





SCDS406A - DECEMBER 2018 - REVISED FEBRUARY 2024

**TMUX1109** 

# TMUX1109 5V, ±2.5V, Low-Leakage-Current, 4:1, 2-Channel Precision Multiplexer

### 1 Features

Single supply range: 1.08V to 5.5V

Dual supply range: ±2.75V Low leakage current: 3pA Low charge injection: 1pC Low on-resistance: 1.8Ω

-40°C to +125°C operating temperature

1.8V logic compatible

Fail-safe logic

Rail to rail operation

Bidirectional signal path

Break-before-make switching

ESD protection HBM: 2000V

# 2 Applications

Ultrasound scanners

Patient monitoring and diagnostics

Optical networking

Optical test equipment

Remote radio unit

Wired networking

ATE test equipment

Factory automation and industrial controls

Programmable logic controllers (PLC)

Analog input modules

Sonar receivers

Motor drive

Servo drive position feedback

#### 4 Channels per ADC 1kΩ Risot 4:1 10Ω Differential ADC A MUX Rf2 1kΩ ADS9224R Dual, Simultaneous-Sampling Low-Latency SAR ADC TMUX1109 1kΩ AINP B 10Ω 4.1 Differential MUX AINM B Rf2 1kΩ 4 Channels per ADC

Application Example

## 3 Description

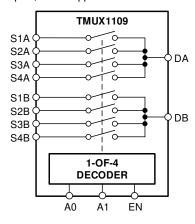
The TMUX1109 is a precision complementary metaloxide semiconductor (CMOS) multiplexer (MUX). The TMUX1109 offers differential 4:1 or dual 4:1 single-ended channels. Wide operating supply of 1.08V to 5.5V allows for use in a broad array of applications from medical equipment to industrial systems. The device supports bidirectional analog and digital signals on the source (Sx) and drain (D) pins ranging from GND to V<sub>DD</sub>. All logic inputs have 1.8V logic compatible thresholds, allowing for both TTL and CMOS logic compatibility when operating in the valid supply voltage range. Fail-Safe Logic circuitry allows voltages on the control pins to be applied before the supply pin, protecting the device from potential damage.

The TMUX1109 is part of the precision switches and multiplexers family of devices. These devices have very low on and off leakage currents and low charge injection, allowing them to be used in high precision measurement applications. A low supply current of 8nA and small package options enable use in portable applications.

#### **Package Information**

PART NUMBER	JMBER PACKAGE <sup>(1)</sup> PACKAGE SIZ		
TMUX1109	PW (TSSOP, 16)	5mm × 6.4mm	
TMUXTTU9	RSV (QFN, 16)	2.6mm × 1.8mm	

- For more information, see Section 10 (1)
- The package size (length × width) is a nominal value and includes pins, where applicable.



**Block Diagram** 



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# **4 Pin Configuration and Functions**

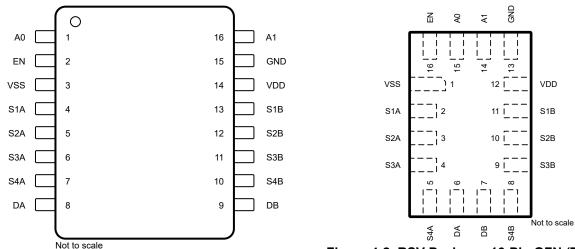


Figure 4-1. PW Package, 16-Pin TSSOP (Top View)

Figure 4-2. RSV Package, 16-Pin QFN (Top View)

**Table 4-1. Pin Functions** 

	PIN		TYPE(1)	DECODINATION	
NAME	TSSOP	UQFN	I TPE(''	DESCRIPTION	
A0	1	15	I	Address line 0. Controls the switch configuration as listed in Table 6-1.	
EN	2	16	1	Active high logic input. When this pin is low, all switches are turned off. When this pin is high, the A[1:0] address inputs determine which switch is turned on.	
VSS	3	1	Р	Negative power supply. This pin is the most negative power-supply potential. For reliable operation, connect a decoupling capacitor ranging from $0.1\mu F$ to $10\mu F$ between $V_{SS}$ and GNI $V_{SS}$ can be connected to ground for single supply applications.	
S1A	4	2	I/O	Source pin 1A. Can be an input or output.	
S2A	5	3	I/O	Source pin 2A. Can be an input or output.	
S3A	6	4	I/O	Source pin 3A. Can be an input or output.	
S4A	7	5	I/O	Source pin 4A. Can be an input or output.	
DA	8	6	I/O	Drain pin A. Can be an input or output.	
DB	9	7	I/O	Drain pin B. Can be an input or output.	
S4B	10	8	I/O	Source pin 4B. Can be an input or output.	
S3B	11	9	I/O	Source pin 3B. Can be an input or output.	
S2B	12	10	I/O	Source pin 2B. Can be an input or output.	
S1B	13	11	I/O	Source pin 1B. Can be an input or output.	
VDD	14	12	Р	Positive power supply. This pin is the most positive power-supply potential. For reliable operation, connect a decoupling capacitor ranging from $0.1\mu F$ to $10\mu F$ between $V_{DD}$ and GND.	
GND	15	13	Р	Ground (0V) reference	
A1	16	14	I	Address line 1. Controls the switch configuration as listed in Table 6-1.	

(1) I = input, O = output, I/O = input and output, P = power



# **5 Specifications**

## 5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1) (2) (3)

		MIN	MAX	UNIT
V <sub>DD</sub> -V <sub>SS</sub>		-0.5	6	V
V <sub>DD</sub>	Supply voltage	-0.5	6	V
V <sub>SS</sub>		-3.0	0.3	V
V <sub>SEL</sub> or V <sub>EN</sub>	Logic control input pin voltage (EN, A0, A1)	-0.5	6	V
I <sub>SEL</sub> or I <sub>EN</sub>	Logic control input pin current (EN, A0, A1)	-30	30	mA
V <sub>S</sub> or V <sub>D</sub>	Source or drain voltage (Sx, Dx)	-0.5	V <sub>DD</sub> +0.5	V
I <sub>S</sub> or I <sub>D (CONT)</sub>	Source or drain continuous current (Sx, Dx)	IDC ± 10 % <sup>(4)</sup>	IDC ± 10 % <sup>(4)</sup>	mA
Is or I <sub>D (PEAK)</sub>	Source and drain peak current: (1 ms period max, 10% duty cycle maximum) (Sx, SxA, SxB, D, DA, DB)	Ipeak ± 10 % <sup>(4)</sup>	Ipeak ± 10 % <sup>(4)</sup>	mA
T <sub>stg</sub>	Storage temperature	-65	150	°C
P <sub>tot</sub>	Total power dissipation <sup>(5)</sup> (6)		500	mW
T <sub>J</sub>	Junction temperature		150	°C

- (1) Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) The algebraic convention, whereby the most negative value is a minimum and the most positive value is a maximum.
- (3) All voltages are with respect to ground, unless otherwise specified.
- 4) Refer to Recommended Operating Conditions for I<sub>DC</sub> and I<sub>Peak</sub> ratings.
- (5) For TSSOP package: P<sub>tot</sub> derates linearly above TA=90°C by 8.41mW/°C
- (6) For QFN package: Ptot derates linearly above TA=82°C by 7.43mW/°C

### 5.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub> Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins <sup>(1)</sup>	±2000	V	
	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>	±750	V	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

## **5.3 Recommended Operating Conditions**

over operating free-air temperature range (unless otherwise noted)

			М	N NOM	MAX	UNIT
$V_{DD}$	Positive power supply voltage (single)	·	1.0	18	5.5	V
V <sub>SS</sub>	Negative power supply voltage (dual)		-2.7	75	0	V
V <sub>DD</sub> - V <sub>SS</sub>	Supply rail voltage difference		1.0	18	5.5	V
V <sub>S</sub> or V <sub>D</sub>	Signal path input/output voltage (source or drain	pin) (Sx, Dx)	V	SS	$V_{DD}$	V
V <sub>SEL</sub> or V <sub>EN</sub>	Logic control input pin voltage			0	5.5	V
T <sub>A</sub>	Ambient temperature		-4	.0	125	°C
		Tj = 25°C		150		mA
	Continuous current through switch	Tj = 85°C		120		mA
IDC		Tj = 125°C		60		mA
	Tj = 130°C			50		mA

# **5.3 Recommended Operating Conditions (continued)**

over operating free-air temperature range (unless otherwise noted)

			MIN NOM	MAX UNIT
Peak current through so duty cycle maximum)		Tj = 25°C	300	mA
	Peak current through switch(1 ms period max, 10%	Tj = 85°C	300	mA
	duty cycle maximum)	Tj = 125°C	180	mA
		Tj = 130°C	160	mA

## **5.4 Thermal Information**

THERMAL METRIC <sup>(1)</sup>		TMU		
		PW (TSSOP)	RSV (QFN)	UNIT
		16 PINS	16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	118.9	134.6	°C/W
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	49.3	74.3	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	65.2	62.8	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	7.6	4.3	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	64.6	61.1	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	N/A	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

