







**TMUX7219M** SCDS455 - MAY 2022

# TMUX7219M 44-V, Latch-Up Immune, Extended Temperature, 2:1 (SPDT) Precision Switch with 1.8-V Logic

#### 1 Features

Dual supply range: ±4.5 V to ±22 V Single supply range: 4.5 V to 44 V

-55°C to +125°C operating temperature

Low on-resistance: 2.1  $\Omega$ Low charge injection: -10 pC

High current support: 330 mA (maximum)

Latch-up immune

1.8 V logic compatible

Integrated pull-up and pull-down resistor on logic

Fail-safe logic

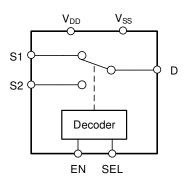
Rail-to-rail operation

Bidirectional signal path

Break-before-make switching

## 2 Applications

- Avionics flight control unit
- Aircraft cockpit display
- Standalone avionics precision flight control
- Interconnect and distribution box
- Aerospace and defense



**Functional Block Diagram** 

## 3 Description

The TMUX7219M is a complementary metal-oxide semiconductor (CMOS) switch with latch-up immunity in a single channel, 2:1 (SPDT) configuration. The device works with a single supply (4.5 V to 44 V), dual supplies (±4.5 V to ±22 V), or asymmetric supplies (such as  $V_{DD}$  = 12 V,  $V_{SS}$  = -5 V). The TMUX7219M supports bidirectional analog and digital signals on the source (Sx) and drain (D) pins ranging from  $V_{SS}$  to  $V_{DD}$ .

The TMUX7219M can be enabled or disabled by controlling the EN pin. When disabled, both signal path switches are off. When enabled, the SEL pin can be used to turn on signal path 1 (S1 to D) or signal path 2 (S2 to D). All logic control inputs support logic levels from 1.8 V to V<sub>DD</sub>, ensuring both TTL and CMOS logic compatibility when operating in the valid supply voltage range. Fail-Safe Logic circuitry allows voltages on the control pins to be applied before the supply pin, protecting the device from potential damage.

The TMUX72xx family provides latch-up immunity, preventing undesirable high current events between parasitic structures within the device typically caused by overvoltage events. A latch-up condition typically continues until the power supply rails are turned off and can lead to device failure. The latch-up immunity feature allows the TMUX72xx family of switches and multiplexers to be used in harsh environments. Additionally, the TMUX7219M is rated for extended temperature use down to -55°C, making it ideal for harsh industrial and aerospace applications.

### Device Information<sup>(1)</sup>

DEVICE NUMBER	PACKAGE	BODY SIZE (NOM)
TMUX7219M	VSSOP (8)	3.00 mm × 3.00 mm

For all available packages, see the package option addendum at the end of the data sheet.



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# **4 Revision History**

DATE	REVISION	NOTES
May 2022	*	Initial Release



# **5 Pin Configuration and Functions**

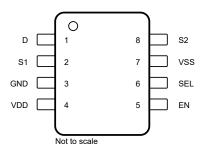


Figure 5-1. DGK Package 8-Pin VSSOP (Top View)

Table 5-1. Pin Functions

PIN		TYPE <sup>(1)</sup>	DESCRIPTION <sup>(2)</sup>
NAME	NO.	TYPE\"	DESCRIPTION
D	1	I/O	Drain pin. Can be an input or output.
S1	2	I/O	Source pin 1. Can be an input or output.
GND	3	Р	Ground (0 V) reference
V <sub>DD</sub>	4	Р	Positive power supply. This pin is the most positive power-supply potential. For reliable operation, connect a decoupling capacitor ranging from 0.1 $\mu$ F to 10 $\mu$ F between $V_{DD}$ and GND.
EN	5	I	Active high logic enable, has internal pull-up resistor. When this pin is low, all switches are turned off. When this pin is high, the SEL logic input determine which switch is turned on.
SEL	6	I	Logic control input, has internal pull-down resistor. Controls the switch connection as shown in Section 8.5.
V <sub>SS</sub>	7	Р	Negative power supply. This pin is the most negative power-supply potential. In single-supply applications, this pin can be connected to ground. For reliable operation, connect a decoupling capacitor ranging from 0.1 µF to 10 µF between V <sub>SS</sub> and GND.
S2	8	I/O	Source pin 2. Can be an input or output.

- (1) I = input, O = output, I/O = input and output, P = power.
- (2) Refer to Section 8.4 for what to do with unused pins.



## **6 Specifications**

## 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1) (2)

		MIN	MAX	UNIT
V <sub>DD</sub> – V <sub>SS</sub>			48	V
V <sub>DD</sub>	Supply voltage	-0.5	48	V
V <sub>SS</sub>		-48	0.5	V
V <sub>SEL</sub> or V <sub>EN</sub>	Logic control input pin voltage (SEL, EN) <sup>(3)</sup>	-0.5	48	V
I <sub>SEL</sub> or I <sub>EN</sub>	Logic control input pin current (SEL, EN) <sup>(3)</sup>	-30	30	mA
V <sub>S</sub> or V <sub>D</sub>	Source or drain voltage (Sx, D) <sup>(3)</sup>	V <sub>SS</sub> -0.5	V <sub>DD</sub> +0.5	V
I <sub>IK</sub>	Diode clamp current <sup>(3)</sup>	-30	30	mA
I <sub>S</sub> or I <sub>D (CONT)</sub>	Source or drain continuous current (Sx, D)		I <sub>DC</sub> + 10 % <sup>(4)</sup>	mA
T <sub>A</sub>	Ambient temperature	-55	150	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C
TJ	Junction temperature		150	°C
P <sub>tot</sub>	Total power dissipation <sup>(5)</sup>		460	mW

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute maximum ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If briefly operating outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not sustain damage, but it may not be fully functional. Operating the device in this manner may affect device reliability, functionality, performance, and shorten the device lifetime.
- (2) All voltages are with respect to ground, unless otherwise specified.
- (3) Pins are diode-clamped to the power-supply rails. Over voltage signals must be voltage and current limited to maximum ratings.
- (4) Refer to Source or Drain Continuous Current table for I<sub>DC</sub> specifications.
- (5) For DGK package:  $P_{tot}$  derates linearily above  $T_A = 70^{\circ}\text{C}$  by 6.7mW/°C.

## 6.2 ESD Ratings

			VALUE	UNIT
V Flootroototia diaabarra	Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins <sup>(1)</sup>	±2000	\/	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC JS-002, all pins <sup>(2)</sup>	±500	<b>v</b>

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

#### 6.3 Thermal Information

		TMUX7219M	
	THERMAL METRIC(1)	DGK (VSSOP)	UNIT
		8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	152.1	°C/W
R <sub>0</sub> JC(top)	Junction-to-case (top) thermal resistance	48.4	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	73.2	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	4.1	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	71.8	°C/W
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	N/A	°C/W

 For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

Product Folder Links: TMUX7219M



# **6.4 Recommended Operating Conditions**

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM MAX	UNIT
V <sub>DD</sub> – V <sub>SS</sub> (1)	Power supply voltage differential	4.5	44	V
$V_{DD}$	Positive power supply voltage	4.5	44	V
V <sub>S</sub> or V <sub>D</sub>	Signal path input/output voltage (source or drain pin) (Sx, D)	V <sub>SS</sub>	$V_{DD}$	V
V <sub>SEL</sub> or V <sub>EN</sub>	Address or enable pin voltage	0	44	V
I <sub>S</sub> or I <sub>D (CONT)</sub>	Source or drain continuous current (Sx, D)		I <sub>DC</sub> <sup>(2)</sup>	mA
T <sub>A</sub>	Ambient temperature	-55	125	°C

 $V_{DD}$  and  $V_{SS}$  can be any value as long as 4.5 V  $\leq$  ( $V_{DD} - V_{SS}$ )  $\leq$  44 V, and the minimum  $V_{DD}$  is met. Refer to *Source or Drain Continuous Current* table for  $I_{DC}$  specifications.

## **6.5 Source or Drain Continuous Current**

at supply voltage of V<sub>DD</sub> ± 10%, V<sub>SS</sub> ± 10 % (unless otherwise noted)

CONTINI	CONTINUOUS CURRENT PER CHANNEL (I <sub>DC</sub> )		T <sub>Δ</sub> = 85°C	T <sub>A</sub> = 125°C	UNIT
PACKAGE	TEST CONDITIONS	T <sub>A</sub> = 25°C	1A - 65 C	1A - 125 C	ONII
	+44 V Single Supply <sup>(1)</sup>	330	210	120	mA
	±15 V Dual Supply	330	210	120	mA
DGK (VSSOP)	+12 V Single Supply	240	160	100	mA
	±5 V Dual Supply	240	160	100	mA
	+5 V Single Supply	180	120	80	mA

<sup>(1)</sup> Specified for nominal supply voltage only.



