







SN65HVD10, SN65HVD11, SN65HVD12 SN75HVD10, SN75HVD11, SN75HVD12

SLLS505P - FEBRUARY 2002 - REVISED FEBRUARY 2022

SNx5HVD1x 3.3-V RS-485 Transceivers

1 Features

- Operates with a 3.3-V supply
- Bus-pin ESD protection exceeds 16-kV HBM
- 1/8 Unit-load option available (up to 256 nodes on
- Optional driver output transition times for signaling rates ¹of 1 Mbps, 10 Mbps, and 32 Mbps
- Meets or exceeds the requirements of ANSI TIA/ EIA-485-A
- Bus-pin short-circuit protection from -7 V to
- Low-current standby mode: 1 µA, typical
- Open-circuit, idle-bus, and shorted-bus fail-safe
- Thermal shutdown protection
- Glitch-free power-up and power-down protection for hot-plugging applications
- SN75176 footprint

2 Applications

- Digital motor control
- **Utility meters**
- Chassis-to-chassis interconnects
- Electronic security stations
- Industrial process control
- **Building automation**
- Point-of-sale (POS) terminals and networks

3 Description

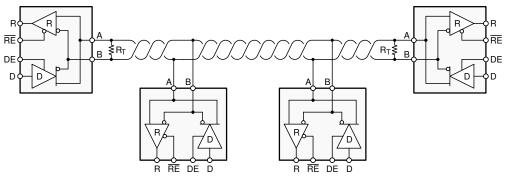
The SN65HVD10, SN75HVD10, SN65HVD11. SN75HVD11, SN65HVD12, and SN75HVD12 bus transceivers all combine a 3-state differential line driver, as well as a differential input line receiver that operates with a single 3.3-V power supply. They are designed for balanced transmission lines and meet or exceed ANSI standard TIA/EIA-485-A and ISO 8482:1993. These differential bus transceivers are monolithic integrated circuits, designed for bidirectional data communication on multipoint bustransmission lines. The drivers and receivers have active-high and active-low enables, that can be externally connected together to function as direction control. Very low device standby supply current, can be achieved by disabling the driver and the receiver.

The driver differential outputs and receiver differential inputs connect internally to form a differential input/ output (I/O) bus port, that is designed to offer minimum loading to the bus whenever the driver is disabled or V_{CC} = 0. These parts feature wide positive and negative common-mode voltage ranges, making them suitable for party-line applications.

Device Information

PART NUMBER	PACKAGE ⁽¹⁾	BODY SIZE (NOM)
SN65HVD10		
SN65HVD11	SOIC (8)	4.90 mm × 3.91 mm
SN65HVD12		
SN75HVD10		
SN75HVD11	PDIP (8)	9.81 mm × 6.35 mm
SN75HVD12		

For all available packages, see the orderable addendum at the end of the data sheet.



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Typical Application Diagram

¹ The signaling rate of a line is the number of voltage transitions that are made per second expressed in the units bps (bits per second).