# 5.5 Electrical Characteristics ( $V_{DD}$ = 5V ±10 %)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
ANALC	G SWITCH		_	<u>'</u>			
		$V_S = 0V \text{ to } V_{DD}$	25°C		1.8	4	Ω
$R_{ON}$	On-resistance	I <sub>SD</sub> = 10mA	–40°C to +85°C			4.5	Ω
		Refer to On-Resistance	-40°C to +125°C			4.9	Ω
		$V_S = 0V \text{ to } V_{DD}$	25°C		0.18		Ω
$\Delta R_{ON}$	On-resistance matching between channels	I <sub>SD</sub> = 10mA	-40°C to +85°C			0.4	Ω
	Giaineis	Refer to On-Resistance	-40°C to +125°C			0.5	Ω
	On-resistance flatness	$V_S = 0V \text{ to } V_{DD}$ $I_{SD} = 10\text{mA}$ Refer to On-Resistance	25°C		0.85		Ω
R <sub>ON</sub>			-40°C to +85°C			1.6	Ω
FLAT			-40°C to +125°C			1.6	Ω
		$V_{DD} = 5V$ Switch Off $V_D = 4.5V / 1.5V$ $V_S = 1.5V / 4.5V$ Refer to Off-Leakage Current	25°C	-0.08	±0.005	0.08	nA
la	Source off leakage current <sup>(1)</sup>		-40°C to +85°C	-0.3		0.3	nA
I <sub>S(OFF)</sub>	Source on leakage current		-40°C to +125°C	-0.9		0.9	nA
		V <sub>DD</sub> = 5V	25°C	-0.1	±0.01	0.1	nA
I	Drain off leakage current <sup>(1)</sup>	Switch Off V <sub>D</sub> = 4.5V / 1.5V	–40°C to +85°C	-0.75		0.75	nA
I <sub>D(OFF)</sub>	Drain oil leakage current	$V_D = 4.5 \text{V} / 1.5 \text{V}$ $V_S = 1.5 \text{V} / 4.5 \text{V}$ Refer to Off-Leakage Current	-40°C to +125°C	-3.5		3.5	nA
		V <sub>DD</sub> = 5V	25°C	-0.025	±0.003	0.025	nA
I <sub>D(ON)</sub>	Channel on leakage current	Switch On $V_D = V_S = 2.5V$ Refer to On-Leakage Current	–40°C to +85°C	-0.3		0.3	nA
I <sub>S(ON)</sub>	_		-40°C to +125°C	-0.75		0.75	nA



# 5.5 Electrical Characteristics (V<sub>DD</sub> = 5V ±10 %) (continued)

at  $T_{\Delta} = 25^{\circ}C$ ,  $V_{DD} = 5V$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
		V <sub>DD</sub> = 5V	25°C	-0.1	±0.01	0.1	nA
I <sub>D(ON)</sub>	Channel on leakage current	Switch On	-40°C to +85°C	-0.75		0.75	nA
I <sub>S(ON)</sub>	-	$V_D = V_S = 4.5V / 1.5V$ Refer to On-Leakage Current	-40°C to +125°C	-3		3	nA
LOGIC	INPUTS (EN, A0, A1)						
V <sub>IH</sub>	Input logic high		10001 10500	1.49		5.5	V
V <sub>IL</sub>	Input logic low		-40°C to +125°C	0		0.87	V
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		25°C		±0.005		μA
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		-40°C to +125°C			±0.05	μA
C <sub>IN</sub>	Logic input capacitance		25°C -40°C to +125°C		1	2	pF pF
POWER	R SUPPLY						<u> </u>
			25°C		0.008		μA
I <sub>DD</sub>	V <sub>DD</sub> supply current	Logic inputs = 0V or 5.5V	-40°C to +125°C			1	μA
DYNAM	IIC CHARACTERISTICS			1			
		V <sub>S</sub> = 3V	25°C		14		ns
t <sub>TRAN</sub>	Transition time between channels	$R_L = 200\Omega$ , $C_L = 15pF$ Refer to Transition Time	-40°C to +85°C			18	ns
			-40°C to +125°C			19	ns
		V <sub>S</sub> = 3V	25°C		8		ns
t <sub>OPEN</sub>	Break before make time	R <sub>L</sub> = 200Ω, C <sub>L</sub> = 15pF Refer to Break-Before-Make	-40°C to +85°C	1			ns
(BBM)			-40°C to +125°C	1			ns
	Enable turn-on time	$V_S = 3V$ $R_L = 200\Omega$ , $C_L = 15pF$ Refer to tON(EN) and tOFF(EN)	25°C		12		ns
t <sub>ON(EN)</sub>			-40°C to +85°C			19	ns
			-40°C to +125°C			20	ns
		V <sub>S</sub> = 3V	25°C		6		ns
t <sub>OFF(EN)</sub>	Enable turn-off time	$R_L = 200\Omega$ , $C_L = 15pF$	–40°C to +85°C			8	ns
		Refer to tON(EN) and tOFF(EN)	-40°C to +125°C			9	ns
$Q_{C}$	Charge Injection	$V_S$ = 1V $R_S$ = 0 $\Omega$ , $C_L$ = 1nF Refer to Charge Injection	25°C		-1		рC
0	Off landstine	$R_L = 50\Omega$ , $C_L = 5pF$ f = 1MHz Refer to Off Isolation	25°C		-65		dB
O <sub>ISO</sub>	Off Isolation	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Off Isolation	25°C		-45		dB
v		R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 1MHz Refer to Crosstalk	25°C		-90		dB
X <sub>TALK</sub>	Crosstalk	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Crosstalk	25°C		-80		dB
BW	Bandwidth	$R_L = 50\Omega$ , $C_L = 5pF$ Refer to Bandwidth	25°C		135		MHz
C <sub>SOFF</sub>	Source off capacitance	f = 1MHz	25°C		7.5		pF
C <sub>DOFF</sub>	Drain off capacitance	f = 1MHz	25°C		32		pF

Product Folder Links: TMUX1109

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# 5.5 Electrical Characteristics (V<sub>DD</sub> = 5V ±10 %) (continued)

at  $T_A = 25$ °C,  $V_{DD} = 5V$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
C <sub>SON</sub> C <sub>DON</sub>	On capacitance	f = 1MHz	25°C		38		pF

<sup>(1)</sup> When  $V_S$  is 4.5V,  $V_D$  is 1.5V, and vice versa.

# 5.6 Electrical Characteristics ( $V_{DD}$ = 3.3V ±10 %)

at  $T_A = 25$ °C,  $V_{DD} = 3.3V$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
ANALC	G SWITCH						
		V <sub>S</sub> = 0V to V <sub>DD</sub>	25°C		4	8.75	Ω
R <sub>ON</sub>	On-resistance	I <sub>SD</sub> = 10mA	-40°C to +85°C			9.5	Ω
		Refer to On-Resistance	-40°C to +125°C			9.75	Ω
		$V_S = 0V \text{ to } V_{DD}$	25°C		0.13		Ω
$\Delta R_{ON}$	On-resistance matching between channels	I <sub>SD</sub> = 10mA	-40°C to +85°C			0.4	Ω
	onarmois .	Refer to On-Resistance	-40°C to +125°C			0.5	Ω
		V <sub>S</sub> = 0V to V <sub>DD</sub>	25°C		1.9		Ω
R <sub>ON</sub>	On-resistance flatness	I <sub>SD</sub> = 10mA	-40°C to +85°C		2		Ω
FLAT		Refer to On-Resistance	-40°C to +125°C		2.2		Ω
		V <sub>DD</sub> = 3.3V	25°C	-0.05	±0.001	0.05	nA
la (acc)	Source off leakage current	Switch Off V <sub>D</sub> = 3V / 1V	-40°C to +85°C	-0.1		0.1	nA
I <sub>S(OFF)</sub>	Source on leakage current	V <sub>S</sub> = 1V / 3V Refer to Off-Leakage Current	-40°C to +125°C	-0.5		0.5	nA
		V <sub>DD</sub> = 3.3V	25°C	-0.1	±0.005	0.1	nA
Invoces	Drain off leakage current	Switch Off V <sub>D</sub> = 3V / 1V	-40°C to +85°C	-0.5		0.5	nA
I <sub>D(OFF)</sub>		V <sub>S</sub> = 1V / 3V Refer to Off-Leakage Current	-40°C to +125°C	-2		2	nA
		V <sub>DD</sub> = 3.3V	25°C	-0.1	±0.005	0.1	nA
$I_{D(ON)}$ $I_{S(ON)}$	Channel on leakage current	Switch On V <sub>D</sub> = V <sub>S</sub> = 3V / 1V	-40°C to +85°C	-0.5		0.5	nA
15(UN)		Refer to On-Leakage Current —40°C to +125°C	-2		2	nA	
LOGIC	INPUTS (EN, A0, A1)		•			'	
V <sub>IH</sub>	Input logic high		-40°C to +125°C	1.35		5.5	V
V <sub>IL</sub>	Input logic low		-40 C to +125 C	0		0.8	V
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		25°C		±0.005		μΑ
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		-40°C to +125°C			±0.05	μΑ
C	Logic input capacitance		25°C		1		pF
C <sub>IN</sub>	Logic input capacitance		-40°C to +125°C			2	pF
POWE	RSUPPLY						
1	V supply current	Logic inputs = 0V or 5.5V	25°C		0.006		μA
IDD	V <sub>DD</sub> supply current	Logic inputs - UV OI 3.3V	-40°C to +125°C			1	μΑ
DYNAN	MIC CHARACTERISTICS						
		V <sub>S</sub> = 2V	25°C		15		ns
t <sub>TRAN</sub>	Transition time between channels	$R_{L} = 200\Omega$ , $C_{L} = 15pF$	-40°C to +85°C			23	ns
		Refer to Transition Time	-40°C to +125°C			23	ns