# 6.6 ±15 V Dual Supply: Electrical Characteristics

 $V_{DD}$  = +15 V ± 10%,  $V_{SS}$  = -15 V ±10%, GND = 0 V (unless otherwise noted) Typical at  $V_{DD}$  = +15 V,  $V_{SS}$  = -15 V,  $T_A$  = 25°C (unless otherwise noted)

rypiodi d	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
ANALOG	SWITCH					I	
		V <sub>S</sub> = -10 V to +10 V	25°C		2.1	2.9	Ω
R <sub>ON</sub>	On-resistance	$I_D = -10 \text{ mA}$	-40°C to +85°C			3.8	Ω
		Refer to On-Resistance	-55°C to +125°C			4.5	Ω
		V <sub>S</sub> = -10 V to +10 V	25°C	4.5   9   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5   1.5	Ω		
$\Delta R_{ON}$	On-resistance mismatch between channels	$I_D = -10 \text{ mA}$	-40°C to +85°C			0.3	Ω
	Chamieis	Refer to On-Resistance	-55°C to +125°C			0.35	Ω
		V <sub>S</sub> = -10 V to +10 V	25°C		0.5	0.6	Ω
R <sub>ON FLAT</sub>	On-resistance flatness	$I_S = -10 \text{ mA}$	-40°C to +85°C			0.7	Ω
		Refer to On-Resistance	-55°C to +125°C			2.1 2.9 3.8 4.5 0.05 0.25 0.3 0.35 0.5 0.6 0.7 0.85 0.01 0.05 0.15 1.6 1.5 0.05 1 1.8 18 18 18 18 18 18 18 18 18 18 18 18 18	Ω
R <sub>ON DRIFT</sub>	On-resistance drift	V <sub>S</sub> = 0 V, I <sub>S</sub> = -10 mA Refer to On-Resistance	–55°C to +125°C		0.01		Ω/°C
		V <sub>DD</sub> = 16.5 V, V <sub>SS</sub> = -16.5 V	25°C	-0.15	0.05	0.15	nA
I <sub>S(OFF)</sub>	Source off leakage current <sup>(1)</sup>	Switch state is off $V_S = +10 \text{ V} / -10 \text{ V}$	-40°C to +85°C	-1.6		1.6	nA
S(OFF)	V <sub>D</sub> = -10 V / + 10 V Refer to Off-Leakage Current	-55°C to +125°C	-15		15	nA	
	Drain off leakage current <sup>(1)</sup>	$V_{DD}$ = 16.5 V, $V_{SS}$ = -16.5 V Switch state is off $V_{S}$ = +10 V / -10 V $V_{D}$ = -10 V / + 10 V Refer to Off-Leakage Current	25°C	-1	0.05	1	nA
Invoces			-40°C to +85°C	-3		3	nA
-D(OFF)			-55°C to +125°C	-26		26	nA
		V <sub>DD</sub> = 16.5 V, V <sub>SS</sub> = -16.5 V	25°C	-1	0.04	1	nA
$I_{S(ON)}$ $I_{D(ON)}$	Channel on leakage current <sup>(2)</sup>	Switch state is on $V_S = V_D = \pm 10 \text{ V}$	-40°C to +85°C	-1.8		1.8	nA
-D(ON)		Refer to On-Leakage Current	–55°C to +125°C	-18		18	nA
LOGIC INF	PUTS (SEL / EN pins)						
V <sub>IH</sub>	Logic voltage high		-55°C to +125°C	1.3		44	V
V <sub>IL</sub>	Logic voltage low		–55°C to +125°C	0		0.8	V
I <sub>IH</sub>	Input leakage current		-55°C to +125°C		0.005	2	μA
I <sub>IL</sub>	Input leakage current		–55°C to +125°C	-1	-0.005		μA
C <sub>IN</sub>	Logic input capacitance		–55°C to +125°C		3		pF
POWER S	UPPLY						
			25°C		30	40	μA
$I_{DD}$	V <sub>DD</sub> supply current	$V_{DD} = 16.5 \text{ V}, V_{SS} = -16.5 \text{ V}$ Logic inputs = 0 V, 5 V, or $V_{DD}$	-40°C to +85°C			48	μΑ
			–55°C to +125°C			62	μΑ
			25°C		3	10	μΑ
I <sub>SS</sub>	V <sub>SS</sub> supply current	$V_{DD} = 16.5 \text{ V}, V_{SS} = -16.5 \text{ V}$ Logic inputs = 0 V, 5 V, or $V_{DD}$	-40°C to +85°C			15	μΑ
			-55°C to +125°C			25	μA

 <sup>(1)</sup> When V<sub>S</sub> is positive, V<sub>D</sub> is negative, or when V<sub>S</sub> is negative, V<sub>D</sub> is positive.
 (2) When V<sub>S</sub> is at a voltage potential, V<sub>D</sub> is floating, or when V<sub>D</sub> is at a voltage potential, V<sub>S</sub> is floating.



# 6.7 ±15 V Dual Supply: Switching Characteristics

 $V_{DD}$  = +15 V ± 10%,  $V_{SS}$  = -15 V ±10%, GND = 0 V (unless otherwise noted) Typical at  $V_{DD}$  = +15 V,  $V_{SS}$  = -15 V,  $T_A$  = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
		V <sub>S</sub> = 10 V	25°C		120	175	ns
TRAN	Transition time from control input	$R_L = 300 \Omega$ , $C_L = 35 pF$	-40°C to +85°C			190	ns
		Refer to Transition Time	-55°C to +125°C			210	ns
		V <sub>S</sub> = 10 V	25°C		100	170	ns
ON (EN)	Turn-on time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C			185	ns
		Refer to Turn-on and Turn-off Time	-55°C to +125°C			120 175 190 210 170 185 200 100 180 195 210 50  0.19 0.2 0.22 700107555117106 40 0.1864 2005 33 48	ns
		V <sub>S</sub> = 10 V	25°C		100	180	ns
OFF (EN)	Turn-off time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C			195	ns
		Refer to Turn-on and Turn-off Time	-55°C to +125°C			210	ns
		V <sub>S</sub> = 10 V,	25°C		50		ns
ВВМ	Break-before-make time delay	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C	1			ns
		Refer to Break-Before-Make	-55°C to +125°C	1	200 100 180 195 210 50 0.19 0.2 0.22 700 -10 -75 -55 -117	ns	
		V rice time = 100mg	25°C		0.19		ms
T <sub>ON (VDD)</sub>	Device turn on time	$V_{DD}$ rise time = 100ns $R_L$ = 300 $\Omega$ , $C_L$ = 35 pF	-40°C to +85°C		0.2		ms
, ,	(V <sub>DD</sub> to output)	Refer to Turn-on (VDD) Time	-55°C to +125°C		200 100 180 195 210 50 1 1 1 0.19 0.2 0.22 700 -10 -75 -55 -117 -106 40	ms	
t <sub>PD</sub>	Propagation delay	$R_L = 50 \Omega$ , $C_L = 5 pF$ Refer to Propagation Delay	25°C		700		ps
Q <sub>INJ</sub>	Charge injection	V <sub>D</sub> = 0 V, C <sub>L</sub> = 1 nF Refer to Charge Injection	25°C		-10		рС
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 100 kHz$ Refer to Off Isolation	25°C		-75		dB
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 1 MHz$ Refer to Off Isolation	25°C		-55		dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 100 kHz$ Refer to Crosstalk	25°C		-117		dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 1 MHz$ Refer to Crosstalk	25°C		-106		dB
BW	-3dB Bandwidth	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ Refer to Bandwidth	25°C		40		MHz
L	Insertion loss	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 1 MHz$	25°C		-0.18		dB
ACPSRR	AC Power Supply Rejection Ratio	$V_{PP}$ = 0.62 V on $V_{DD}$ and $V_{SS}$ $R_L$ = 50 $\Omega$ , $C_L$ = 5 pF, f = 1 MHz Refer to ACPSRR	25°C		-64		dB
THD+N	Total Harmonic Distortion + Noise	$V_{PP}=15$ V, $V_{BIAS}=0$ V $R_L=10$ k $\Omega$ , $C_L=5$ pF, $f=20$ Hz to 20 kHz Refer to THD + Noise	25°C		0.0005		%
C <sub>S(OFF)</sub>	Source off capacitance	V <sub>S</sub> = 0 V, f = 1 MHz	25°C		33		pF
C <sub>D(OFF)</sub>	Drain off capacitance	V <sub>S</sub> = 0 V, f = 1 MHz	25°C		48		pF
C <sub>S(ON)</sub> , C <sub>D(ON)</sub>	On capacitance	V <sub>S</sub> = 0 V, f = 1 MHz	25°C		148		pF



# 6.8 ±20 V Dual Supply: Electrical Characteristics

 $V_{DD} = +20 \text{ V} \pm 10\%, \ V_{SS} = -20 \text{ V} \pm 10\%, \ \text{GND} = 0 \text{ V} \ \text{(unless otherwise noted)}$  Typical at  $V_{DD} = +20 \text{ V}, \ V_{SS} = -20 \text{ V}, \ T_A = 25^{\circ}\text{C} \ \text{(unless otherwise noted)}$ 

rypiodi d	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
ANALOG	SWITCH						
		V. = 15 V to +15 V	25°C		1.9	2.7	Ω
R <sub>ON</sub>	On-resistance	$I_D = -10 \text{ mA}$	-40°C to +85°C			3.5	Ω
		Refer to On-Resistance	-55°C to +125°C			4.2	Ω
		V. = 15 V to +15 V	25°C		0.04	1.9 2.7 3.5 4.2 0.04 0.22 0.28 0.3 0.3 0.75 0.9 1.2 009 0.05 1.5 4 24 0.1 2 8 44 0.1 2 5 29	Ω
$R_{ON}  \text{On-resistance}  \begin{array}{c} V_S = -15 \text{ V to } +15 \text{ V} \\ I_D = -10 \text{ mA} \\ \text{Refer to On-Resistance} \\ \end{array}  \begin{array}{c} -40^{\circ}\text{C to} \\ -55^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} 25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -40^{\circ}\text{C to} \\ -55^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -40^{\circ}\text{C to} \\ -55^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -40^{\circ}\text{C to} \\ -55^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -40^{\circ}\text{C to} \\ -55^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -40^{\circ}\text{C to} \\ -55^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -40^{\circ}\text{C to} \\ -55^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C} \\ \end{array} \\ \begin{array}{c} -25^{\circ}\text{C to} \\ \end{array} \\ \begin{array}{c} -25^{\circ}$	-40°C to +85°C			0.28	Ω		
	Sita mole	Refer to On-Resistance	-55°C to +125°C			0.3	Ω
		V <sub>2</sub> = -15 V to +15 V	25°C		0.3	0.75	Ω
R <sub>ON FLAT</sub>	On-resistance flatness	$I_S = -10 \text{ mA}$	-40°C to +85°C			0.9	Ω
		Refer to On-Resistance	-55°C to +125°C			1.9 2.7 3.5 4.2 0.04 0.22 0.28 0.3 0.3 0.75 0.9 1.2 0.009 0.05 1.5 4 24 0.1 2 8 44 0.1 2 5 29 44 0.8 0.005 2 0.005 3 34 44 50 65 4 9	Ω
R <sub>ON DRIFT</sub>	On-resistance drift		-55°C to +125°C		0.009		Ω/°C
			25°C	-1.5	0.05	1.5	nA
le(OEE)	Source off leakage current <sup>(1)</sup>		-40°C to +85°C	-4		4	nA
S(OFF)	coarse on rounage carrons	$V_D = -15 \text{ V} / + 15 \text{ V}$	-55°C to +125°C	-24		24	nA
	Drain off leakage current <sup>(1)</sup>	Switch state is off $V_S = +15 \text{ V} / -15 \text{ V}$ $V_D = -15 \text{ V} / + 15 \text{ V}$	25°C	-2	0.1	2	nA
I <sub>D(OFF)</sub> Drain off leakage current <sup>(1)</sup>			-40°C to +85°C	-8		8	nA
			-55°C to +125°C	-44		44	nA
			25°C	-2	0.1	2	nA
	Channel on leakage current <sup>(2)</sup>		-40°C to +85°C	-5		5	nA
-D(ON)			-55°C to +125°C	-29		29	nA
LOGIC INF	PUTS (SEL / EN pins)			•			
V <sub>IH</sub>	Logic voltage high		-55°C to +125°C	1.3		44	V
V <sub>IL</sub>	Logic voltage low		-55°C to +125°C	0		0.8	V
I <sub>IH</sub>	Input leakage current		-55°C to +125°C		0.005	2	μA
I <sub>IL</sub>	Input leakage current		-55°C to +125°C	-1	-0.005		μA
C <sub>IN</sub>	Logic input capacitance		-55°C to +125°C		3		pF
POWER S	UPPLY		·				
			25°C		34	44	μA
$I_{DD}$	V <sub>DD</sub> supply current	$V_{DD} = 22 \text{ V}, V_{SS} = -22 \text{ V}$ Logic inputs = 0 V, 5 V, or $V_{DD}$	-40°C to +85°C			50	μΑ
		3	-55°C to +125°C			65	μΑ
			25°C		4	9	μΑ
I <sub>SS</sub>	V <sub>SS</sub> supply current	$V_{DD}$ = 22 V, $V_{SS}$ = -22 V Logic inputs = 0 V, 5 V, or $V_{DD}$	-40°C to +85°C			12	μΑ
			-55°C to +125°C			25	μA