Table of Contents

1 Features	1	9.2 Functional Block Diagram	17
2 Applications	1	9.3 Feature Description	17
3 Description	1	9.4 Device Functional Modes	17
4 Revision History	<mark>2</mark>	10 Application and Implementation	19
5 Device Comparison Table	3	10.1 Application Information	19
6 Pin Configuration and Functions		10.2 Typical Application	
7 Specifications	4	11 Power Supply Recommendations	<mark>23</mark>
7.1 Absolute Maximum Ratings	4	12 Layout	
7.2 ESD Ratings		12.1 Layout Guidelines	
7.3 Recommended Operating Conditions	4	12.2 Layout Example	
7.4 Thermal Information		12.3 Thermal Considerations	
7.5 Driver Electrical Characteristics	<mark>5</mark>	13 Device and Documentation Support	
7.6 Receiver Electrical Characteristics		13.1 Device Support	26
7.7 Power Dissipation Characteristics		13.2 Related Links	
7.8 Driver Switching Characteristics		13.3 Receiving Notification of Documentation	
7.9 Receiver Switching Characteristics	8	13.4 Support Resources	26
7.10 Dissipation Ratings		13.5 Trademarks	<mark>26</mark>
7.11 Typical Characteristics	9	13.6 Electrostatic Discharge Caution	<mark>26</mark>
8 Parameter Measurement Information	11	13.7 Glossary	<mark>26</mark>
9 Detailed Description	17	14 Mechanical, Packaging, and Orderable	
9.1 Overview	17	Information	26
4 Revision History Changes from Revision O (February 2017) to Changed the <i>Thermal Information</i> table	Revisio	on P (February 2022)	Page5
Changes from Revision O (February 2017) to Changed the <i>Thermal Information</i> table			5
Changes from Revision O (February 2017) to	ision O ((February 2017)	5 Page
Changes from Revision O (February 2017) to Changed the <i>Thermal Information</i> table Changes from Revision N (July 2015) to Revi Added MIN value of -55°C to the Storage tell	ision O (February 2017) e in Absolute Maximum Ratings	Page4
Changes from Revision O (February 2017) to Changed the <i>Thermal Information</i> table Changes from Revision N (July 2015) to Revi Added MIN value of -55°C to the Storage ter Changes from Revision M (July 2013) to Rev	ision O (mperatur	February 2017) re in Absolute Maximum Ratings	Page
Changes from Revision O (February 2017) to Changed the Thermal Information table Changes from Revision N (July 2015) to Revi Added MIN value of -55°C to the Storage ter Changes from Revision M (July 2013) to Rev Added Pin Configuration and Functions section Functional Modes, Application and Implement section, Device and Documentation Support	ision O (mperatur rision N (ion, ESD ntation series section,	February 2017) e in Absolute Maximum Ratings	Page4 Page evice on, Layout formation
Changes from Revision O (February 2017) to Changed the Thermal Information table Changes from Revision N (July 2015) to Revi Added MIN value of -55°C to the Storage ter Changes from Revision M (July 2013) to Rev Added Pin Configuration and Functions section Functional Modes, Application and Implement section, Device and Documentation Support	ision O (mperatur rision N (ion, ESD ntation se	February 2017) Te in Absolute Maximum Ratings	Page4 Page evice on, Layout formation
 Changes from Revision O (February 2017) to Changed the Thermal Information table Changes from Revision N (July 2015) to Revious Added MIN value of -55°C to the Storage terman Revision M (July 2013) to Revious Added Pin Configuration and Functions section. Device and Documentation Support section. Changes from Revision L (July 2013) to Revious Changes from Revision Change	rision O (mperatur rision N (ion, ESD ntation series ection,	February 2017) Te in Absolute Maximum Ratings	Page4 Page evice on, Layout formation1 Page
 Changes from Revision O (February 2017) to Changed the Thermal Information table Changes from Revision N (July 2015) to Revious Added MIN value of -55°C to the Storage terman Added Pin Configuration and Functions section. Functional Modes, Application and Implement section, Device and Documentation Support section Changes from Revision L (July 2013) to Revious Changed the V_{IT+} TYP value From: -0.65 V Changes from Revision K (September 2011) 	ision O (mperatur rision N (ion, ESD ntation ser section, ision M (To: -0.06	February 2017) Te in Absolute Maximum Ratings	Page4 Page evice on, Layout formation1 Page6 Page
 Changes from Revision O (February 2017) to Changed the Thermal Information table Changes from Revision N (July 2015) to Revious Added MIN value of -55°C to the Storage terman Added Pin Configuration and Functions section. Functional Modes, Application and Implement section, Device and Documentation Support section Changes from Revision L (July 2013) to Revious Changed the V_{IT+} TYP value From: -0.65 V Changes from Revision K (September 2011) to Added TYP = -0.65 V to V_{IT+} 	rision O (mperatur rision N (ion, ESD ntation ser section,	(February 2017) The in Absolute Maximum Ratings	Page
 Changes from Revision O (February 2017) to Changed the Thermal Information table Changes from Revision N (July 2015) to Revious Added MIN value of -55°C to the Storage terman Added Pin Configuration and Functions section. Functional Modes, Application and Implement section, Device and Documentation Support section Changes from Revision L (July 2013) to Revious Changed the V_{IT+} TYP value From: -0.65 V Changes from Revision K (September 2011) to Added TYP = -0.65 V to V_{IT+} 	rision O (mperatur rision N (ion, ESD ntation ser section,	February 2017) Te in Absolute Maximum Ratings	Page
Changes from Revision O (February 2017) to Changed the Thermal Information table Changes from Revision N (July 2015) to Revi Added MIN value of -55°C to the Storage ter Changes from Revision M (July 2013) to Rev Added Pin Configuration and Functions section Functional Modes, Application and Implement section, Device and Documentation Support section Changes from Revision L (July 2013) to Revi Changed the V _{IT+} TYP value From: -0.65 V Changes from Revision K (September 2011) Added TYP = -0.65 V to V _{IT+} Added TYP = -0.1 V to V _{IT-} Changes from Revision J (February 2009) to	ision O (mperatur rision N (ion, ESD ntation set section, ision M (To: -0.06	February 2017) Te in Absolute Maximum Ratings	Page



5 Device Comparison Table

PART NUMBER		SIGNALING RATE	UNIT LOADS	TA	SOIC MARKING
SOIC ⁽¹⁾	PDIP	SIGNALING RATE	ISHALING NATE		SOIC WARRING
SN65HVD10D	SN65HVD10P	32 Mbps	1/2		VP10
SN65HVD11D	SN65HVD11P	10 Mbps	1/8	–40°C to 85°C	VP11
SN65HVD12D	SN65HVD12P	1 Mbps	1/8		VP12
SN75HVD10D	SN75HVD10P	32 Mbps	1/2		VN10
SN75HVD11D	SN75HVD11P	10 Mbps	1/8	−0°C to 70°C	VN11
SN75HVD12D	SN75HVD12P	1 Mbps	1/8		VN12
SN65HVD10QD	SN65HVD10QP	32 Mbps	1/2	-40°C to 125°C	VP10Q
SN65HVD11QD	SN65HVD11QP	10 Mbps	1/8	-40 C to 125 C	VP11Q

⁽¹⁾ The D package is available as a tape and reel. Add an R suffix to the part number (that is, SN75HVD11DR) for this option.