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# 5.6 Electrical Characteristics (V<sub>DD</sub> = 3.3V ±10 %) (continued)

at  $T_A = 25$ °C,  $V_{DD} = 3.3V$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
		V <sub>S</sub> = 2V	25°C		9		ns
t <sub>OPEN</sub>	Break before make time	$R_{L} = 200\Omega, C_{L} = 15pF$	–40°C to +85°C	1			ns
(BBM)		Refer to Break-Before-Make	-40°C to +125°C	1			ns
		V <sub>S</sub> = 2V	25°C		14		ns
t <sub>ON(EN)</sub>	Enable turn-on time	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C			25	ns
		Refer to tON(EN) and tOFF(EN)	-40°C to +125°C			25	ns
		V <sub>S</sub> = 2V	25°C		7		ns
t <sub>OFF(EN)</sub>	Enable turn-off time	$R_L = 200\Omega$ , $C_L = 15pF$	–40°C to +85°C			12	ns
		Refer to tON(EN) and tOFF(EN)	-40°C to +125°C			12	ns
Q <sub>C</sub>	Charge Injection	$V_S = 1V$ $R_S = 0\Omega$ , $C_L = 1nF$ Refer to Charge Injection	25°C	-1			рС
_		$R_L = 50\Omega$ , $C_L = 5pF$ f = 1MHz Refer to Off Isolation	25°C		-65		dB
O <sub>ISO</sub>	Off Isolation	$R_L = 50\Omega$ , $C_L = 5pF$ f = 10MHz Refer to Off Isolation	25°C		-45		dB
· ·	O t. II.	$R_L = 50\Omega$ , $C_L = 5pF$ f = 1MHz Refer to Crosstalk	25°C		-90		dB
X <sub>TALK</sub>	Crosstalk	$R_L = 50\Omega$ , $C_L = 5pF$ f = 10MHz Refer to Crosstalk	25°C		-80		dB
BW	Bandwidth	$R_L = 50\Omega$ , $C_L = 5pF$ Refer to Bandwidth 25°C 135			MHz		
C <sub>SOFF</sub>	Source off capacitance	f = 1MHz	25°C		7		pF
C <sub>DOFF</sub>	Drain off capacitance	f = 1MHz	25°C		32		pF
C <sub>SON</sub> C <sub>DON</sub>	On capacitance	f = 1MHz	25°C		38		pF

# 5.7 Electrical Characteristics ( $V_{DD}$ = 2.5V ±10 %), ( $V_{SS}$ = -2.5V ±10 %)

at  $T_A = 25$ °C,  $V_{DD} = +2.5$ V,  $V_{SS} = -2.5$ V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
ANALC	G SWITCH		<u> </u>				
		$V_S = V_{SS}$ to $V_{DD}$	25°C		1.8	4	Ω
$R_{ON}$	On-resistance	I <sub>SD</sub> = 10mA	–40°C to +85°C			4.5	Ω
		Refer to On-Resistance	-40°C to +125°C			4.9	Ω
	On-resistance matching between channels $ \begin{aligned} & V_S = V_{SS} \text{ to } V_{DD} \\ & I_{SD} = 10 \text{mA} \\ & \text{Refer to On-Resistance} \end{aligned} $	Ve = Vee to Vee	25°C		0.18		Ω
$\Delta R_{ON}$		I <sub>SD</sub> = 10mA	–40°C to +85°C			0.4	Ω
		Refer to On-Resistance	-40°C to +125°C			0.5	Ω
		$V_S = V_{SS}$ to $V_{DD}$	25°C		0.85		Ω
R <sub>ON</sub>	On-resistance flatness	I <sub>SD</sub> = 10mA	–40°C to +85°C			1.6	Ω
FLAT		Refer to On-Resistance	-40°C to +125°C			1.6	Ω
		$V_{DD}$ = +2.5V, $V_{SS}$ = -2.5V	25°C	-0.08	±0.005	0.08	nA
la	Source off leakage current	Switch Off $V_D = +2V / -1V$	-40°C to +85°C	-0.3		0.3	nA
I <sub>S(OFF)</sub>	Course on rearrage current	$V_S = -1V / +2V$ Refer to Off-Leakage Current	-40°C to +125°C	-0.9		0.9	nA

# 5.7 Electrical Characteristics ( $V_{DD}$ = 2.5V ±10 %), ( $V_{SS}$ = -2.5V ±10 %) (continued)

at  $T_A = 25$ °C,  $V_{DD} = +2.5$ V,  $V_{SS} = -2.5$ V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
		V <sub>DD</sub> = +2.5V, V <sub>SS</sub> = -2.5V	25°C	-0.1	±0.01	0.1	nA
l_	Drain off leakage current	Switch Off $V_D = +2V / -1V$	–40°C to +85°C	-0.75		0.75	nA
I <sub>D(OFF)</sub>	Drain on leakage current	V <sub>S</sub> = -1V / +2V Refer to Off-Leakage Current	-40°C to +125°C	-3.5		3.5	nA
		V <sub>DD</sub> = +2.5V, V <sub>SS</sub> = -2.5V	25°C	-0.1	±0.01	0.1	nA
$I_{D(ON)}$ $I_{S(ON)}$	Channel on leakage current	Switch On $V_D = V_S = +2V / -1V$	-40°C to +85°C	-0.75		0.75	nA
·3(ON)		Refer to On-Leakage Current	-40°C to +125°C	-3		3	nA
LOGIC	INPUTS (EN, A0, A1)						
V <sub>IH</sub>	Input logic high		-40°C to +125°C	1.2		2.75	V
V <sub>IL</sub>	Input logic low		40 C to +123 C	0		0.73	V
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		25°C		±0.005		μΑ
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		-40°C to +125°C			±0.05	μΑ
C <sub>IN</sub>	Logic input capacitance		25°C		1		pF
→IN	20gio input oapaoitanoo		–40°C to +125°C			2	pF
POWER	SUPPLY		_				
I <sub>DD</sub> V <sub>I</sub>	V <sub>DD</sub> supply current	Logic inputs = 0V or 2.75V	25°C		0.008		μΑ
	VDD supply current	Logic inputs = 0 v or 2.75 v	-40°C to +125°C			1	μΑ
I <sub>SS</sub>	V <sub>SS</sub> supply current	Logic inputs = 0V or 2.75V	25°C		0.008		μΑ
iss	VSS supply current	Logic inputs – 0 v or 2.75 v	-40°C to +125°C			1	μΑ
DYNAM	IC CHARACTERISTICS						
		V <sub>S</sub> = 1.5V	25°C		14		ns
t <sub>TRAN</sub>	Transition time between channels	$R_L = 200\Omega$ , $C_L = 15pF$	–40°C to +85°C			21	ns
		Refer to Transition Time	–40°C to +125°C			21	ns
		V <sub>S</sub> = 1.5V	25°C		8		ns
t <sub>open</sub> (BBM)	Break before make time	$R_L = 200\Omega$ , $C_L = 15pF$	–40°C to +85°C	1			ns
(55)		Refer to Break-Before-Make	–40°C to +125°C	1			ns
		V <sub>S</sub> = 1.5V	25°C		14		ns
t <sub>ON(EN)</sub>	Enable turn-on time	$R_L = 200\Omega$ , $C_L = 15pF$	–40°C to +85°C			21	ns
		Refer to tON(EN) and tOFF(EN)	–40°C to +125°C			22	ns
		V <sub>S</sub> = 1.5V	25°C		8		ns
t <sub>OFF(EN)</sub>	Enable turn-off time	$R_{I} = 200\Omega, C_{I} = 15pF$	–40°C to +85°C			11	ns
		Refer to tON(EN)and tOFF(EN)	-40°C to +125°C			12	ns
$Q_{C}$	Charge Injection	$V_S = -1V$ $R_S = 0\Omega$ , $C_L = 1nF$ Refer to Charge Injection	25°C		-1		рС
0	Off legistion	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 1MHz Refer to Off Isolation	25°C		-65		dB
O <sub>ISO</sub>	Off Isolation	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Off Isolation	25°C		-45		dB



# 5.7 Electrical Characteristics ( $V_{DD}$ = 2.5V ±10 %), ( $V_{SS}$ = -2.5V ±10 %) (continued)

at  $T_A = 25$ °C,  $V_{DD} = +2.5$ V,  $V_{SS} = -2.5$ V (unless otherwise noted)

	<u> </u>	1				
	PARAMETER	TEST CONDITIONS	TA	MIN TYP	MAX	UNIT
X <sub>TALK</sub>	Crosstalk	$R_L = 50\Omega$ , $C_L = 5pF$ f = 1MHz Refer to Crosstalk	25°C	-90		dB
	Crosstaik	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Crosstalk	25°C	-80		dB
BW	Bandwidth	$R_L = 50\Omega$ , $C_L = 5pF$ Refer to Bandwidth	25°C	135		MHz
C <sub>SOFF</sub>	Source off capacitance	f = 1MHz	25°C	7		pF
C <sub>DOFF</sub>	Drain off capacitance	f = 1MHz	25°C	32		pF
C <sub>SON</sub> C <sub>DON</sub>	On capacitance	f = 1MHz	25°C	38		pF