 <sup>(1)</sup> When V<sub>S</sub> is positive, V<sub>D</sub> is negative, or when V<sub>S</sub> is negative, V<sub>D</sub> is positive.
 (2) When V<sub>S</sub> is at a voltage potential, V<sub>D</sub> is floating, or when V<sub>D</sub> is at a voltage potential, V<sub>S</sub> is floating.



# 6.9 ±20 V Dual Supply: Switching Characteristics

 $V_{DD} = +20 \text{ V} \pm 10\%, \ V_{SS} = -20 \text{ V} \pm 10\%, \ \text{GND} = 0 \text{ V} \ \text{(unless otherwise noted)}$  Typical at  $V_{DD} = +20 \text{ V}, \ V_{SS} = -20 \text{ V}, \ T_A = 25^{\circ}\text{C} \ \text{(unless otherwise noted)}$ 

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
		V <sub>S</sub> = 10 V	25°C		110	175	ns
TRAN	Transition time from control input	$R_L = 300 \Omega$ , $C_L = 35 pF$	-40°C to +85°C			190	ns
		Refer to Transition Time	-55°C to +125°C			205	ns
		V <sub>S</sub> = 10 V	25°C		110	170	ns
ON (EN)	Turn-on time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C			185	ns
		Refer to Turn-on and Turn-off Time	-55°C to +125°C			200	ns
		V <sub>S</sub> = 10 V	25°C		90	180	ns
OFF (EN)	Turn-off time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C			190	ns
		Refer to Turn-on and Turn-off Time	-55°C to +125°C			200	ns
		V <sub>S</sub> = 10 V,	25°C		55		ns
ВВМ	Break-before-make time delay	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C	1			ns
		Refer to Break-Before-Make	-55°C to +125°C	1			ns
		V <sub>DD</sub> rise time = 100ns	25°C		0.18		ms
T <sub>ON (VDD)</sub>	Device turn on time (V <sub>DD</sub> to output)	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C		0.2		ms
	(VDD to surpar)	Refer to Turn-on (VDD) Time	-55°C to +125°C		0.22		ms
t <sub>PD</sub>	Propagation delay	$R_L = 50 \Omega$ , $C_L = 5 pF$ Refer to Propagation Delay	25°C		715		ps
Q <sub>INJ</sub>	Charge injection	V <sub>D</sub> = 0 V, C <sub>L</sub> = 1 nF Refer to Charge Injection	25°C	-15			рС
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 100 kHz$ Refer to Off Isolation	25°C	-75			dB
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 1 MHz$ Refer to Off Isolation	25°C		<b>–</b> 55		dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 100 kHz$ Refer to Crosstalk	25°C		-117		dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 1 MHz$ Refer to Crosstalk	25°C		-106		dB
BW	-3dB Bandwidth	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , Refer to Bandwidth	25°C		38		MHz
L	Insertion loss	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 0 V$ , $f = 1 MHz$	25°C		-0.16		dB
ACPSRR	AC Power Supply Rejection Ratio	$\begin{array}{l} V_{PP}=0.62~V~on~V_{DD}~and~V_{SS}\\ R_L=50~\Omega~,~C_L=5~pF,\\ f=1~MHz\\ Refer~to~ACPSRR \end{array} \label{eq:VPP}$		-63		dB	
THD+N	Total Harmonic Distortion + Noise	$ \begin{array}{l} V_{PP} = 20 \text{ V}, V_{BIAS} = 0 \text{ V} \\ R_{L} = 10 \text{ k}\Omega, C_{L} = 5 \text{ pF}, \\ f = 20 \text{ Hz to } 20 \text{ kHz} \\ \text{Refer to THD + Noise} \end{array} $		0.0005		%	
C <sub>S(OFF)</sub>	Source off capacitance	V <sub>S</sub> = 0 V, f = 1 MHz	25°C		32		pF
C <sub>D(OFF)</sub>	Drain off capacitance	V <sub>S</sub> = 0 V, f = 1 MHz	25°C		45		pF
C <sub>S(ON),</sub> C <sub>D(ON)</sub>	On capacitance	V <sub>S</sub> = 0 V, f = 1 MHz	25°C		146		pF



# 6.10 44 V Single Supply: Electrical Characteristics

 $V_{DD}$  = +44 V,  $V_{SS}$  = 0 V, GND = 0 V (unless otherwise noted) Typical at  $V_{DD}$  = +44 V,  $V_{SS}$  = 0 V,  $T_A$  = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
ANALOG	SWITCH						
		V <sub>S</sub> = 0 V to 40 V	25°C		2.2	2.8	Ω
R <sub>ON</sub>	On-resistance	$I_D = -10 \text{ mA}$	-40°C to +85°C			3.6	Ω
		Refer to On-Resistance	-55°C to +125°C			4.2	Ω
		V <sub>S</sub> = 0 V to 40 V	25°C		0.1	0.2	Ω
$\Delta R_{ON}$	On-resistance mismatch between channels	$I_D = -10 \text{ mA}$	-40°C to +85°C			0.3	Ω
		Refer to On-Resistance	-55°C to +125°C			0.35	Ω
		V <sub>S</sub> = 0 V to 40 V	25°C		0.2	1	Ω
R <sub>ON FLAT</sub>	On-resistance flatness	$I_{D} = -10 \text{ mA}$	-40°C to +85°C			1.3	Ω
		Refer to On-Resistance	-55°C to +125°C			1.5	Ω
R <sub>ON DRIFT</sub>	On-resistance drift	V <sub>S</sub> = 22 V, I <sub>S</sub> = -10 mA Refer to On-Resistance	-55°C to +125°C		0.008		Ω/°C
	Source off leakage current <sup>(1)</sup>	V <sub>DD</sub> = 44 V, V <sub>SS</sub> = 0 V	25°C	-5	0.05	5	nA
I <sub>S(OFF)</sub>		Switch state is off $V_S = 40 \text{ V} / 1 \text{ V}$	-40°C to +85°C	-10		10	nA
·5(OFF)		V <sub>D</sub> = 1 V / 40 V Refer to Off-Leakage Current	-55°C to +125°C	-35		35	nA
	Drain off leakage current <sup>(1)</sup>	V <sub>DD</sub> = 44 V, V <sub>SS</sub> = 0 V	25°C	-8	0.05	8	nA
I <sub>D(OFF)</sub>		Switch state is off V <sub>S</sub> = 40 V / 1 V	-40°C to +85°C	-12		12	nA
D(O(1)		V <sub>D</sub> = 1 V / 40 V Refer to Off-Leakage Current	-55°C to +125°C	-70		70	nA
		V <sub>DD</sub> = 44 V, V <sub>SS</sub> = 0 V	25°C	-8	0.05	8	nA
I <sub>S(ON)</sub> I <sub>D(ON)</sub>	Channel on leakage current <sup>(2)</sup>	Switch state is on $V_S = V_D = 40 \text{ V or } 1 \text{ V}$	-40°C to +85°C	-10		10	nA
-D(ON)		Refer to On-Leakage Current	-55°C to +125°C	-45		45	nA
LOGIC INF	PUTS (SEL / EN pins)						
V <sub>IH</sub>	Logic voltage high		-55°C to +125°C	1.3		44	V
V <sub>IL</sub>	Logic voltage low		-55°C to +125°C	0		0.8	V
I <sub>IH</sub>	Input leakage current		-55°C to +125°C		0.005	2	μA
I <sub>IL</sub>	Input leakage current		-55°C to +125°C	-1	-0.005		μΑ
C <sub>IN</sub>	Logic input capacitance		-55°C to +125°C		3		pF
POWER S	UPPLY						
			25°C		17	50	μA
$I_{DD}$	V <sub>DD</sub> supply current	$V_{DD}$ = 44 V, $V_{SS}$ = 0 V Logic inputs = 0 V, 5 V, or $V_{DD}$	-40°C to +85°C			60	μA
		3	-55°C to +125°C			75	μA

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<sup>(1)</sup> When V<sub>S</sub> is 40 V, V<sub>D</sub> is 1 V, or when V<sub>S</sub> is 1 V, V<sub>D</sub> is 40 V.
(2) When V<sub>S</sub> is at a voltage potential, V<sub>D</sub> is floating, or when V<sub>D</sub> is at a voltage potential, V<sub>S</sub> is floating.