6 Pin Configuration and Functions

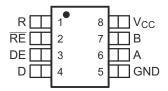


Figure 6-1. D, JD, or HKJ Package 8-Pin SOIC or PDIP (Top View)

Table 6-1. Pin Functions

PIN		TYPE	DESCRIPTION	
NAME	NO.	IIFE	DESCRIPTION	
A	6	Bus input/output	Driver output or receiver input (complementary to B)	
В	7	Bus input/output	Driver output or receiver input (complementary to A)	
D	4	Digital input	Driver data input	
DE	3	Digital input	Active-high driver enable	
GND	5	Reference potential	Local device ground	
R	1	Digital output	Receive data output	
RE	2	Digital input	Active-low receiver enable	
V _{CC}	8	Supply	3-V to 3.6-V supply	



7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range unless otherwise noted (1) (2)

		MIN	MAX	UNIT
V _{CC}	Supply voltage	-0.3	6	V
	Voltage at A or B	-9	14	V
	Input voltage at D, DE, R, or RE	-0.5	V _{CC} + 0.5	V
	Voltage input, transient pulse, A and B, through 100 Ω , see Figure 8-12	-50	50	V
Io	Receiver output current	-11	11	mA
	Continuous total power dissipation	See S	ection 7.10	
TJ	Junction temperature		170	°C
T _{stg}	Storage temperature	– 55	145	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Section 7.3 is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

				VALUE	UNIT
	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	Pins 5, 6, and 7	±16000		
	Electrostatic	Human body model (HBNI), per ANSI/ESDA/JEDEC 35-001	All pins	±4000	
V _(ESD)	discharge	Charged device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	All pins	±1000	V
		Electrical fast transient/burst ⁽³⁾	Pins 5, 6, and 7	±4000	

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

over operating free-air temperature range unless otherwise noted

			MIN	NOM	MAX	UNIT
V _{CC}	Supply voltage		3		3.6	
V _I or V _{IC}	Voltage at any bus terminal (sepa	rately or common mode)	-7 ⁽¹⁾		12	
V _{IH}	High-level input voltage	D, DE, RE	2		V _{CC}	V
V _{IL}	Low-level input voltage	D, DE, RE	0		0.8	
V _{ID}	Differential input voltage	See Figure 8-8	-12		12	
	High lavel autout august	Driver	-60			Л
I _{OH}	High-level output current	Receiver	-8			mA
	Low lovel output ourrent	Driver			60	m Λ
l _{OL}	Low-level output current	Receiver			8	mA
R _L	Differential load resistance	1	54	60		Ω
C _L	Differential load capacitance			50		pF
		HVD10			32	
	Signaling rate	HVD11			10	Mbps
		HVD12			1	
T _J ⁽²⁾	Junction temperature	1			145	°C

⁽¹⁾ The algebraic convention, in which the least positive (most negative) limit is designated as minimum is used in this data sheet.

⁽²⁾ All voltage values, except differential I/O bus voltages, are with respect to network ground terminal.

⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

⁽³⁾ Tested in accordance with IEC 61000-4-4.

⁽²⁾ See thermal characteristics table for information regarding this specification.



7.4 Thermal Information

		SNx5H	VD1xx	
	THERMAL METRIC ⁽¹⁾	D (SOIC)	P (PDIP)	UNIT
		8 Pins	8 Pins	
R _{θJA}	Junction-to-ambient thermal resistance	116.7	84.3	°C/W
R _θ JC(top)	Junction-to-case (top) thermal resistance	56.3	65.4	°C/W
R _{θJB}	Junction-to-board thermal resistance	63.4	62.1	°C/W
Ψлт	Junction-to-top characterization parameter	8.8	31.3	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	62.6	60.4	°C/W

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

7.5 Driver Electrical Characteristics

Over recommended operating conditions unless otherwise noted

	PARAMETER		TEST	CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
V _{IK}	Input clamp voltage		I _I = -18 mA		-1.5			V
			I _O = 0		2		V_{CC}	
V _{OD}	Differential output voltage ⁽²⁾		$R_L = 54 \Omega$, See Figu	R_L = 54 Ω, See Figure 8-1				V
			$V_{\text{test}} = -7 \text{ V to } 12 \text{ V},$	See Figure 8-2	1.5			
Δ V _{OD}	Change in magnitude of differe voltage	ntial output	See Figure 8-1 and	Figure 8-2	-0.2		0.2	V
V _{OC(PP)}	Peak-to-peak common-mode c voltage	utput				400		mV
V _{OC(SS)}	Steady-state common-mode or	utput voltage	See Figure 8-3		1.4		2.5	V
$\Delta V_{OC(SS)}$	Change in steady-state commo output voltage	on-mode			-0.05		0.05	V
I _{OZ}	High-impedance output current		See receiver input of	currents				
	Input current	D			-100		0	
l _l	input current	DE			0		100	μA
Ios	Short-circuit output current	•	-7 V ≤ V _O ≤ 12 V		-250		250	mA
C _(OD)	Differential output capacitance		$V_{OD} = 0.4 \sin(4E6\pi)$	t) + 0.5 V, DE at 0 V		16		pF
			RE at V _{CC} , D and DE at V _{CC,} No load	Receiver disabled and driver enabled		9	15.5	mA
Icc	Supply current		RE at V _{CC} , D at V _{CC} , DE at 0 V, No load	Receiver disabled and driver disabled (standby)		1	5	μА
			RE at 0 V, D and DE at V _{CC} , No load	Receiver enabled and driver enabled		9	15.5	mA

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply.

⁽²⁾ For $T_A > 85^{\circ}C$, V_{CC} is ±5%.