# 5.8 Electrical Characteristics ( $V_{DD} = 1.8V \pm 10 \%$ )

at  $T_{\Delta}$  = 25°C,  $V_{DD}$  = 1.8V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
ANALO	G SWITCH						
		$V_S = 0V \text{ to } V_{DD}$	25°C		40		Ω
R <sub>ON</sub>	On-resistance	$I_{SD} = 10 \text{mA}$	–40°C to +85°C			80	Ω
		Refer to On-Resistance	-40°C to +125°C			80	Ω
		$V_S = 0V \text{ to } V_{DD}$	25°C		0.4		Ω
$\Delta R_{ON}$	On-resistance matching between channels	I <sub>SD</sub> = 10mA	-40°C to +85°C			1.5	Ω
	Sharmolo .	Refer to On-Resistance	-40°C to +125°C			1.5	Ω
	Source off leakage current	V <sub>DD</sub> = 1.98V	25°C	-0.05	±0.003	0.05	nΑ
Is/OFF)		Switch Off V <sub>D</sub> = 1.62V / 1V	–40°C to +85°C	-0.1		0.1	nA
Source off leakage curre	Source on loakage carroin	V <sub>S</sub> = 1V / 1.62V Refer to Off-Leakage Current	-40°C to +125°C	-0.5		0.5	nA
		V <sub>DD</sub> = 1.98V	25°C	-0.1	±0.005	0.1	nA
I <sub>D(OFF)</sub>	Drain off leakage current	Switch Off V <sub>D</sub> = 1.62V / 1V	-40°C to +85°C	-0.5		0.5	nA
	Drain on loakage carrent	$V_S = 1V / 1.62V$ Refer to Off-Leakage Current	-40°C to +125°C	-2		2	nA
		V <sub>DD</sub> = 1.98V	25°C	-0.1	±0.005	0.1	nA
I <sub>D(ON)</sub>	Channel on leakage current	Switch On $V_D = V_S = 1.62V / 1V$	–40°C to +85°C	-0.5		0.5	nA
I <sub>S(ON)</sub>		Refer to On-Leakage Current	-40°C to +125°C	-2		2	nA
LOGIC	INPUTS (EN, A0, A1)			1			
V <sub>IH</sub>	Input logic high		4000 to 140500	1.07		5.5	V
V <sub>IL</sub>	Input logic low		-40°C to +125°C	0		0.68	V
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		25°C		±0.005		μA
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		-40°C to +125°C			±0.05	μΑ
	Logic input capacitance		25°C		1		pF
C <sub>IN</sub>	Logic iliput capacitatice		-40°C to +125°C			2	pF
POWER	RSUPPLY					'	
	V aupply aurrent	supply current Logic inputs = 0V or 5.5V	25°C		0.001		μΑ
I <sub>DD</sub>	V <sub>DD</sub> supply current		-40°C to +125°C			0.85	μΑ

# 5.8 Electrical Characteristics (V<sub>DD</sub> = 1.8V ±10 %) (continued)

at T<sub>A</sub> = 25°C, V<sub>DD</sub> = 1.8V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
DYNAM	IC CHARACTERISTICS						
		V <sub>S</sub> = 1V	25°C		28		ns
t <sub>TRAN</sub>	Transition time between channels	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C			48	ns
		Refer to Transition Time	-40°C to +125°C			48	ns
		V <sub>S</sub> = 1V	25°C		16		ns
t <sub>OPEN</sub>	Break before make time	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C	1			ns
(BBM)		Refer to Break-Before-Make	-40°C to +125°C	1			ns
		V <sub>S</sub> = 1V	25°C		28		ns
t <sub>ON(EN)</sub>	Enable turn-on time	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C			48	ns
		Refer to tON(EN) and tOFF(EN)	–40°C to +125°C			48	ns
		V <sub>S</sub> = 1V	25°C		16		ns
t <sub>OFF(EN)</sub>	Enable turn-off time	$R_{L} = 200\Omega$ , $C_{L} = 15pF$	–40°C to +85°C			27	ns
		Refer to tON(EN) and tOFF(EN)	-40°C to +125°C			27	ns
Q <sub>C</sub>	Charge Injection	$V_S = 1V$ $R_S = 0\Omega$ , $C_L = 1nF$ Refer to Charge Injection	25°C		-0.5		рC
0		R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 1MHz Refer to Off Isolation	25°C		-65		dB
O <sub>ISO</sub>	Off Isolation	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Off Isolation	25°C		<b>–45</b>		dB
v	0	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 1MHz Refer to Crosstalk	25°C		-90		dB
X <sub>TALK</sub>	Crosstalk	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Crosstalk	25°C		-80		dB
BW	Bandwidth	$R_L = 50\Omega$ , $C_L = 5pF$ Refer to Bandwidth	25°C 135			MHz	
C <sub>SOFF</sub>	Source off capacitance	f = 1MHz	25°C		7		pF
C <sub>DOFF</sub>	Drain off capacitance	f = 1MHz	25°C		32		pF
C <sub>SON</sub> C <sub>DON</sub>	On capacitance	f = 1MHz	25°C				pF

# 5.9 Electrical Characteristics ( $V_{DD} = 1.2V \pm 10 \%$ )

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
ANALO	G SWITCH						
		$V_S = 0V \text{ to } V_{DD}$	25°C		70		Ω
R <sub>ON</sub>	On-resistance	I <sub>SD</sub> = 10mA	-40°C to +85°C			105	Ω
		Refer to On-Resistance	-40°C to +125°C			105	Ω
	V <sub>0</sub> = (	$V_S = 0V \text{ to } V_{DD}$	25°C		0.4		Ω
ΔR <sub>ON</sub>	On-resistance matching between channels	I <sub>SD</sub> = 10mA Refer to On-Resistance	-40°C to +85°C			1.5	Ω
	Chamillo		-40°C to +125°C			1.5	Ω
		V <sub>DD</sub> = 1.32V	25°C	-0.05	±0.003	0.05	nA
la.a==	Source off leakage current <sup>(1)</sup>	Switch Off V <sub>D</sub> = 1V / 0.8V	-40°C to +85°C	-0.1		0.1	nA
IS(OFF)	Source on loakage danone	V <sub>S</sub> = 0.8V / 1V Refer to Off-Leakage Current	-40°C to +125°C	-0.5		0.5	nA

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# 5.9 Electrical Characteristics ( $V_{DD}$ = 1.2V ±10 %) (continued)

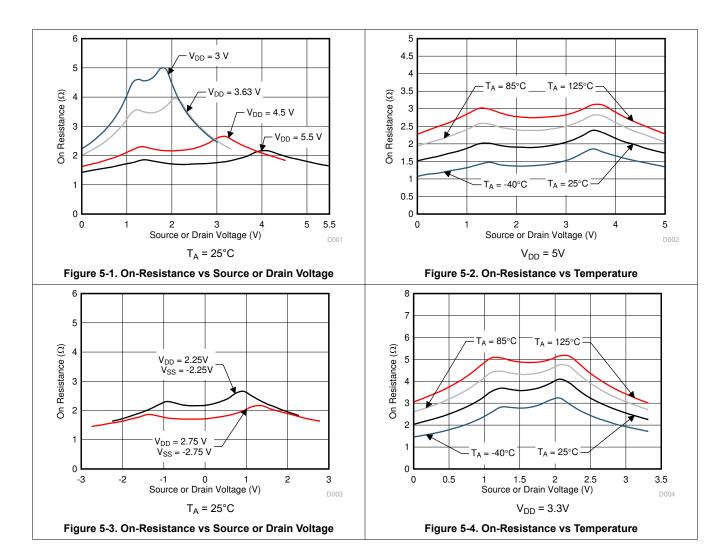
	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
		V <sub>DD</sub> = 1.32V	25°C	-0.1	±0.005	0.1	nA
l	Drain off leakage current <sup>(1)</sup>	Switch Off $V_D = 1V / 0.8V$	-40°C to +85°C	-0.5		0.5	nA
I <sub>D(OFF)</sub>	Drain on leakage current	V <sub>S</sub> = 0.8V / 1V Refer to Off-Leakage Current	-40°C to +125°C	-2		2	nA
		V <sub>DD</sub> = 1.32V	25°C	-0.1	±0.005	0.1	nA
$I_{D(ON)}$ $I_{S(ON)}$	Channel on leakage current	Switch On $V_D = V_S = 1V / 0.8V$	-40°C to +85°C	-0.5		0.5	nA
·3(ON)		Refer to On-Leakage Current	-40°C to +125°C	-2		2	nA
LOGIC I	NPUTS (EN, A0, A1)		•				
$V_{IH}$	Input logic high		-40°C to +125°C	0.96		5.5	V
$V_{IL}$	Input logic low		40 0 to 1123 0	0		0.36	V
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		25°C		±0.005		μΑ
I <sub>IH</sub> I <sub>IL</sub>	Input leakage current		-40°C to +125°C			±0.05	μΑ
<u></u>	Logic input conscitones		25°C		1		pF
C <sub>IN</sub>	Logic input capacitance		-40°C to +125°C			2	pF
POWER	SUPPLY						
	V gupply gurrent	Logic inputs = 0V or 5.5V	25°C		0.001		μΑ
I <sub>DD</sub>	V <sub>DD</sub> supply current	Logic inputs – 0V or 5.5V	-40°C to +125°C			0.7	μΑ
DYNAM	IC CHARACTERISTICS						
		V <sub>S</sub> = 1V	25°C		60		ns
t <sub>TRAN</sub>	Transition time between channels	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C			210	ns
		Refer to Transition Time	-40°C to +125°C			210	ns
		V <sub>S</sub> = 1V	25°C		28		ns
t <sub>OPEN</sub>	Break before make time	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C	1			ns
(BBM)		Refer to Break-Before-Make	-40°C to +125°C	1			ns
		V <sub>S</sub> = 1V	25°C		60		ns
t <sub>ON(EN)</sub>	Enable turn-on time	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C			190	ns
		Refer to tON(EN) and tOFF(EN)	-40°C to +125°C			190	ns
		V <sub>S</sub> = 1V	25°C		45		ns
t <sub>OFF(EN)</sub>	Enable turn-off time	$R_L = 200\Omega$ , $C_L = 15pF$	-40°C to +85°C			150	ns
		Refer to tON(EN) and tOFF(EN)	-40°C to +125°C		-	150	ns
Q <sub>C</sub>	Charge Injection	$V_S = 1V$ $R_S = 0\Omega$ , $C_L = 1nF$ Refer to Charge Injection	25°C		-0.5		рС
		R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 1MHz Refer to Off Isolation	25°C		-65		dB
O <sub>ISO</sub>	Off Isolation	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Off Isolation	25°C		-45		dB
<b>V</b>	Croostelly	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 1MHz Refer to Crosstalk	25°C		-90		dB
X <sub>TALK</sub>	Crosstalk	R <sub>L</sub> = 50Ω, C <sub>L</sub> = 5pF f = 10MHz Refer to Crosstalk	25°C		-80		dB
BW	Bandwidth	$R_L = 50\Omega$ , $C_L = 5pF$ Refer to Bandwidth	25°C		135		MHz