# 6.11 44 V Single Supply: Switching Characteristics

 $V_{DD}$  = +44 V,  $V_{SS}$  = 0 V, GND = 0 V (unless otherwise noted) Typical at  $V_{DD}$  = +44 V,  $V_{SS}$  = 0 V,  $T_A$  = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	MIN TY	P MAX	UNIT
		V <sub>S</sub> = 18 V	25°C	12	175	ns
t <sub>TRAN</sub>	Transition time from control input	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C		190	ns
		Refer to Transition Time	-55°C to +125°C		205	ns
		V <sub>S</sub> = 18 V	25°C	12	168	ns
t <sub>ON (EN)</sub>	Turn-on time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C		185	ns
		Refer to Turn-on and Turn-off Time	–55°C to +125°C		195	ns
		V <sub>S</sub> = 18 V	25°C	12	180	ns
t <sub>OFF (EN)</sub>	Turn-off time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C		200	ns
		Refer to Turn-on and Turn-off Time	–55°C to +125°C		205	ns
		V <sub>S</sub> = 18 V,	25°C	4	5	ns
t <sub>BBM</sub>	Break-before-make time delay	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C	1		ns
		Refer to Break-Before-Make	-55°C to +125°C	1		ns
		V rice time = 1ue	25°C	0.1	5	ms
T <sub>ON (VDD)</sub>	Device turn on time	$V_{DD}$ rise time = 1μs $R_L$ = 300 Ω, $C_L$ = 35 pF	-40°C to +85°C	0.1	7	ms
. ,	(V <sub>DD</sub> to output)	Refer to Turn-on (VDD) Time	-55°C to +125°C	0.1	9	ms
t <sub>PD</sub>	Propagation delay	$R_L = 50 \Omega$ , $C_L = 5 pF$ Refer to Propagation Delay	25°C	93	)	ps
Q <sub>INJ</sub>	Charge injection	V <sub>D</sub> = 22 V, C <sub>L</sub> = 1 nF Refer to Charge Injection				pC
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 100 kHz$ Refer to Off Isolation	25°C -75		5	dB
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 1 MHz$ Refer to Off Isolation	25°C	-5	5	dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 100 kHz$ Refer to Crosstalk	25°C	-11	7	dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 1 MHz$ Refer to Crosstalk	25°C	-10	3	dB
BW	-3dB Bandwidth	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ Refer to Bandwidth	25°C	3	7	MHz
IL	Insertion loss	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 1 MHz$	25°C	-0.1	3	dB
ACPSRR	AC Power Supply Rejection Ratio	$V_{PP}$ = 0.62 V on $V_{DD}$ and $V_{SS}$ R <sub>L</sub> = 50 $\Omega$ , C <sub>L</sub> = 5 pF, f = 1 MHz Refer to ACPSRR		-6	)	dB
THD+N	Total Harmonic Distortion + Noise	$\begin{array}{c} V_{PP} = 22 \ V, V_{BIAS} = 22 \ V \\ R_{L} = 10 \ k\Omega \ , C_{L} = 5 \ pF, \\ f = 20 \ Hz \ to \ 20 \ kHz \\ Refer \ to \ THD + Noise \end{array} \hspace{2cm} 25^{\circ}C \end{array} \hspace{2cm} ($		0.000	1	%
C <sub>S(OFF)</sub>	Source off capacitance	V <sub>S</sub> = 6 V, f = 1 MHz	25°C	3	1	pF
C <sub>D(OFF)</sub>	Drain off capacitance	V <sub>S</sub> = 6 V, f = 1 MHz	25°C	4	3	pF
C <sub>S(ON)</sub> , C <sub>D(ON)</sub>	On capacitance	V <sub>S</sub> = 6 V, f = 1 MHz	25°C	14	6	pF



# 6.12 12 V Single Supply: Electrical Characteristics

 $V_{DD}$  = +12 V ± 10%,  $V_{SS}$  = 0 V, GND = 0 V (unless otherwise noted) Typical at  $V_{DD}$  = +12 V,  $V_{SS}$  = 0 V,  $T_A$  = 25°C (unless otherwise noted)

. , p	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
ANALOG	SWITCH	<u> </u>					
		V <sub>S</sub> = 0 V to 10 V	25°C		4.6	6	Ω
R <sub>ON</sub>	On-resistance	$I_D = -10 \text{ mA}$	-40°C to +85°C			7.5	Ω
		Refer to On-Resistance	-55°C to +125°C			8.4	Ω
		V <sub>S</sub> = 0 V to 10 V	25°C		0.08	0.2	Ω
$\Delta R_{ON}$	On-resistance mismatch between channels	$I_{D} = -10 \text{ mA}$	-40°C to +85°C			0.32	Ω
	Grannels	Refer to On-Resistance	-55°C to +125°C			0.35	Ω
		V <sub>S</sub> = 0 V to 10 V	25°C		1.2	2	Ω
R <sub>ON FLAT</sub>	On-resistance flatness	$I_{S} = -10 \text{ mA}$	-40°C to +85°C			2.2	Ω
		Refer to On-Resistance	-55°C to +125°C			2.4	Ω
R <sub>ON DRIFT</sub>	On-resistance drift	V <sub>S</sub> = 6 V, I <sub>S</sub> = -10 mA Refer to On-Resistance	-55°C to +125°C		0.017		Ω/°C
	Source off leakage current <sup>(1)</sup>	V <sub>DD</sub> = 13.2 V, V <sub>SS</sub> = 0 V	25°C	-0.5	0.05	0.5	nA
I <sub>S(OFF)</sub>		Switch state is off V <sub>S</sub> = 10 V / 1 V	-40°C to +85°C	-2		2	nA
·3(OFF)		V <sub>D</sub> = 1 V / 10 V Refer to Off-Leakage Current	-55°C to +125°C	-12		12	nA
	Drain off leakage current <sup>(1)</sup>	V <sub>DD</sub> = 13.2 V, V <sub>SS</sub> = 0 V	25°C	-0.5	0.05	0.5	nA
I <sub>D(OFF)</sub>		Switch state is off V <sub>S</sub> = 10 V / 1 V	-40°C to +85°C	-3		3	nA
·D(OFF)		V <sub>D</sub> = 1 V / 10 V Refer to Off-Leakage Current	-55°C to +125°C	-23		23	nA
		V <sub>DD</sub> = 13.2 V, V <sub>SS</sub> = 0 V	25°C	-1.5	0.05	1.5	nA
$I_{S(ON)}$ $I_{D(ON)}$	Channel on leakage current <sup>(2)</sup>	Switch state is on $V_S = V_D = 10 \text{ V or } 1 \text{ V}$	-40°C to +85°C	-3		3	nA
-D(ON)		Refer to On-Leakage Current	-55°C to +125°C	-15		15	nA
LOGIC INF	PUTS (SEL / EN pins)					·	
V <sub>IH</sub>	Logic voltage high		-55°C to +125°C	1.3		44	V
V <sub>IL</sub>	Logic voltage low		-55°C to +125°C	0		0.8	V
I <sub>IH</sub>	Input leakage current		-55°C to +125°C		0.005	2	μA
I <sub>IL</sub>	Input leakage current		-55°C to +125°C	-1	-0.005		μΑ
C <sub>IN</sub>	Logic input capacitance		-55°C to +125°C		3		pF
POWER S	UPPLY						
		40.0444 511	25°C		10	35	μΑ
$I_{DD}$	V <sub>DD</sub> supply current	$V_{DD}$ = 13.2 V, $V_{SS}$ = 0 V Logic inputs = 0 V, 5 V, or $V_{DD}$	-40°C to +85°C			45	μΑ
		3 , 3 , 3 , 3 , 3 , 3	-55°C to +125°C			55	μA

<sup>(1)</sup> When  $V_S$  is 10 V,  $V_D$  is 1 V, or when  $V_S$  is 1 V,  $V_D$  is 10 V.

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When  $V_S$  is at a voltage potential,  $V_D$  is floating, or when  $V_D$  is at a voltage potential,  $V_S$  is floating.



# 6.13 12 V Single Supply: Switching Characteristics

 $V_{DD}$  = +12 V ± 10%,  $V_{SS}$  = 0 V, GND = 0 V (unless otherwise noted) Typical at  $V_{DD}$  = +12 V,  $V_{SS}$  = 0 V,  $T_A$  = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
		V <sub>S</sub> = 8 V	25°C		180	185	ns
t <sub>TRAN</sub>	Transition time from control input	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C			215	ns
		Refer to Transition Time	-55°C to +125°C			235	ns
		V <sub>S</sub> = 8 V	25°C		120	180	ns
t <sub>ON (EN)</sub>	Turn-on time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C			210	ns
		Refer to Turn-on and Turn-off Time	-55°C to +125°C			230	ns
		V <sub>S</sub> = 8 V	25°C		130	210	ns
t <sub>OFF (EN)</sub>	Turn-off time from enable	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C			235	ns
		Refer to Turn-on and Turn-off Time	-55°C to +125°C			250	ns
		V <sub>S</sub> = 8 V,	25°C		40		ns
ввм	Break-before-make time delay	$R_L = 300 \Omega, C_L = 35 pF$	-40°C to +85°C	1			ns
		Refer to Break-Before-Make	-55°C to +125°C	1			ns
		V rice time = 100pe	25°C		0.19		ms
T <sub>ON (VDD)</sub>	Device turn on time	$V_{DD}$ rise time = 100ns R <sub>L</sub> = 300 $\Omega$ , C <sub>L</sub> = 35 pF	-40°C to +85°C		0.2		ms
` ,	(V <sub>DD</sub> to output)	Refer to Turn-on (VDD) Time	-55°C to +125°C		0.22		ms
t <sub>PD</sub>	Propagation delay	$R_L = 50 \Omega$ , $C_L = 5 pF$ Refer to Propagation Delay					ps
Q <sub>INJ</sub>	Charge injection	V <sub>D</sub> = 6 V, C <sub>L</sub> = 1 nF Refer to Charge Injection	25°C	-6			рС
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 100 kHz$ Refer to Off Isolation	25°C	-75			dB
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 1 MHz$ 25°C Refer to Off Isolation		-55		dB	
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 100 kHz$ Refer to Crosstalk	25°C		-117		dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 1MHz$ Refer to Crosstalk	25°C		-106		dB
BW	-3dB Bandwidth	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ Refer to Bandwidth	25°C		42		MHz
I <sub>L</sub>	Insertion loss	$R_L = 50 \Omega$ , $C_L = 5 pF$ $V_S = 6 V$ , $f = 1 MHz$	25°C		-0.3		dB
ACPSRR	AC Power Supply Rejection Ratio	$\begin{array}{l} V_{PP}=0.62~V~on~V_{DD}~and~V_{SS}\\ R_L=50~\Omega~,~C_L=5~pF,\\ f=1~MHz\\ Refer~to~ACPSRR \end{array} \hspace{2cm} 25^{\circ}C \hspace{1cm} -$		-65		dB	
THD+N	Total Harmonic Distortion + Noise	$\begin{array}{c} V_{PP}=6~V,~V_{BIAS}=6~V\\ R_L=10~k\Omega~,~C_L=5~pF,\\ f=20~Hz~to~20~kHz\\ Refer~to~THD~+~Noise \end{array} \hspace{2cm} 25^{\circ}C \\ \end{array}$			0.0009		%
C <sub>S(OFF)</sub>	Source off capacitance	V <sub>S</sub> = 6 V, f = 1 MHz	25°C		38		pF
C <sub>D(OFF)</sub>	Drain off capacitance	V <sub>S</sub> = 6 V, f = 1 MHz	25°C		56		pF
C <sub>S(ON)</sub> , C <sub>D(ON)</sub>	On capacitance	V <sub>S</sub> = 6 V, f = 1 MHz	25°C		150		pF