7.6 Receiver Electrical Characteristics

Over recommended operating conditions unless otherwise noted

	PARAMETER	TE	ST CONDIT	TONS	MIN	TYP ⁽¹⁾	MAX	UNIT
V _{IT+}	Positive-going input threshold voltage	I _O = -8 mA				-0.065	-0.01	
V _{IT} _	Negative-going input threshold voltage	I _O = 8 mA	I _O = 8 mA		-0.2	-0.1		V
V _{hys}	Hysteresis voltage (V _{IT+} – V _{IT-})					35		mV
V _{IK}	Enable-input clamp voltage	I _I = -18 mA			-1.5			V
V _{OH}	High-level output voltage	V _{ID} = 200 mV, I _{OH} :	= –8 mA, se	e Figure 8-8	2.4			V
V _{OL}	Low-level output voltage	V_{ID} = -200 mV, I_{OL}	= 8 mA, se	e Figure 8-8			0.4	V
I _{OZ}	High-impedance-state output current	$V_O = 0$ or V_{CC} , \overline{RE}	at V _{CC}		-1		1	μA
		V _A or V _B = 12 V				0.05	0.11	
		V_A or $V_B = 12 \text{ V}, V_C$	_{CC} = 0 V	HVD11, HVD12,		0.06	0.13	m Λ
	Bus input current	V_A or $V_B = -7 V$		Other inputs at 0 V	-0.1	-0.05		mA
		V_A or $V_B = -7$ V, $V_{CC} = 0$ V			-0.05	-0.04		
1		$V_A \text{ or } V_B = 12 \text{ V}$ $V_A \text{ or } V_B = 12 \text{ V}, V_{CC} = 0 \text{ V}$				0.2	0.5	
				HVD10,		0.25	0.5	mA
		V_A or $V_B = -7 V$		Other inputs at 0 V	-0.4	-0.2		mA
		V_A or $V_B = -7 \text{ V}, V_C$	_{CC} = 0 V		-0.4	-0.15		
I _{IH}	High-level input current, RE	V _{IH} = 2 V			-30		0	μA
I _{IL}	Low-level input current, RE	V _{IL} = 0.8 V			-30		0	μA
C _{ID}	Differential input capacitance	V _{ID} = 0.4 sin(4E6π	t) + 0.5 V, D	E at 0 V		15		pF
		RE at 0 V D and DE at 0 V No load	Receiver enabled and driver disabled			4	8	mA
I _{CC}	Supply current	RE at V _{CC} D at V _{CC} DE at 0 V No load	Receiver disabled and driver disabled (standby)			1	5	μΑ
		RE at 0 V D and DE at V _{CC} No load	Receiver e	enabled and driver		9	15.5	mA

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply.

7.7 Power Dissipation Characteristics

	ower Bioorpation one						
	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
P _D Device		R_L = 60 Ω, C_L = 50 pF,	HVD10 (32Mbps)		198	250	
	Device power dissipation	DE at V _{CC} , RE at 0 V, Input to D is a 50% duty-cycle square wave at indicated signaling	HVD11 (10Mbps)		141	176	mW
		rate	HVD12 (500 kbps)		133	161	
_	Ambient eir temperature(1)	High-K board, no airflow	D pkg	-40		116	°C
T _A	Ambient air temperature ⁽¹⁾	No airflow ⁽²⁾	P pkg	-40		123	C
T _{JSD}	Thermal shutdown junction ter	mperature ⁽¹⁾			165		°C

⁽¹⁾ See Section 12.3.1 section for an explanation of these parameters.

⁽²⁾ JESD51-10, Test Boards for Through-Hole Perimeter Leaded Package Thermal Measurements.



7.8 Driver Switching Characteristics

Over recommended operating conditions unless otherwise noted

	PARAMETER		TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
		HVD10		5	8.5	16	
t _{PLH}	Propagation delay time, low-to-high-level output	HVD11		18	25	40	ns
	low-to-riigh-level output	HVD12		135	200	300	
		HVD10		5	8.5	16	
t _{PHL}	Propagation delay time, high-to-low-level output	HVD11		18	25	40	ns
	riigir to low level output	HVD12		135	200	300	
		HVD10		3	4.5	10	
t _r	Differential output signal rise time	HVD11	$R_L = 54 \Omega$, $C_L = 50 pF$ See Figure 8-4	10	20	30	ns
	noo umo	HVD12	Joseph Marie Control	100	100 170 300		
		HVD10		3	4.5	10	
t _f	Differential output signal fall time	HVD11		10	20	30	ns
	idii diiio	HVD12		100	170	300	
		HVD10				1.5	
t _{sk(p)}	Pulse skew (t _{PHL} – t _{PLH})	HVD11				2.5	ns
		HVD12				7	
		HVD10				6	6
t _{sk(pp)} (2)	Part-to-part skew	HVD11				11	ns
		HVD12				100	
	Propagation delay	HVD10				31	
t _{PZH}	time, high-impedance-to-	HVD11				55	ns
	high-level output	HVD12	R_L = 110 Ω, \overline{RE} at 0 V			300	
	Propagation delay	HVD10	See Figure 8-5			25	
t _{PHZ}	time, high-level-to-high-	HVD11				55	ns
	impedance output	HVD12				300	
	Propagation delay	HVD10				26	
t _{PZL}	time, high-impedance-to-	HVD11				55	ns
	low-level output	HVD12	R_L = 110 Ω, \overline{RE} at 0 V			300	
	Propagation delay	HVD10 See Figure 8-6			26		
t _{PLZ}	time, low-level-to-high-	HVD11				75	ns
 I	impedance output	HVD12				400	
t _{PZH}	Propagation delay time, stalevel output	andby-to-high-	R _L = 110 Ω, RE at 3 V See Figure 8-5			6	μs
t _{PZL}	Propagation delay time, sta	andby-to-low-	R _L = 110 Ω, RE at 3 V See Figure 8-6			6	μs

¹⁾ All typical values are at 25°C and with a 3.3-V supply.

⁽²⁾ $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.