# 5.9 Electrical Characteristics (V<sub>DD</sub> = 1.2V ±10 %) (continued)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
C <sub>SOFF</sub>	Source off capacitance	f = 1MHz	25°C		7		pF
C <sub>DOFF</sub>	Drain off capacitance	f = 1MHz	25°C		32		pF
C <sub>SON</sub> C <sub>DON</sub>	On capacitance	f = 1MHz	25°C		38		pF

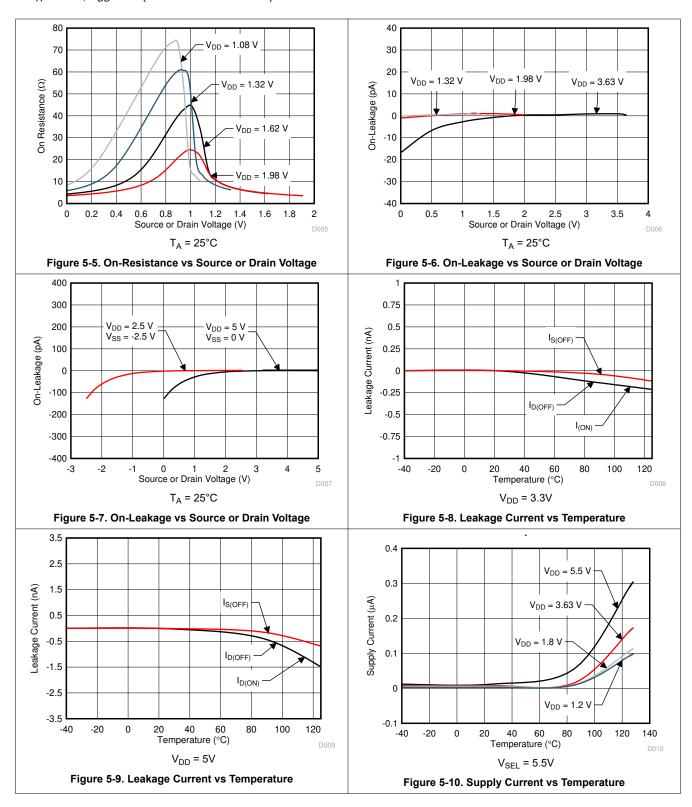
(1) When  $V_S$  is 1V,  $V_D$  is 0.8V, and vice versa.

## **Typical Characteristics**

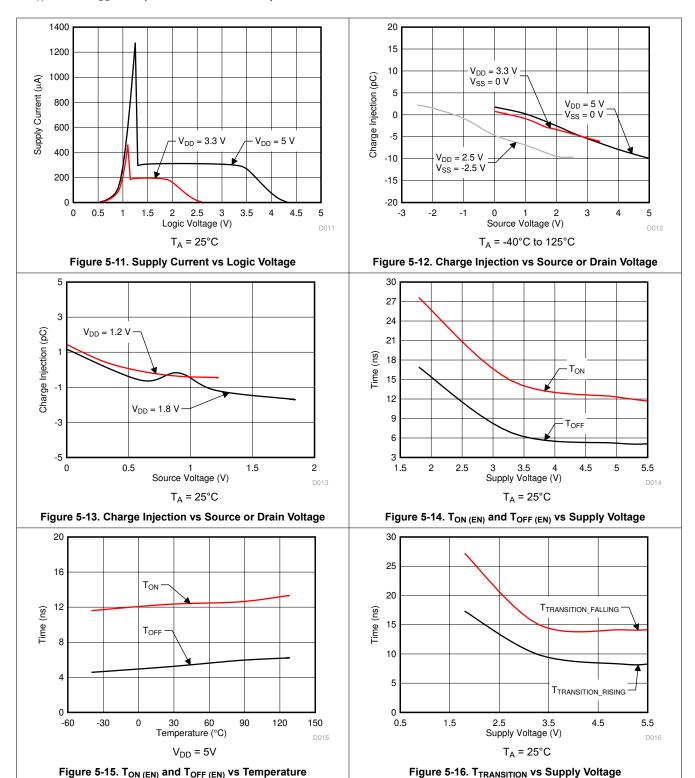




# **Typical Characteristics**



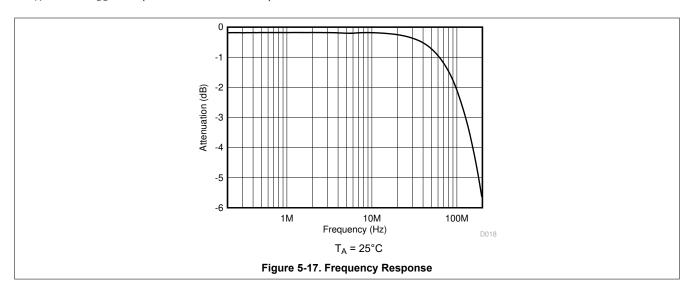
# **Typical Characteristics**





# **Typical Characteristics**

at  $T_A = 25$ °C,  $V_{DD} = 5V$  (unless otherwise noted)



## **6 Detailed Description**

### 6.1 Overview

#### 6.1.1 On-Resistance

The on-resistance of a device is the ohmic resistance between the source (Sx) and drain (D) pins of the device. The on-resistance varies with input voltage and supply voltage. The symbol  $R_{ON}$  is used to denote on-resistance. The measurement setup used to measure  $R_{ON}$  is shown in Figure 6-1. Voltage (V) and current ( $I_{SD}$ ) are measured using this setup, and  $R_{ON}$  is computed with  $R_{ON} = V / I_{SD}$ :

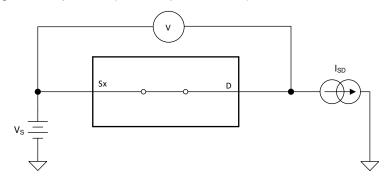


Figure 6-1. On-Resistance Measurement Setup

## 6.1.2 Off-Leakage Current

There are two types of leakage currents associated with a switch during the off state:

- 1. Source off-leakage current
- 2. Drain off-leakage current

Source leakage current is defined as the leakage current flowing into or out of the source pin when the switch is off. This current is denoted by the symbol  $I_{S(OFF)}$ .

Drain leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is off. This current is denoted by the symbol  $I_{D(OFF)}$ .

The setup used to measure both off-leakage currents is shown in Figure 6-2.

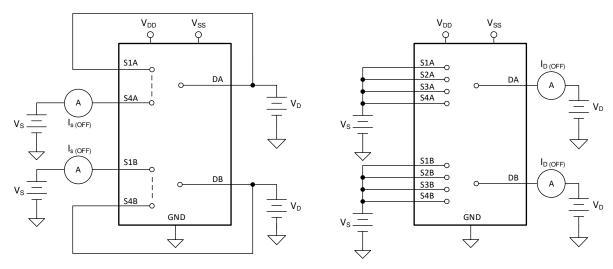


Figure 6-2. Off-Leakage Measurement Setup

#### 6.1.3 On-Leakage Current

Source on-leakage current is defined as the leakage current flowing into or out of the source pin when the switch is on. This current is denoted by the symbol  $I_{S(ON)}$ .

Drain on-leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is on. This current is denoted by the symbol  $I_{D(ON)}$ .

Either the source pin or drain pin is left floating during the measurement. Figure 6-3 shows the circuit used for measuring the on-leakage current, denoted by  $I_{S(ON)}$  or  $I_{D(ON)}$ .