## 6.14 Typical Characteristics

at  $T_A = 25$ °C

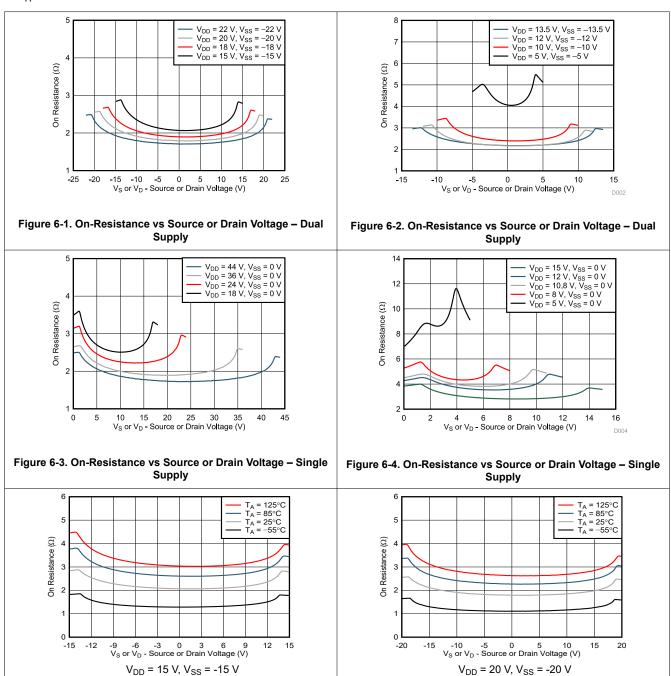
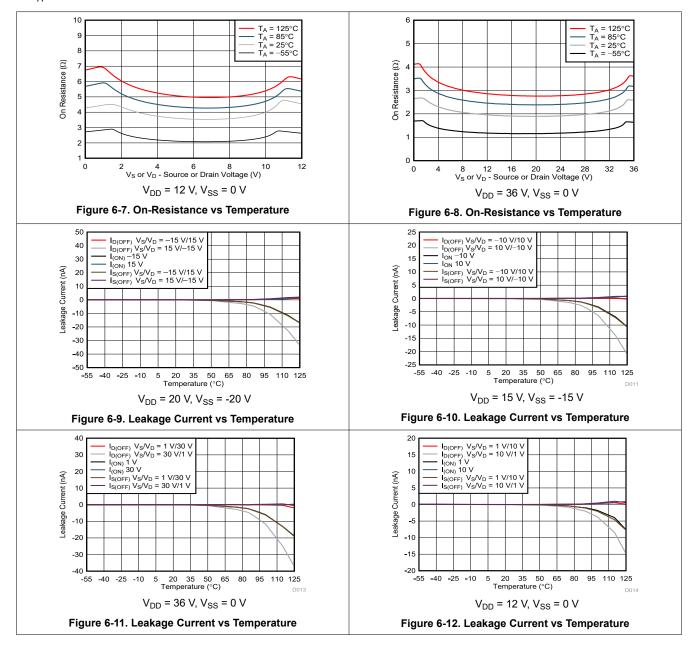


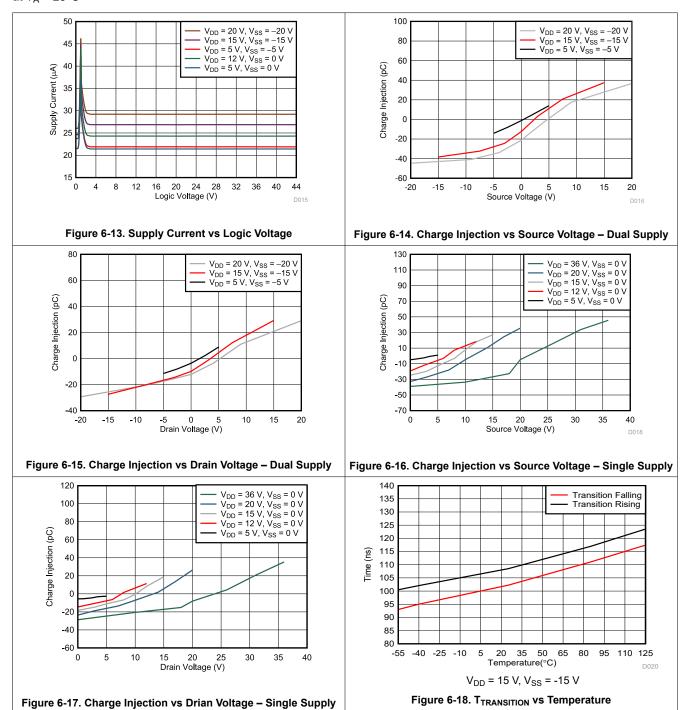
Figure 6-5. On-Resistance vs Temperature

Figure 6-6. On-Resistance vs Temperature

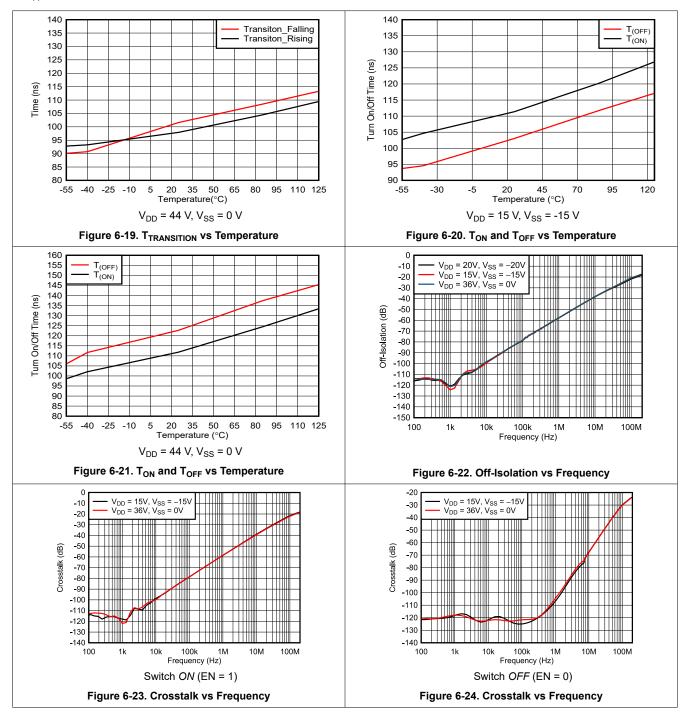




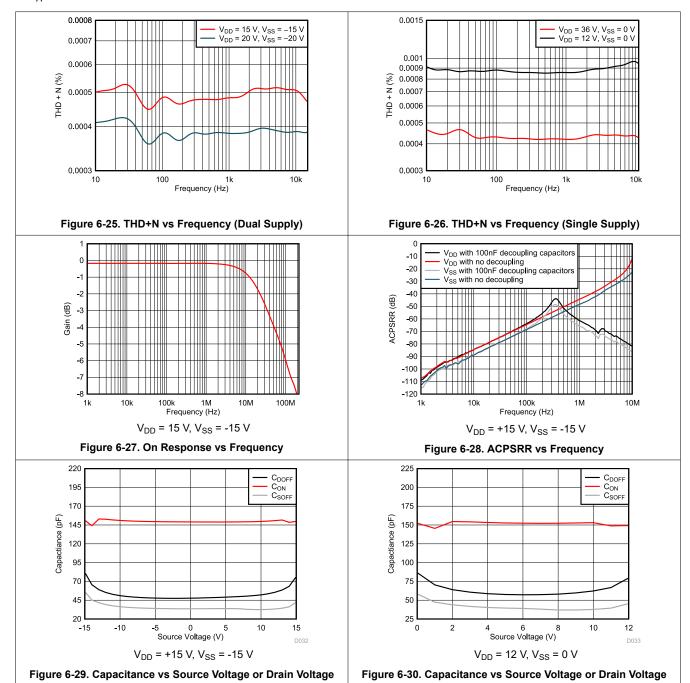












### 7 Parameter Measurement Information

### 7.1 On-Resistance

The on-resistance of a device is the ohmic resistance between the source (Sx) and drain (D) pins of the device. The on-resistance varies with input voltage and supply voltage. The symbol  $R_{ON}$  is used to denote on-resistance. Figure 7-1 shows the measurement setup used to measure  $R_{ON}$ . Voltage (V) and current ( $I_{SD}$ ) are measured using the following setup, where  $R_{ON}$  is computed as  $R_{ON} = V / I_{SD}$ :

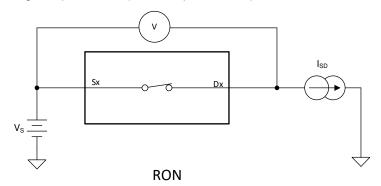


Figure 7-1. On-Resistance

## 7.2 Off-Leakage Current

There are two types of leakage currents associated with a switch during the off state:

- 1. Source off-leakage current.
- 2. Drain off-leakage current.

Source leakage current is defined as the leakage current flowing into or out of the source pin when the switch is off. This current is denoted by the symbol  $I_{S(OFF)}$ .