7.9 Receiver Switching Characteristics

Over recommended operating conditions unless otherwise noted

	PARAMETER		TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT	
t _{PLH}	Propagation delay time, low-to-high-level output	HVD10		12.5	20	25	ns	
t _{PHL}	Propagation delay time, high-to-low-level output	HVD10		12.5	20	25	115	
t _{PLH}	Propagation delay time, low-to-high-level output	HVD11 HVD12	$V_{ID} = -1.5 \text{ V to } 1.5 \text{ V}$ $C_L = 15 \text{ pF}$	30	55	70	ns	
t _{PHL}	Propagation delay time, high-to-low-level output	HVD11 HVD12	See Figure 8-9	30	55	70	ns	
		HVD10				1.5		
t _{sk(p)}	Pulse skew (t _{PHL} – t _{PLH})	HVD11				4	ns	
		HVD12				4		
		HVD10				8		
t _{sk(pp)} (2)	Part-to-part skew	HVD11				15	ns	
		HVD12				15		
t _r	Output signal rise time		C _L = 15 pF	1	2	5	no	
t _f	Output signal fall time		See Figure 8-9	1	2	5	ns	
t _{PZH} (1)	Output enable time to high	level				15		
t _{PZL} (1)	Output enable time to low I	evel	C _L = 15 pF, DE at 3 V			15	20	
t _{PHZ}	Output disable time from hi	igh level	See Figure 8-10			20	ns	
t _{PLZ}	Output disable time from lo	w level				15		
t _{PZH} ⁽²⁾	Propagation delay time, standby-to-high-level output		C _L = 15 pF, DE at 0			6	110	
t _{PZL} ⁽²⁾	Propagation delay time, sta	andby-to-low-	See Figure 8-11			6	μs	

⁽¹⁾ All typical values are at 25°C and with a 3.3-V supply

7.10 Dissipation Ratings

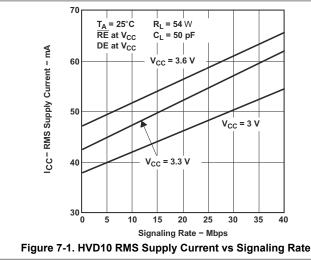
PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ⁽¹⁾ ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING	T _A = 125°C POWER RATING
D ⁽²⁾	597 mW	4.97 mW/°C	373 mW	298 mW	100 mW
D(3)	990 mW	8.26 mW/°C	620 mW	496 mW	165 mW
Р	1290 mW	10.75 mW/°C	806 mW	645 mW	215 mW

- (1) This is the inverse of the junction-to-ambient thermal resistance when board-mounted and with no air flow.
- (2) Tested in accordance with the Low-K thermal metric definitions of EIA/JESD51-3.
- (3) Tested in accordance with the High-K thermal metric definitions of EIA/JESD51-7.

⁽²⁾ t_{sk(pp)} is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.



7.11 Typical Characteristics



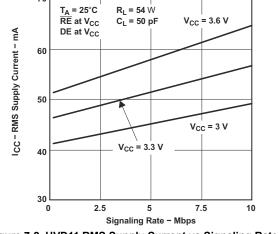


Figure 7-2. HVD11 RMS Supply Current vs Signaling Rate

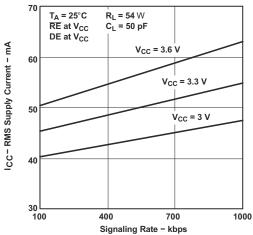


Figure 7-3. HVD12 RMS Supply Current vs Signaling Rate

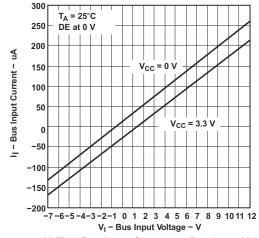


Figure 7-4. HVD10 Bus Input Current vs Bus Input Voltage

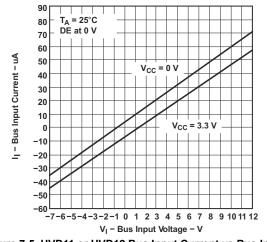


Figure 7-5. HVD11 or HVD12 Bus Input Current vs Bus Input Voltage

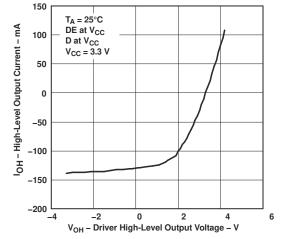


Figure 7-6. High-Level Output Current vs Driver High-Level **Output Voltage**



7.11 Typical Characteristics (continued)

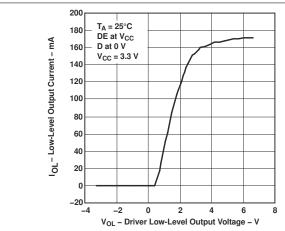


Figure 7-7. Low-Level Output Current vs Driver Low-Level Output Voltage

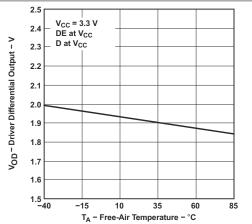


Figure 7-8. Driver Differential Output vs Free-Air Temperature

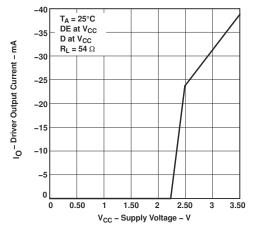


Figure 7-9. Driver Output Current vs Supply Voltage

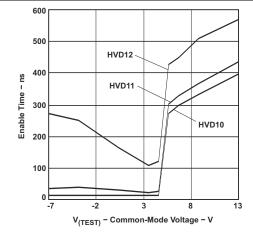
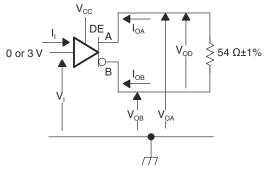