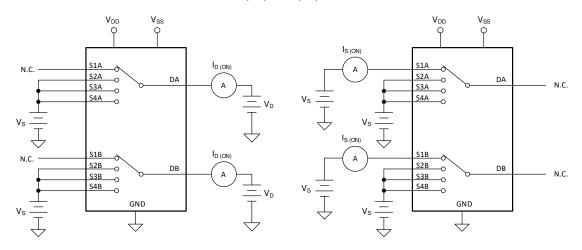


Figure 6-3. On-Leakage Measurement Setup

#### 6.1.4 Transition Time

Transition time is defined as the time taken by the output of the device to rise or fall 10% after the address signal has risen or fallen past the logic threshold. The 10% transition measurement is utilized to provide the timing of the device, system level timing can then account for the time constant added from the load resistance and load capacitance. Figure 6-4 shows the setup used to measure transition time, denoted by the symbol  $t_{TRANSITION}$ .

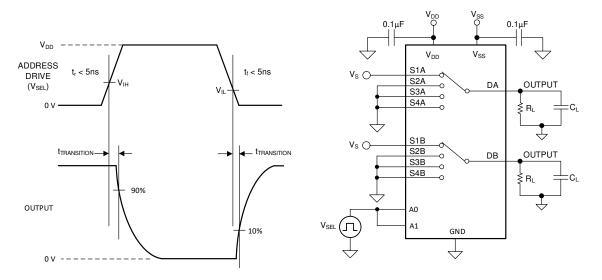


Figure 6-4. Transition-Time Measurement Setup

#### 6.1.5 Break-Before-Make

Break-before-make delay is a safety feature that prevents two inputs from connecting when the device is switching. The output first breaks from the on-state switch before making the connection with the next on-state switch. The time delay between the *break* and the *make* is known as break-before-make delay. Figure 6-5 shows the setup used to measure break-before-make delay, denoted by the symbol topen(BBM).

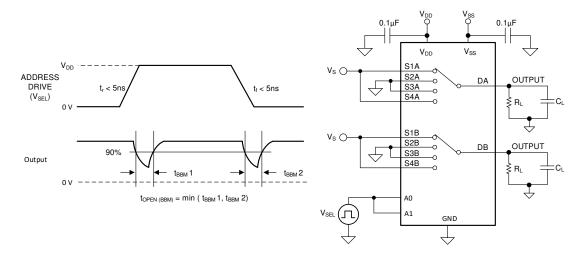


Figure 6-5. Break-Before-Make Delay Measurement Setup

## 6.1.6 t<sub>ON(EN)</sub> and t<sub>OFF(EN)</sub>

Turn-on time is defined as the time taken by the output of the device to rise to 10% after the enable has risen past the logic threshold. The 10% measurement is utilized to provide the timing of the device, system level timing can then account for the time constant added from the load resistance and load capacitance. Figure 6-6 shows the setup used to measure turn-on time, denoted by the symbol  $t_{ON(EN)}$ .

Turn-off time is defined as the time taken by the output of the device to fall to 90% after the enable has fallen past the logic threshold. The 90% measurement is utilized to provide the timing of the device, system level timing can then account for the time constant added from the load resistance and load capacitance. Figure 6-6 shows the setup used to measure turn-off time, denoted by the symbol  $t_{OFF(EN)}$ .

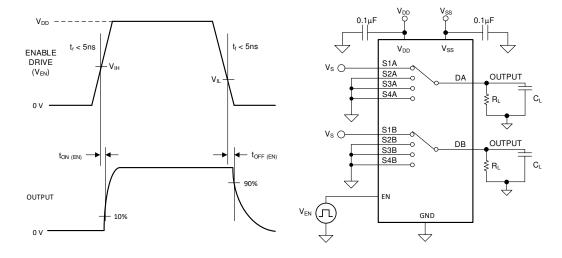


Figure 6-6. Turn-On and Turn-Off Time Measurement Setup

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### 6.1.7 Charge Injection

The TMUX1109 has a transmission-gate topology. Any mismatch in capacitance between the NMOS and PMOS transistors results in a charge injected into the drain or source during the falling or rising edge of the gate signal. The amount of charge injected into the source or drain of the device is known as charge injection, and is denoted by the symbol Q<sub>C</sub>. Figure 6-7 shows the setup used to measure charge injection from source (Sx) to drain (D).

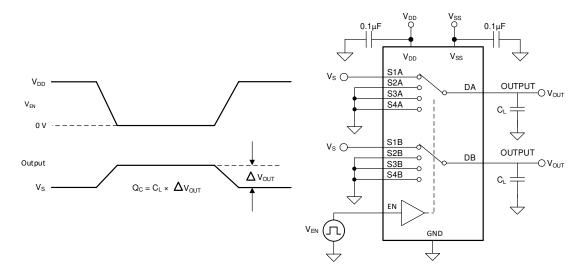


Figure 6-7. Charge-Injection Measurement Setup

### 6.1.8 Off Isolation

Off isolation is defined as the ratio of the signal at the drain pin (D) of the device when a signal is applied to the source pin (Sx) of an off-channel. Figure 6-8 shows the setup used to measure and the equation used to compute off isolation.

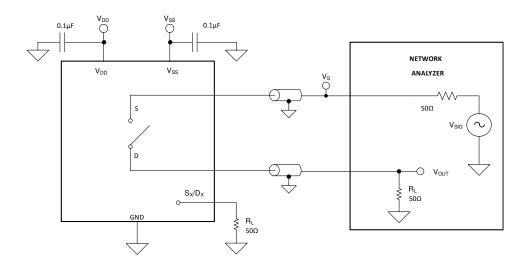


Figure 6-8. Off Isolation Measurement Setup

Off Isolation = 
$$20 \cdot Log\left(\frac{V_{OUT}}{V_{S}}\right)$$
 (1)

#### 6.1.9 Crosstalk

Crosstalk is defined as the ratio of the signal at the drain pin (D) of a different channel, when a signal is applied at the source pin (Sx) of an on-channel. Figure 6-9 shows the setup used to measure, and the equation used to compute crosstalk.

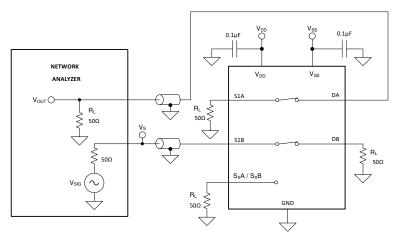


Figure 6-9. Crosstalk Measurement Setup

Channel-to-Channel Crosstalk = 
$$20 \cdot Log\left(\frac{V_{OUT}}{V_{S}}\right)$$
 (2)

#### 6.1.10 Bandwidth

Bandwidth is defined as the range of frequencies that are attenuated by less than 3 dB when the input is applied to the source pin (Sx) of an on-channel, and the output is measured at the drain pin (D) of the device. Figure 6-10 shows the setup used to measure bandwidth.

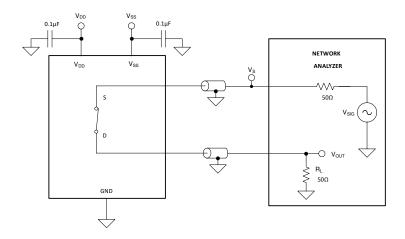


Figure 6-10. Bandwidth Measurement Setup

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### 6.2 Functional Block Diagram

The TMUX1109 is an 4:1, differential (2-channel), multiplexer. Each switch is turned on or off based on the state of the address lines and enable pin.

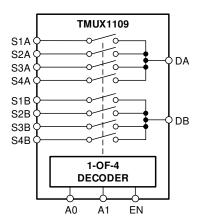


Figure 6-11. TMUX1109 Functional Block Diagram

### **6.3 Feature Description**

#### 6.3.1 Bidirectional Operation

The TMUX1109 conducts equally well from source (Sx) to drain (Dx) or from drain (Dx) to source (Sx). Each channel has very similar characteristics in both directions and supports both analog and digital signals.

### 6.3.2 Rail to Rail Operation

The valid signal path input/output voltage for TMUX1109 ranges from V<sub>SS</sub> to V<sub>DD</sub>.

#### 6.3.3 1.8V Logic Compatible Inputs

The TMUX1109 has 1.8-V logic compatible control for all logic control inputs. The logic input thresholds scale with supply but still provide 1.8-V logic control when operating at 5.5V supply voltage. 1.8-V logic level inputs allows the TMUX1109 to interface with processors that have lower logic I/O rails and eliminates the need for an external translator, which saves both space and BOM cost. For more information on 1.8V logic implementations refer to Simplifying Design with 1.8V logic Muxes and Switches

#### 6.3.4 Fail-Safe Logic

The TMUX1109 supports Fail-Safe Logic on the control input pins (EN, A0, A1) allowing for operation up to 5.5V above  $V_{SS}$ , regardless of the state of the supply pin. This feature allows voltages on the control pins to be applied before the supply pin, protecting the device from potential damage. Fail-Safe Logic minimizes system complexity by removing the need for power supply sequencing on the logic control pins. For example, the Fail-Safe Logic feature allows the select pins of the TMUX1109 to be ramped to 5.5V while  $V_{DD} = 0V$ . Additionally, the feature enables operation of the TMUX1109 with  $V_{DD} = 1.2V$  while allowing the select pins to interface with a logic level of another device up to 5.5V.

# 6.3.5 Ultra-Low Leakage Current

The TMUX1109 provides extremely low on-leakage and off-leakage currents. The TMUX1109 is capable of switching signals from high source-impedance inputs into a high input-impedance op amp with minimal offset error because of the ultralow leakage currents. Figure 6-12 shows typical leakage currents of the TMUX1109 versus temperature.