Drain leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is off. This current is denoted by the symbol  $I_{D(OFF)}$ .

Figure 7-2 shows the setup used to measure both off-leakage currents.

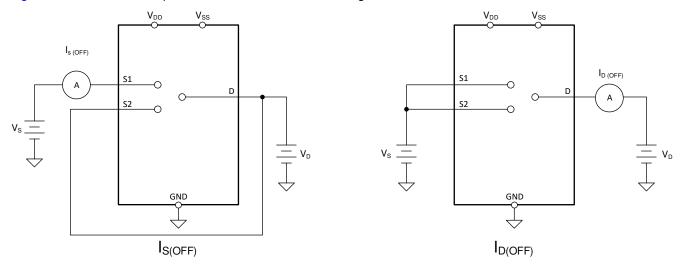


Figure 7-2. Off-Leakage Measurement Setup



### 7.3 On-Leakage Current

Source on-leakage current is defined as the leakage current flowing into or out of the source pin when the switch is on. This current is denoted by the symbol  $I_{S(ON)}$ .

Drain on-leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is on. This current is denoted by the symbol  $I_{D(ON)}$ .

Either the source pin or drain pin is left floating during the measurement. Figure 7-3 shows the circuit used for measuring the on-leakage current, denoted by  $I_{S(ON)}$  or  $I_{D(ON)}$ .

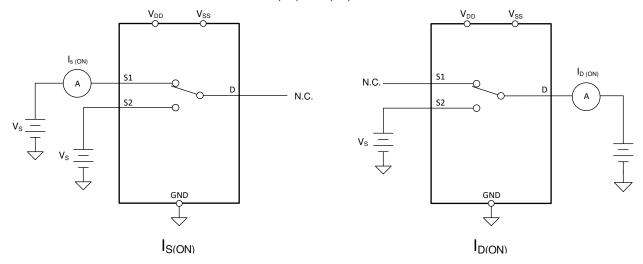


Figure 7-3. On-Leakage Measurement Setup

#### 7.4 Transition Time

Transition time is defined as the time taken by the output of the device to rise or fall 90% after the address signal has risen or fallen past the logic threshold. The 90% transition measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. Figure 7-4 shows the setup used to measure transition time, denoted by the symbol  $t_{TRANSITION}$ .

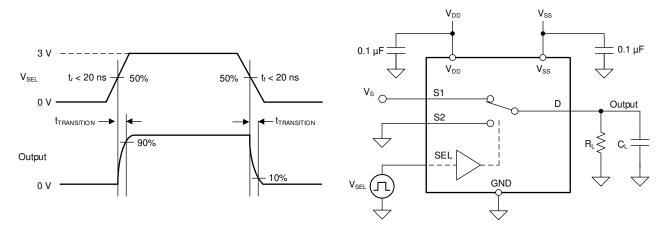


Figure 7-4. Transition-Time Measurement Setup

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7.5 t<sub>ON(EN)</sub> and t<sub>OFF(EN)</sub>

Turn-on time is defined as the time taken by the output of the device to rise to 90% after the enable has risen past the logic threshold. The 90% measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. Figure 7-5 shows the setup used to measure turn-on time, denoted by the symbol  $t_{ON(FN)}$ .

Turn-off time is defined as the time taken by the output of the device to fall to 10% after the enable has fallen past the logic threshold. The 10% measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. Figure 7-5 shows the setup used to measure turn-off time, denoted by the symbol t<sub>OFF(EN)</sub>.

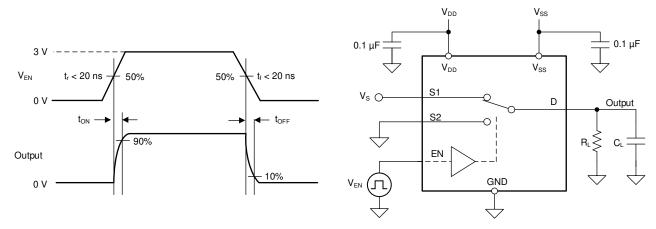


Figure 7-5. Turn-On and Turn-Off Time Measurement Setup

#### 7.6 Break-Before-Make

Break-before-make delay is a safety feature that prevents two inputs from connecting when the device is switching. The output first breaks from the on-state switch before making the connection with the next on-state switch. The time delay between the *break* and the *make* is known as break-before-make delay. Figure 7-6 shows the setup used to measure break-before-make delay, denoted by the symbol t<sub>OPEN(BBM)</sub>.

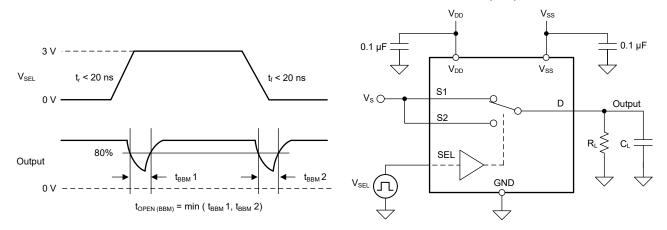


Figure 7-6. Break-Before-Make Delay Measurement Setup



# 7.7 t<sub>ON (VDD)</sub> Time

The  $t_{ON\ (VDD)}$  time is defined as the time taken by the output of the device to rise to 90% after the supply has risen past the supply threshold. The 90% measurement is used to provide the timing of the device turning on in the system. Figure 7-7 shows the setup used to measure turn on time, denoted by the symbol  $t_{ON\ (VDD)}$ .

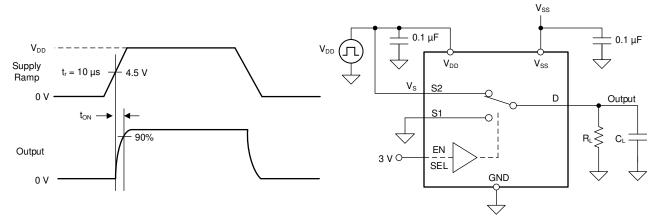


Figure 7-7. t<sub>ON (VDD)</sub> Time Measurement Setup

## 7.8 Propagation Delay

Propagation delay is defined as the time taken by the output of the device to rise or fall 50% after the input signal has risen or fallen past the 50% threshold. Figure 7-8 shows the setup used to measure propagation delay, denoted by the symbol  $t_{PD}$ .

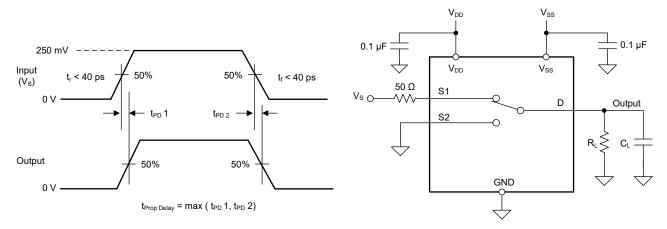


Figure 7-8. Propagation Delay Measurement Setup

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### 7.9 Charge Injection

The TMUX7219M has a transmission-gate topology. Any mismatch in capacitance between the NMOS and PMOS transistors results in a charge injected into the drain or source during the falling or rising edge of the gate signal. The amount of charge injected into the source or drain of the device is known as charge injection, and is denoted by the symbol  $Q_C$ . Figure 7-9 shows the setup used to measure charge injection from source (Sx) to drain (D).

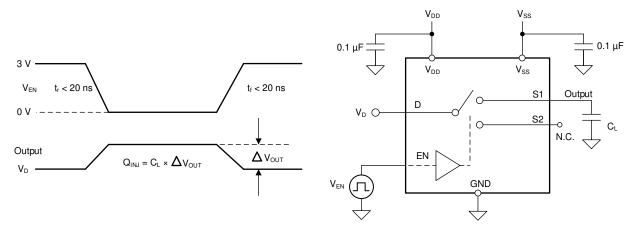


Figure 7-9. Charge-Injection Measurement Setup

#### 7.10 Off Isolation

Off isolation is defined as the ratio of the signal at the drain pin (D) of the device when a signal is applied to the source pin (Sx) of an off-channel. Figure 7-10 shows the setup used to measure, and the equation used to calculate off isolation.

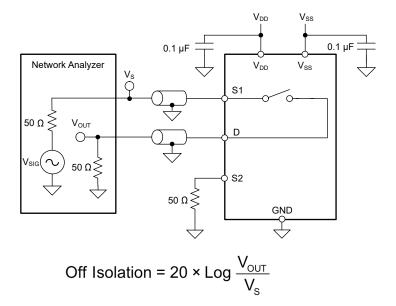


Figure 7-10. Off Isolation Measurement Setup



### 7.11 Crosstalk

Crosstalk is defined as the ratio of the signal at the drain pin (D) of a different channel, when a signal is applied at the source pin (Sx) of an on-channel. Figure 7-11 shows the setup used to measure, and the equation used to calculate crosstalk.

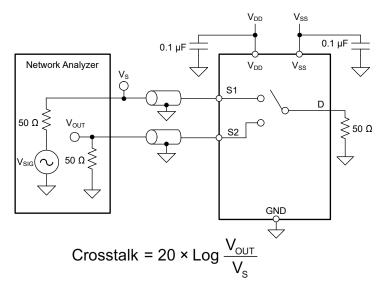


Figure 7-11. Crosstalk Measurement Setup

### 7.12 Bandwidth

Bandwidth is defined as the range of frequencies that are attenuated by less than 3 dB when the input is applied to the source pin (Sx) of an on-channel, and the output is measured at the drain pin (D) of the device. Figure 7-12 shows the setup used to measure bandwidth.

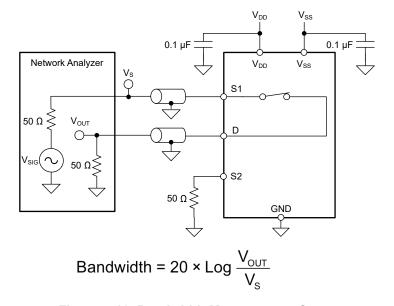


Figure 7-12. Bandwidth Measurement Setup

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#### 7.13 THD + Noise

The total harmonic distortion (THD) of a signal is a measurement of the harmonic distortion, and is defined as the ratio of the sum of the powers of all harmonic components to the power of the fundamental frequency at the mux output.

The on-resistance of the device varies with the amplitude of the input signal and results in distortion when the drain pin is connected to a low-impedance load. Total harmonic distortion plus noise is denoted as THD + N.