Figure 7-10. Enable Time vs Common-Mode Voltage



8 Parameter Measurement Information



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Figure 8-1. Driver V_{OD} Test Circuit and Voltage and Current Definitions

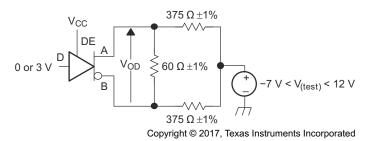
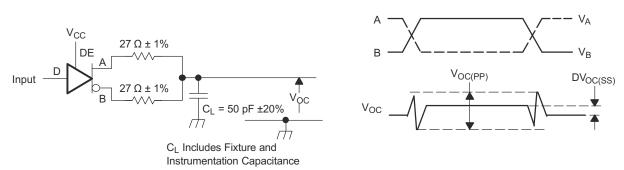


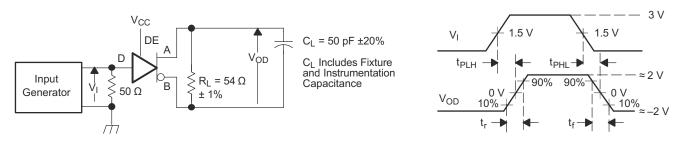
Figure 8-2. Driver V_{OD} With Common-Mode Loading Test Circuit



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Input: PRR = 500 kHz, 50% Duty Cycle, $t_{\rm r}$ < 60 ns, $t_{\rm f}$ < 6 ns $Z_{\rm O}$ = 50 Ω

Figure 8-3. Test Circuit and Definitions for the Driver Common-Mode Output Voltage

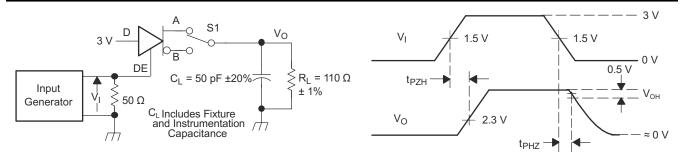


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Generator: PRR = 500 kHz, 50% Duty Cycle, t_r < 60 ns, t_f < 6 ns Z_O = 50 Ω

Figure 8-4. Driver Switching Test Circuit and Voltage Waveforms

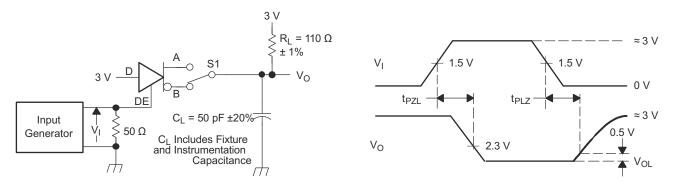




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Generator: PRR = 500 kHz, 50% Duty Cycle, t_r < 60 ns, t_f < 6 ns Z_O = 50 Ω

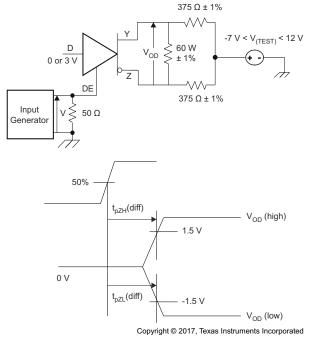
Figure 8-5. Driver High-Level Enable and Disable Time Test Circuit and Voltage Waveforms



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Generator: PRR = 500 kHz, 50% Duty Cycle, t_r < 60 ns, t_f < 6 ns Z_O = 50 Ω

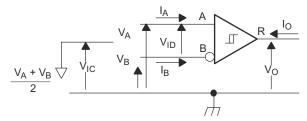
Figure 8-6. Driver Low-Level Output Enable and Disable Time Test Circuit and Voltage Waveforms



The time $t_{PZL(x)}$ is the measure from DE to $V_{OD}(x)$. V_{OD} is valid when it is greater than 1.5 V.

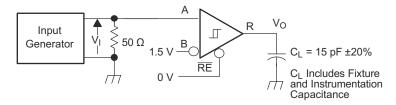
Figure 8-7. Driver Enable Time from DE to V_{OD}





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Figure 8-8. Receiver Voltage and Current Definitions



Generator: PRR = 500 kHz, 50% Duty Cycle, t_r <6 ns, t_f <6 ns, Z_o = 50 Ω

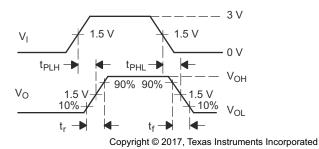
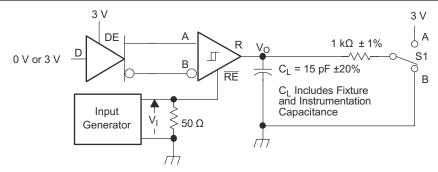


Figure 8-9. Receiver Switching Test Circuit and Voltage Waveforms





Generator: PRR = 500 kHz, 50% Duty Cycle, t_r <6 ns, t_f <6 ns, t_f

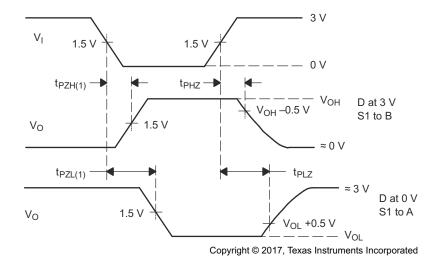
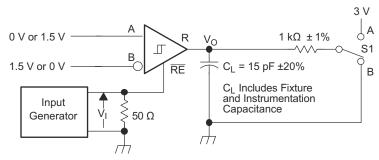


Figure 8-10. Receiver Enable and Disable Time Test Circuit and Voltage Waveforms With Drivers Enabled