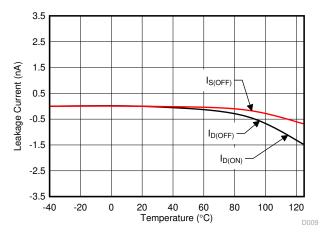


Figure 6-12. Leakage Current vs Temperature

#### 6.3.6 Ultra-Low Charge Injection

The TMUX1109 has a transmission gate topology, as shown in Figure 6-13. Any mismatch in the stray capacitance associated with the NMOS and PMOS causes an output level change whenever the switch is opened or closed.

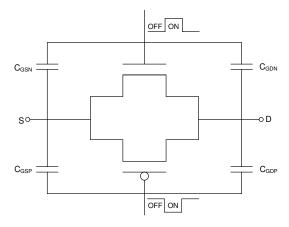


Figure 6-13. Transmission Gate Topology

The TMUX1109 has special charge-injection cancellation circuitry that reduces the source-to-drain charge injection to as low as 1pC at  $V_S$  = 1V as shown in Figure 6-14.

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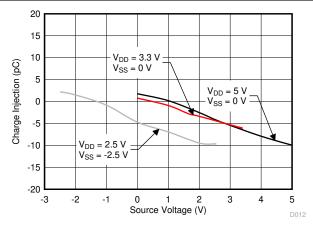


Figure 6-14. Charge Injection vs Source Voltage

### **6.4 Device Functional Modes**

When the EN pin of the TMUX1109 is pulled high, one of the switches is closed based on the state of the address lines. When the EN pin is pulled low, all the switches are in an open state regardless of the state of the address lines.

#### 6.4.1 Truth Tables

Table 6-1, TMUX1109 Truth Table

EN	A1	A0	Selected Input Connected To Drain (DA, DB) Pins		
0	X <sup>(1)</sup>	X <sup>(1)</sup>	All channels are off		
1	0	0	S1A and S1B		
1	0	1	S2A and S2B		
1	1	0	S3A and S3B		
1	1	1	S4A and S4B		

(1) X denotes do not care.



# 7 Application and Implementation

#### **Note**

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

## 7.1 Application Information

The TMUX11xx family offers ulta-low input/output leakage currents and low charge injection. These devices operate up to 5.5V, and offer true rail-to-rail input and output. The TMUX1109 has a low on-capacitance which allows faster settling time when multiplexing inputs in the time domain. These features make the TMUX11xx devices a family of precision, robust, high-performance analog multiplexer for low-voltage applications.

### 7.2 Typical Application

Figure 7-1 shows a 16-bit, simultaneous-sampling data-acquisition system. This example is typical in industrial applications that require sampling simultaneous signals such as optical modules, analog input modules, and motor drive circuits for position feedback. The circuit uses eight fully differential amplifiers (FDAs), a 16-bit, 3-MSPS successive-approximation-resistor (SAR) analog-to-digital converter (ADC), along with a two differential precision multiplexers. Refer to *True Differential*, 4 x 2 MUX, Analog Front End, Simultaneous-Sampling ADC Circuit for more information.

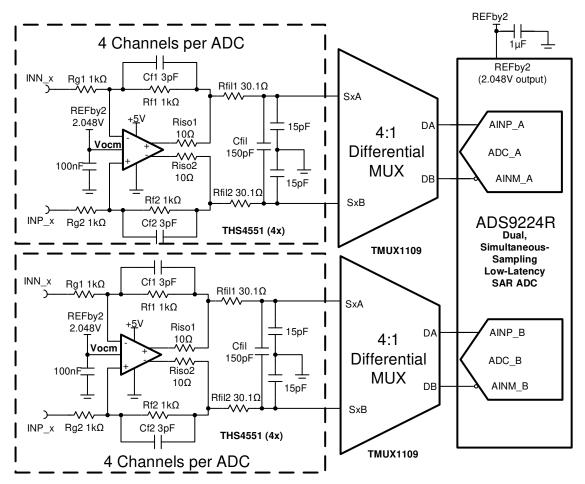


Figure 7-1. Simultaneous-Sampling ADC Circuit

#### 7.2.1 Design Requirements

For this design example, use the parameters listed in Table 7-1.

Table 7-1. Design Parameters

PARAMETERS	VALUES
Supply (V <sub>DD</sub> )	5V
Vref	4.096V
Vocm	2.048V
Max Differential Voltage	3.636V
Control logic thresholds	1.8V compatible

### 7.2.2 Detailed Design Procedure

The TMUX1109 can operate without any external components except for the supply decoupling capacitors. If the device desired power-up state is disabled, then the enable pin should have a weak pull-down resistor and be controlled by the MCU through the GPIO. All inputs being muxed to the ADC must fall within the recommend operating conditions of the TMUX1109 including signal range and continuous current. System level design and component selection are made according to *True Differential*, 4 x 2 MUX, Analog Front End, Simultaneous-Sampling ADC Circuit.

- 1. The ADS9224R was selected because of the dual simultaneous sampling and high throughput (3-MSPS).
- 2. The TMUX1109 4:1 (2x) multiplexer was selected to support 4 differential inputs for each ADC.
- 3. Find ADC full-scale range, resolution and common-mode range specifications.
- 4. Determine the linear range of the FDA (THS4551) based on common-mode and output swing specification.
- 5. Select COG capacitors for all filter capacitors at the ADC input to minimize distortion.
- 6. Select the FDA gain resistors RF1,2 , RG1,2. Use 0.1% 20ppm/°C film resistors or better for good accuracy, low gain drift and to minimize distortion.
- 7. Introduction to SAR ADC Front-End Component Selection covers the methods for selecting the charge bucket circuit Rfil1, Rfil1 and Cfil. These component values are dependent on the amplifier bandwidth, data converter sampling rare, and data converter design. The values shown here will give good settling and AC performance for the amplifier and data converter in this example. If the design is modified, a different RC filter must be selected.
- 8. The THS4551 is commonly used in high-speed precision fully differential SAR applications as it has sufficient bandwidth to settle to charge kickback transients from the ADC input sampling, and multiplexer charge injection and provides the common-mode level shifting to the voltage range of the SAR ADC.
- 9. The TMUX1109 is used in high-speed precision fully differential SAR applications as it has sufficient bandwidth, low charge injection, and low on-resistance and capacitance. Low capacitance supports fast switching between channels and allows the system to settle within required precision in the specified timing.

## 7.2.3 Application Curve

Charge injection impacts system performance and settling characteristics of the charge bucket circuit. A multiplexer with low charge injection and a flat response across input voltage allows the system to settle to the required precision during the ADC acquisition period. Figure 7-2 shows the flat charge injection of the TMUX1109 at multiple supply voltages.

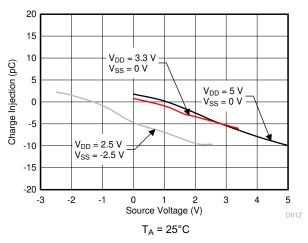


Figure 7-2. Charge Injection vs Source Voltage

## 7.3 Power Supply Recommendations

The TMUX1109 operates across a wide supply range of 1.08V to 5.5V, or ±2.5V. Do not exceed the absolute maximum ratings because stresses beyond the listed ratings can cause permanent damage to the devices.

Power-supply bypassing improves noise margin and prevents switching noise propagation from the  $V_{DD}$  and  $V_{SS}$  supplies to other components. Good power-supply decoupling is important to achieve optimum performance. For improved supply noise immunity, use a supply decoupling capacitor ranging from  $0.1\mu F$  to  $10\mu F$  from  $V_{DD}$  and  $V_{SS}$  to ground. Place the bypass capacitors as close to the power supply pins of the device as possible using low-impedance connections. TI recommends using multi-layer ceramic chip capacitors (MLCCs) that offer low equivalent series resistance (ESR) and inductance (ESL) characteristics for power-supply decoupling purposes. For very sensitive systems, or for systems in harsh noise environments, avoiding the use of vias for connecting the capacitors to the device pins may offer superior noise immunity. The use of multiple vias in parallel lowers the overall inductance and is beneficial for connections to ground planes.

#### 7.4 Layout

#### 7.4.1 Layout Guidelines

When a PCB trace turns a corner at a 90° angle, a reflection can occur. A reflection occurs primarily because of the change of width of the trace. At the apex of the turn, the trace width increases to 1.414 times the width. This increase upsets the transmission-line characteristics, especially the distributed capacitance and self–inductance of the trace which results in the reflection. Not all PCB traces can be straight and therefore some traces must turn corners. Figure 7-3 shows progressively better techniques of rounding corners. Only the last example (BEST) maintains constant trace width and minimizes reflections.

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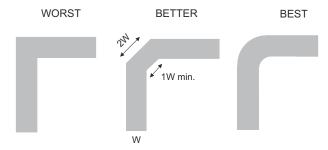


Figure 7-3. Trace Example

Route high-speed signals using a minimum of vias and corners which reduces signal reflections and impedance changes. When a via must be used, increase the clearance size around it to minimize its capacitance. Each via introduces discontinuities in the signal's transmission line and increases the chance of picking up interference from the other layers of the board. Be careful when designing test points, through-hole pins are not recommended at high frequencies.