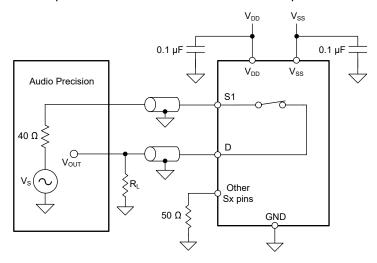
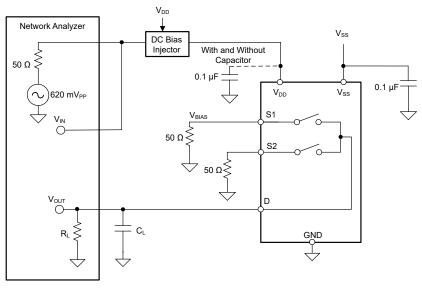


Figure 7-13. THD + N Measurement Setup

# 7.14 Power Supply Rejection Ratio (PSRR)

PSRR measures the ability of a device to prevent noise and spurious signals that appear on the supply voltage pin from coupling to the output of the switch. The DC voltage on the device supply is modulated by a sine wave of 620 mV $_{\rm PP}$ . The ratio of the amplitude of signal on the output to the amplitude of the modulated signal is the ACPSRR. A high ratio represents a high degree of tolerance to supply rail variation.

This helps stabilize the supply and immediately filter as much of the supply noise as possible.



 $PSRR = 20 \times Log \frac{V_{OUT}}{V_{IN}}$ 

Figure 7-14. ACPSRR Measurement Setup



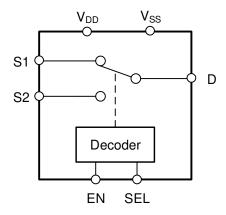
## 8 Detailed Description

#### 8.1 Overview

The TMUX7219M is a 2:1, 1-channel switch. Each input is turned on or turned off based on the state of the select line and enable pin.

## 8.2 Functional Block Diagram

The following figure shows the functional block diagram of the TMUX7219M.



## 8.3 Feature Description

### 8.3.1 Bidirectional Operation

The TMUX7219M conducts equally well from source (Sx) to drain (D) or from drain (D) to source (Sx). Each channel has very similar characteristics in both directions and supports both analog and digital signals.

#### 8.3.2 Rail-to-Rail Operation

The valid signal path input and output voltage for TMUX7219M ranges from  $V_{SS}$  to  $V_{DD}$ .

## 8.3.3 1.8 V Logic Compatible Inputs

The TMUX7219M has 1.8 V logic compatible control for all logic control inputs. 1.8 V logic level inputs allows the device to interface with processors that have lower logic I/O rails and eliminates the need for an external translator, which saves both space and BOM cost. For more information on 1.8 V logic implementations refer to Simplifying Design with 1.8 V logic Muxes and Switches.

#### 8.3.4 Integrated Pull-Up and Pull-Down Resistor on Logic Pins

The TMUX7219M has internal weak pull-up and pull-down resistors to GND to ensure the logic pins are not left floating. The value of this pull-down resistor is approximately 4 M $\Omega$ , but is clamped to about 1  $\mu$ A at higher voltages. The EN pin integrates a pull-up resistor to  $V_{DD}$  and the SEL pin integrates a pull-down resistor. This feature integrates up to two external components and reduces system size and cost.

#### 8.3.5 Fail-Safe Logic

The TMUX7219M supports Fail-Safe Logic on the control input pins (EN and SEL) allowing for operation up to 44 V above ground, regardless of the state of the supply pins. This feature allows voltages on the control pins to be applied before the supply pin, protecting the device from potential damage. Fail-Safe Logic minimizes system complexity by removing the need for power supply sequencing on the logic control pins. For example, the Fail-Safe Logic feature allows the logic input pins of the TMUX7219M to be ramped to +44 V while  $V_{DD}$  and  $V_{SS}$  = 0 V. The logic control inputs are protected against positive faults of up to +44 V in powered-off condition, but do not offer protection against negative overvoltage conditions.

Product Folder Links: TMUX7219M



#### 8.3.6 Latch-Up Immune

Latch-up is a condition where a low impedance path is created between a supply pin and ground. This condition is caused by a trigger (current injection or overvoltage), but once activated, the low impedance path remains even after the trigger is no longer present. This low impedance path may cause system upset or catastrophic damage due to excessive current levels. The latch-up condition typically requires a power cycle to eliminate the low impedance path.

The TMUX72xx family of devices are constructed on Silicon on Insulator (SOI) based process where an oxide layer is added between the PMOS and NMOS transistor of each CMOS switch to prevent parasitic structures from forming. The oxide layer is also known as an insulating trench and prevents triggering of latch up events due to overvoltage or current injections. The latch-up immunity feature allows the TMUX72xx family of switches and multiplexers to be used in harsh environments. For more information on latch-up immunity refer to *Using Latch Up Immune Multiplexers to Help Improve System Reliability*.

### 8.3.7 Ultra-Low Charge Injection

Figure 8-1 shows how the TMUX7219M has a transmission gate topology. Any mismatch in the stray capacitance associated with the NMOS and PMOS causes an output level change whenever the switch is opened or closed.

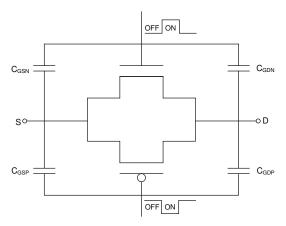


Figure 8-1. Transmission Gate Topology

The TMUX7219M contains specialized architecture to reduce charge injection on the source (Sx). To further reduce charge injection in a sensitive application, a compensation capacitor (Cp) can be added on the drain (D). This will ensure that excess charge from the switch transition will be pushed into the compensation capacitor on the drain (D) instead of the source (Sx). As a general rule, Cp should be 20× larger than the equivalent load capacitance on the source (Sx). Figure 8-2 shows charge injection variation with source voltage with different compensation capacitors on the drain side.

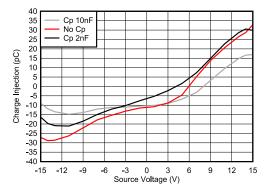


Figure 8-2. Charge Injection Compensation



#### 8.4 Device Functional Modes

When the EN pin of the TMUX7219M is pulled high, one of the switches is closed based on the state of the SEL pin. When the EN pin is pulled low, both of the switches are in an open state regardless of the state of the SEL pin. The control pins can be as high as 44 V.

The TMUX7219M can operate without any external components except for the supply decoupling capacitors. The EN pin has an internal pull-up resistor of 4 M $\Omega$ , and SEL pin has internal pull-down resistor of 4 M $\Omega$ . If unused, EN pin must be tied to V<sub>DD</sub> and SEL pin must be tied to GND to ensure the device does not consume additional current as highlighted in *Implications of Slow or Floating CMOS Inputs*. Unused signal path inputs (S1, S2, or D) should be connected to GND.

#### 8.5 Truth Tables

Table 8-1 show the truth tables for the TMUX7219M.

Table 8-1, TMUX7219M Truth Table

EN	SEL	Selected Source Connected To Drain (D) Pin
0	X <sup>(1)</sup>	All sources are off (HI-Z)
1	0	S1
1	1	S2

(1) X denotes do not care.

Product Folder Links: TMUX7219M

# 9 Application and Implementation

#### **Note**

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

## 9.1 Application Information

TMUX7219M is part of the precision switches and multiplexers family of devices. TMUX7219M offers low RON, low on and off leakage currents, and ultra-low charge injection performance. These properties make TMUX7219M ideal for implementing high precision industrial systems requiring selection of one of two inputs or outputs.

#### 9.2 Typical Applications

#### 9.2.1 Data Acquisition Calibration

One application of the TMUX7219M is in Data Acquisition systems (DAQ). To account for system loss and ensure the lowest possible noise floor, a calibration path is needed. To minimize board space and automate this procedure, many applications utilize a 2:1 (SPDT) switch. Figure 9-1 shows the TMUX7219M configured for switching a calibration path on a precision measurement module.

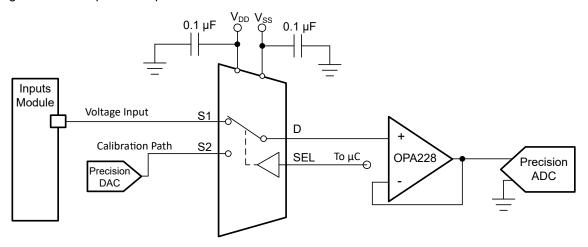


Figure 9-1. Calibration Path Switching for Data Acquisition

# 9.2.1.1 Design Requirements

For the design example, use the parameters listed in Table 9-1.

 PARAMETERS
 VALUES

 Supply (V<sub>DD</sub>)
 15 V

 Supply (V<sub>SS</sub>)
 -15 V

 MUX I/O signal range
 -15 V to 15 V (Rail-to-Rail)

 Control logic thresholds
 1.8 V compatiable (up to 44V)

 EN
 EN pulled high to enable the switch

Table 9-1. Design Parameters



#### 9.2.1.2 Detailed Design Procedure

The TMUX7219M can be operated without any external components except for the supply decoupling capacitors. All inputs passing through the switch must fall within the recommended operating conditions, including signal range and continuous current. For this design, with a dual supply of ±15 V, the signal range can range from -15 V to +15 V. Industrial applications such as factory automation and control and test and measurement benefit from using a 2:1 switch, because it allows additional flexibility in the design. A single 2:1 switch has numerous applications such as switching between an analog signal path and a calibration path, and allowing a single channel to be configured as either an analog input or analog output.

# 10 Power Supply Recommendations

The TMUX7219M operates across a wide supply range of  $\pm 4.5$  V to  $\pm 22$  V (4.5 V to 44 V in single-supply mode). The device also performs well with asymmetrical supplies such as  $V_{DD} = 12$  V and  $V_{SS} = -5$  V.