Generator: PRR = 100 kHz, 50% Duty Cycle, t_r <6 ns, t_f <6 ns, Z_o = 50 Ω

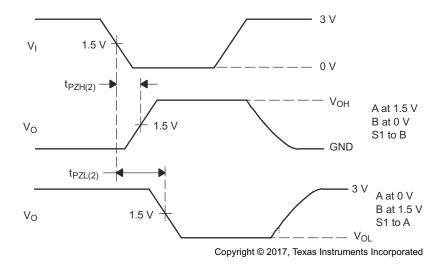
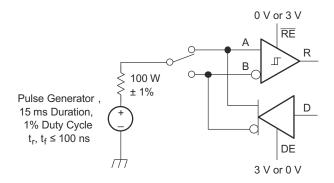


Figure 8-11. Receiver Enable Time From Standby (Driver Disabled)

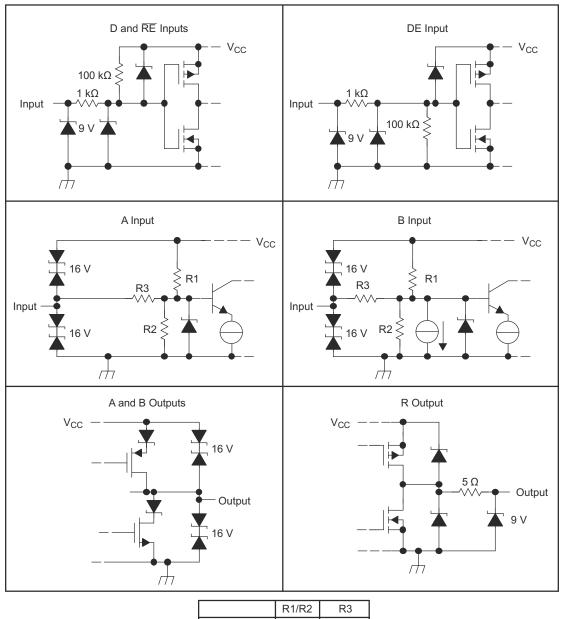


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NOTE: This test is conducted to test survivability only. Data stability at the R output is not specified.

Figure 8-12. Test Circuit, Transient Over Voltage Test





	R1/R2	R3
SN65HVD10	9 kW	45 kW
SN65HVD11	36 kW	180 kW
SN65HVD12	36 kW	180 kW

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Figure 8-13. Equivalent Input and Output Schematic Diagrams



9 Detailed Description

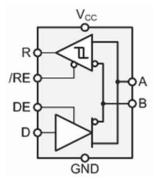
9.1 Overview

The SN65HVD10, SN65HVD11, and SN65HVD12 are 3.3 V, half-duplex, and RS-485 transceivers that are available in 3 speed grades suitable for data transmission up to 32 Mbps, 10 Mbps, and 1 Mbps.

These devices have both active-high driver enables and active-low receiver enables. A standby current of less than

5 µA can be achieved by disabling both driver and receiver.

9.2 Functional Block Diagram



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9.3 Feature Description

Internal ESD protection circuits protect the transceiver bus terminals against ±16-kV Human Body Model (HBM) electrostatic discharges and ±4-kV electrical fast transients (EFT) according to IEC61000-4-4.

The SN65HVD1x half-duplex family provides internal biasing of the receiver input thresholds for open-circuit, bus-idle, or short-circuit fail-safe conditions, as well as a typical receiver hysteresis of 35 mV.

9.4 Device Functional Modes

When the driver enable pin, DE, is logic high, the differential outputs A and B follow the logic states at data input D. A logic high at D causes A to turn high and B to turn low. In this case, the differential output voltage defined as $V_{OD} = V_A - V_B$ is positive. When D is low, the output states reverse, B turns high, A becomes low, and V_{OD} is negative.

When DE is low, both outputs turn high-impedance. In this condition, the logic state at D is irrelevant. The DE pin has an internal pulldown resistor to ground; therefore, when left open, the driver is disabled (high-impedance) by default. The D pin has an internal pullup resistor to V_{CC} ; therefore, when left open while the driver is enabled, output A turns high and B turns low.

OUTPUTS INPUT ENABLE FUNCTION D DE Α В Н Н Η L Actively drive bus High Н L L Н Actively drive bus Low Z Χ Ζ Driver disabled Χ **OPEN** Ζ Ζ Driver disabled by default OPEN Н Н Actively drive bus High by default

Table 9-1. Driver Functions⁽¹⁾

(1) H = high level; L = low level; Z = high impedance; X = irrelevant; ? = indeterminate

When the receiver enable pin, \overline{RE} , is logic low, the receiver is enabled. When the differential input voltage defined as $V_{ID} = V_A - V_B$ is positive and higher than the positive input threshold, V_{IT+} , the receiver output, R,



turns high. When V_{ID} is negative and lower than the negative input threshold, V_{IT-} , the receiver output, R, turns low. If V_{ID} is between V_{IT+} and V_{IT-} , the output is indeterminate.

When \overline{RE} is logic high or left open, the receiver output is high-impedance and the magnitude and polarity of V_{ID} are irrelevant. Internal biasing of the receiver inputs causes the output to go fail-safe-high when the transceiver is disconnected from the bus (open-circuit), the bus lines are shorted (short-circuit), or when the bus is not actively driven (idle bus).