Figure 7-4 shows an example of a PCB layout with the TMUX1109. Some key considerations are:

- Decouple the V<sub>DD</sub> pin with a 0.1µF capacitor, and place the capacitor as close to the pin as possible. Ensure
  that the capacitor voltage rating is sufficient for the V<sub>DD</sub> supply.
- Keep the input lines as short as possible.
- Use a solid ground plane to reduce electromagnetic interference (EMI) noise pickup.
- Do not run sensitive analog traces in parallel with digital traces. Avoid crossing digital and analog traces if possible, and only make perpendicular crossings when necessary.

### 7.4.2 Layout Example

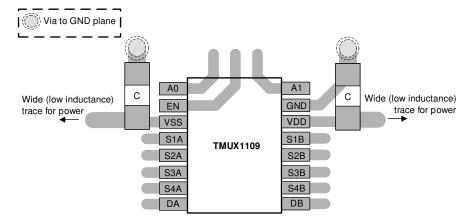


Figure 7-4. TMUX1109 Layout Example

# 8 Device and Documentation Support

## 8.1 Documentation Support

#### 8.1.1 Related Documentation

For releated documentation, see the following:

- Texas Instruments, True Differential, 4 x 2 MUX, Analog Front End, Simultaneous-Sampling ADC Circuit.
- Texas Instruments, Improve Stability Issues with Low CON Multiplexers.
- Texas Instruments, Simplifying Design with 1.8V logic Muxes and Switches.
- Texas Instruments, Eliminate Power Sequencing with Powered-off Protection Signal Switches.
- Texas Instruments, System-Level Protection for High-Voltage Analog Multiplexers.
- Texas Instruments, QFN/SON PCB Attachment.
- Texas Instruments, Quad Flatpack No-Lead Logic Packages.

## 8.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on Notifications to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 8.3 Support Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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## 8.4 Trademarks

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### 8.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 8.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

# 9 Revision History

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NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

### Changes from Revision \* (December 2018) to Revision A (February 2024) **Page** Updated Is or Id (Continuous Current) values......4

# 10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

www.ti.com 7-Oct-2025

#### PACKAGING INFORMATION

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/	MSL rating/	Op temp (°C)	Part marking
	(1)	(2)			(3)	Ball material	Peak reflow		(6)
						(4)	(5)		
TMUX1109PWR	Active	Production	TSSOP (PW)   16	2000   LARGE T&R	Yes	NIPDAU   SN	Level-1-260C-UNLIM	-40 to 125	TM1109
TMUX1109PWR.A	Active	Production	TSSOP (PW)   16	2000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TM1109
TMUX1109RSVR	Active	Production	UQFN (RSV)   16	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1D1
TMUX1109RSVR.A	Active	Production	UQFN (RSV)   16	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1D1
TMUX1109RSVRG4.A	Active	Production	UQFN (RSV)   16	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1D1

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

# **PACKAGE MATERIALS INFORMATION**

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## TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TMUX1109PWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TMUX1109RSVR	UQFN	RSV	16	3000	178.0	13.5	2.1	2.9	0.75	4.0	12.0	Q1

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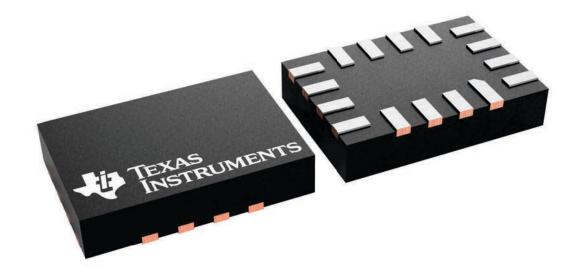
## \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TMUX1109PWR	TSSOP	PW	16	2000	356.0	356.0	35.0
TMUX1109RSVR	UQFN	RSV	16	3000	189.0	185.0	36.0

1.8 x 2.6, 0.4 mm pitch

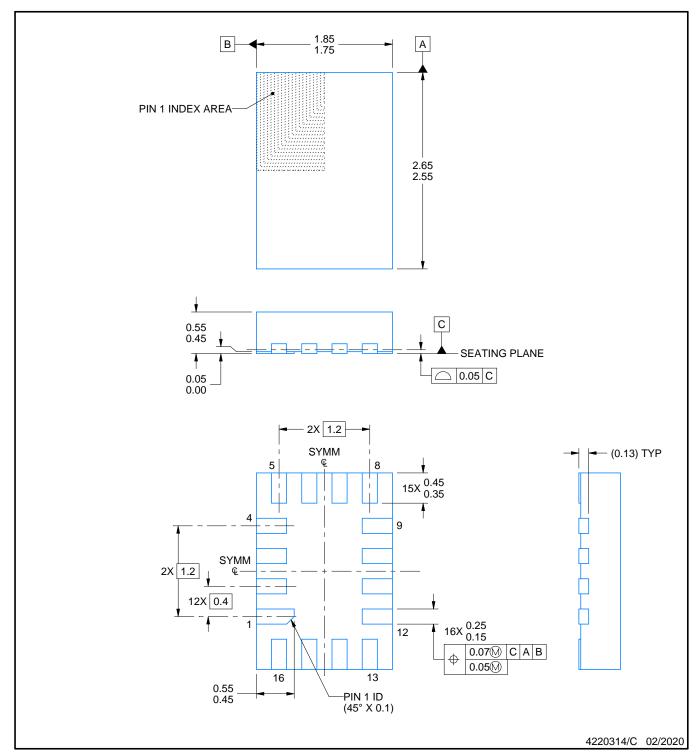
ULTRA THIN QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





ULTRA THIN QUAD FLATPACK - NO LEAD

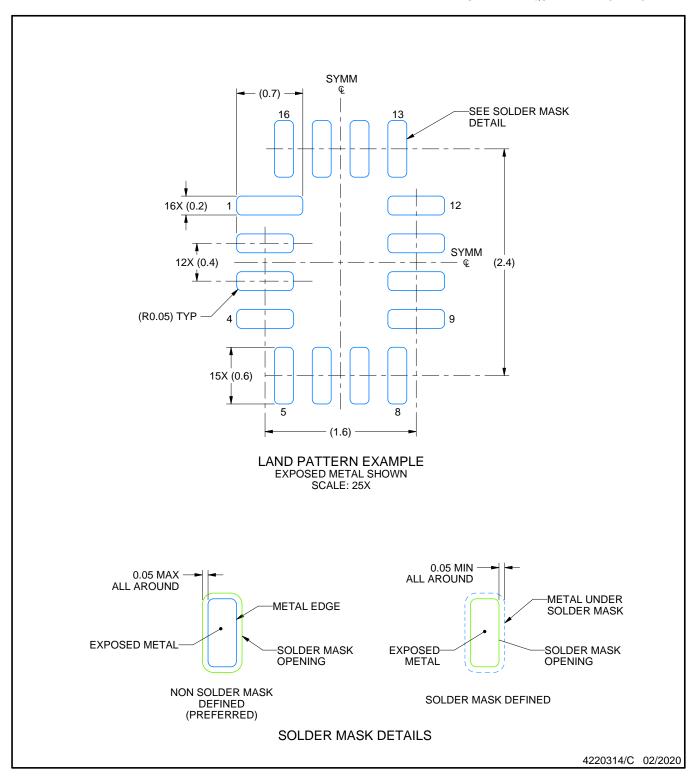


#### NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
   This drawing is subject to change without notice.



ULTRA THIN QUAD FLATPACK - NO LEAD

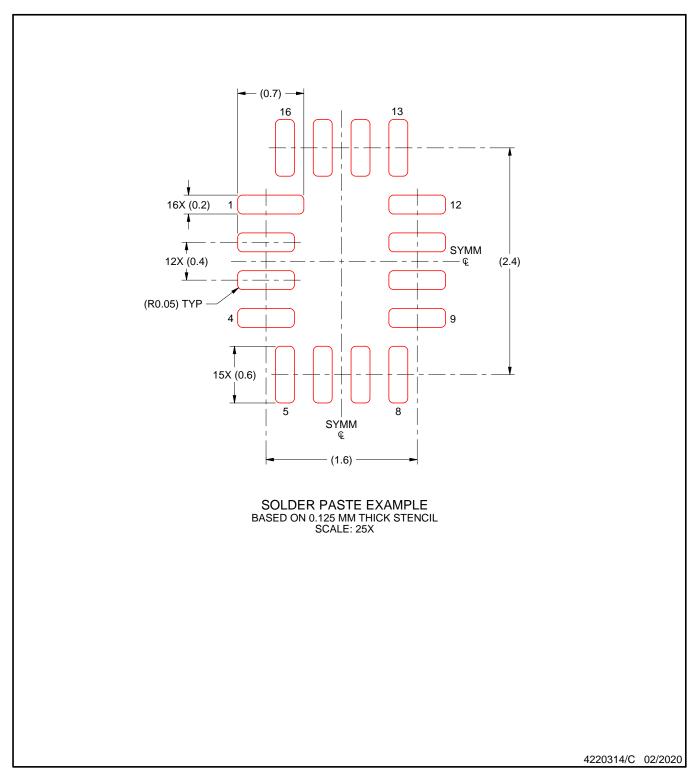


NOTES: (continued)

3. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).



ULTRA THIN QUAD FLATPACK - NO LEAD



NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.





SMALL OUTLINE PACKAGE



### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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