Power-supply bypassing improves noise margin and prevents switching noise propagation from the supply rails to other components. Good power-supply decoupling is important to achieve optimum performance. For improved supply noise immunity, use a supply decoupling capacitor ranging from 0.1  $\mu$ F to 10  $\mu$ F at both the V<sub>DD</sub> and V<sub>SS</sub> pins to ground. Place the bypass capacitors as close to the power supply pins of the device as possible using low-impedance connections. TI recommends using multi-layer ceramic chip capacitors (MLCCs) that offer low equivalent series resistance (ESR) and inductance (ESL) characteristics for power-supply decoupling purposes. For very sensitive systems, or for systems in harsh noise environments, avoiding the use of vias for connecting the capacitors to the device pins may offer superior noise immunity. The use of multiple vias in parallel lowers the overall inductance and is beneficial for connections to ground and power planes. Always ensure the ground (GND) connection is established before supplies are ramped.

Product Folder Links: TMUX7219M



## 11 Layout

## 11.1 Layout Guidelines

When a PCB trace turns a corner at a 90° angle, a reflection can occur. A reflection occurs primarily because of the change of width of the trace. At the apex of the turn, the trace width increases to 1.414 times the width. This increase upsets the transmission-line characteristics, especially the distributed capacitance and self–inductance of the trace which results in the reflection. Not all PCB traces can be straight and therefore some traces must turn corners. Figure 11-1 shows progressively better techniques of rounding corners. Only the last example (BEST) maintains constant trace width and minimizes reflections.

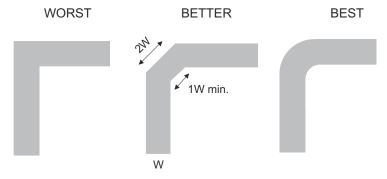


Figure 11-1. Trace Example

Route high-speed signals using a minimum of vias and corners which reduces signal reflections and impedance changes. When a via must be used, increase the clearance size around it to minimize its capacitance. Each via introduces discontinuities in the signal's transmission line and increases the chance of picking up interference from the other layers of the board. Be careful when designing test points, through-hole pins are not recommended at high frequencies.

Figure 11-2 illustrates an example of a PCB layout with the TMUX7219M. Some key considerations are:

- For reliable operation, connect a decoupling capacitor ranging from 0.1 μF to 10 μF between VDD/VSS and GND. TI recommends placing the lowest value capacitor as close to the pin as possible. Make sure that the capacitor voltage rating is sufficient for the supply voltage.
- Keep the input lines as short as possible.
- Use a solid ground plane to help reduce electromagnetic interference (EMI) noise pickup.
- Do not run sensitive analog traces in parallel with digital traces. Avoid crossing digital and analog traces if possible, and only make perpendicular crossings when necessary.
- Using multiple vias in parallel will lower the overall inductance and is beneficial for connection to ground planes.

#### 11.2 Layout Example

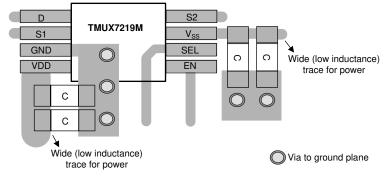


Figure 11-2. TMUX7219MDGK Layout Example



# 12 Device and Documentation Support

## **12.1 Documentation Support**

#### 12.1.1 Related Documentation

For related documentation, see the following:

- · Texas Instruments, Improve Stability Issues with Low CON Multiplexers application brief
- Texas Instruments, Improving Signal Measurement Accuracy in Automated Test Equipment application brief
- · Texas Instruments, Multiplexers and Signal Switches Glossary application report
- Texas Instruments, QFN/SON PCB Attachment application report
- Texas Instruments, Quad Flatpack No-Lead Logic Packages application report
- Texas Instruments, Simplifying Design with 1.8 V logic Muxes and Switches application brief
- Texas Instruments, System-Level Protection for High-Voltage Analog Multiplexers application report

## 12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 12.3 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### 12.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.

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#### 12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 12.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

# 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

Product Folder Links: TMUX7219M

www.ti.com 9-Nov-2025

#### PACKAGING INFORMATION

Orderable part number	Status (1)	Material type	Package   Pins	Package qty   Carrier	RoHS (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
TMUX7219MDGKR	Active	Production	VSSOP (DGK)   8	2500   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-55 to 125	X219
TMUX7219MDGKR.B	Active	Production	VSSOP (DGK)   8	2500   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-55 to 125	X219

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

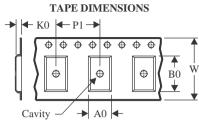
<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

# **PACKAGE MATERIALS INFORMATION**

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## TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

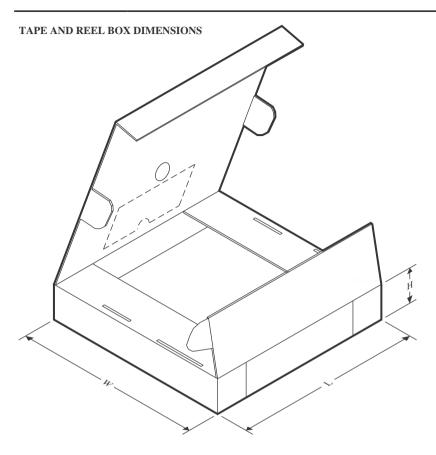


#### \*All dimensions are nominal

	Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
١	TMUX7219MDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.25	3.35	1.25	8.0	12.0	Q1

**PACKAGE MATERIALS INFORMATION** 

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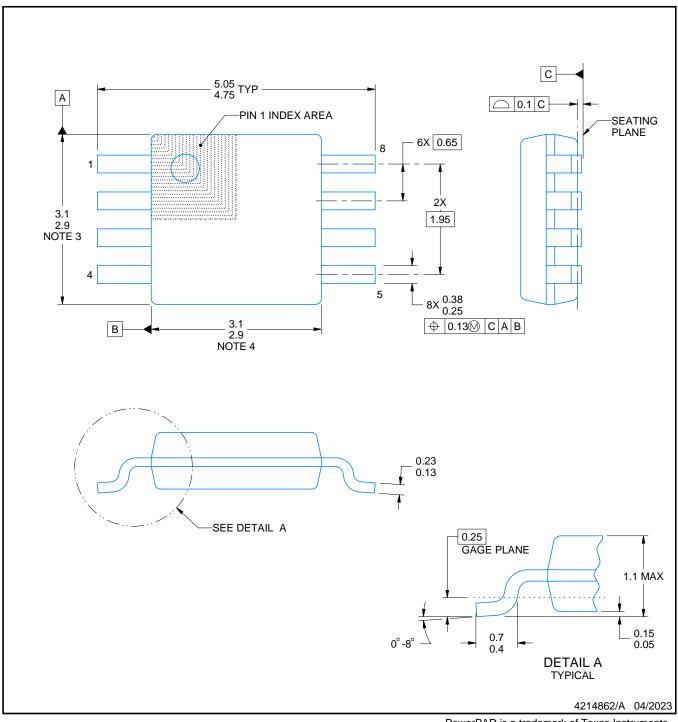


### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TMUX7219MDGKR	VSSOP	DGK	8	2500	366.0	364.0	50.0



SMALL OUTLINE PACKAGE



#### NOTES:

PowerPAD is a trademark of Texas Instruments.

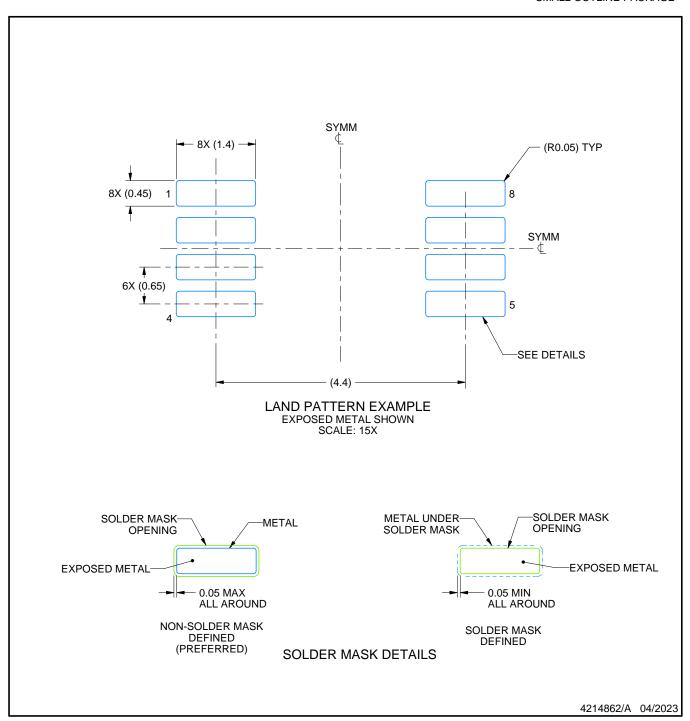
- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-187.



SMALL OUTLINE PACKAGE

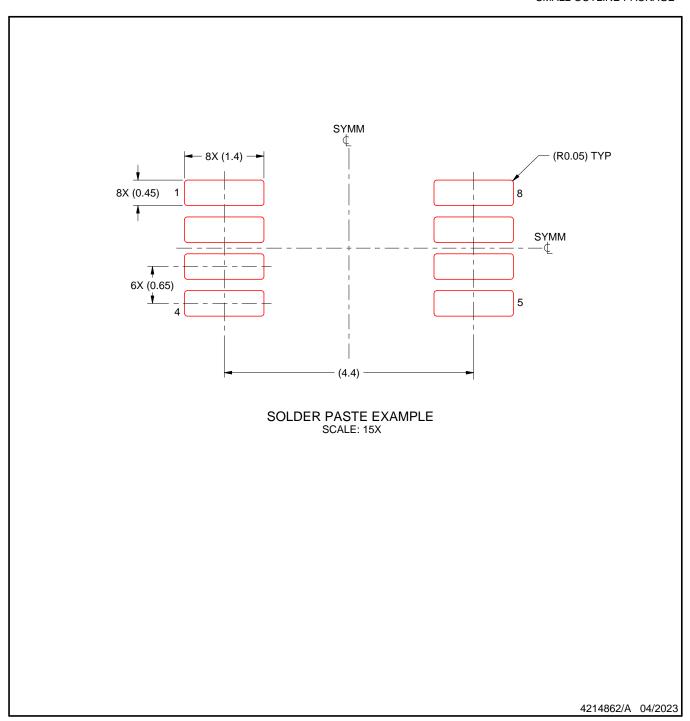


NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
- 8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
- 9. Size of metal pad may vary due to creepage requirement.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 12. Board assembly site may have different recommendations for stencil design.



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Last updated 10/2025