Table 9-2. Receiver Functions(1)

ENABLE RE	OUTPUT R	FUNCTION									
L	Н	Receive valid bus High									
L	?	Indeterminate bus state									
L	L	Receive valid bus Low									
Н	Z	Receiver disabled									
OPEN	Z	Receiver disabled by default									
L	Н	Fail-safe high output									
L	Н	Fail-safe high output									
	RE L L L	RE R L H L ? L L H Z OPEN Z L H									

⁽¹⁾ H = high level; L = low level; Z = high impedance; X = irrelevant; ? = indeterminate

9.4.1 Low-Power Standby Mode

When both the driver and receiver are disabled (DE low and \overline{RE} high) the device is in standby mode. If the enable inputs are in this state for less than 60 ns, the device does not enter standby mode. This guards against inadvertently entering standby mode during driver or receiver enabling. Only when the enable inputs are held in this state for 300 ns or more, the device is assured to be in standby mode. In this low-power standby mode, most internal circuitry is powered down, and the supply current is typically less than 1 μ A. When either the driver or the receiver is re-enabled, the internal circuitry becomes active.



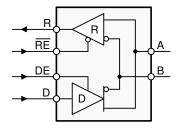
10 Application and Implementation

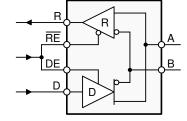
Note

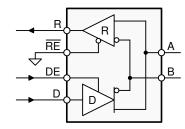
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

10.1 Application Information

The SN65HVD10, 'HVD11, and 'HVD12 are half-duplex RS-485 transceivers commonly used for asynchronous data transmissions. The driver and receiver enable pins allow the configuration of different operating modes.







- a) Independent driver and receiver enable signals
- b) Combined enable signals for use as directional control pin
- c) Receiver always on

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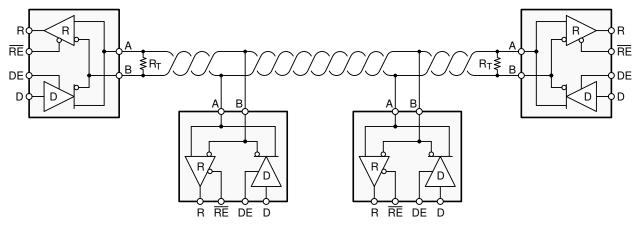
Figure 10-1. Half-Duplex Transceiver Configurations

- 1. Using independent enable lines provides the most flexible control, as it allows the driver and the receiver to be turned on and off individually. While this configuration requires two control lines, it allows selective listening into the bus traffic, whether the driver is transmitting data or not.
- 2. Combining the enable signals simplify the interface to the controller, by forming a single direction-control signal. In this configuration, the transceiver operates as a driver when the direction-control line is high, and as a receiver when the direction-control line is low.
- 3. Only one line is required when connecting the receiver-enable input to ground and controlling only the driver-enable input. In this configuration, a node not only receives the data from the bus, but also the data it sends and can verify that the correct data have been transmitted.



10.2 Typical Application

An RS-485 bus consists of multiple transceivers connected in parallel to a bus cable. To eliminate line reflections, each cable end is terminated with a termination resistor, R_T , whose value matches the characteristic impedance, Z_0 , of the cable. This method, known as parallel termination, allows higher data rates over a longer cable length.



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Figure 10-2. Typical RS-485 Network With Half-Duplex Transceivers

10.2.1 Design Requirements

RS-485 is a robust electrical standard suitable for long-distance networking, that may be used in a wide range of applications with varying requirements, such as distance, data rate, and number of nodes.

10.2.1.1 Data Rate and Bus Length

There is an inverse relationship between data rate and bus length, meaning the higher the data rate, the shorter the cable length; and conversely, the lower the data rate, the longer the cable may be without introducing data errors. While most RS-485 systems use data rates between 10 kbps and 100 kbps, some applications require data rates up to 250 kbps at distances of 4000 feet and longer. Longer distances are possible by allowing small signal jitter of up to 5 or 10%.

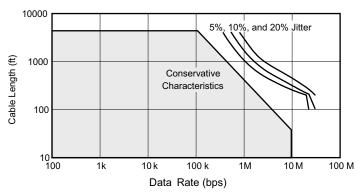


Figure 10-3. Cable Length vs Data Rate Characteristic



10.2.1.2 Stub Length

When connecting a node to the bus, the distance between the transceiver inputs and the cable trunk, known as the stub, should be as short as possible. Stubs present a nonterminated piece of bus line, which can introduce reflections as the length of the stub increases. As a general guideline, the electrical length or round-trip delay of a stub should be less than one-tenth of the rise time of the driver, therefore giving a maximum physical stub length as shown in Equation 1.

$$L_{(STUB)} \le 0.1 \times t_r \times v \times c \tag{1}$$

where

- t_r is the 10/90 rise time of the driver
- v is the signal velocity of the cable or trace as a factor of c
- c is the speed of light (3 × 10⁸ m/s)

Per Equation 1, Table 10-1 lists the maximum cable-stub lengths for the minimum-driver output rise-times of the SN65HVD1x full-duplex family of transceivers for a signal velocity of 78%.

		=			
DEVICE	MINIMUM DRIVER OUTPUT	MAXIMUM STUB LENGTH			
DEVICE	RISE TIME (ns)	(m)	(ft)		
SN65HVD10	3	0.07	0.23		
SN65HVD11	10	0.23	0.75		
SN65HVD12	100	2.34	7.67		

Table 10-1. Maximum Stub Length

10.2.1.3 Bus Loading

The RS-485 standard specifies that a compliant driver must be able to driver 32 unit loads (UL), where 1 unit load represents a load impedance of approximately 12 k Ω . SN65HVD11 and HVD12 are both 1/8 UL transceivers, which means that up to 256 receivers can be connected to the bus. The SN65HVD10 is a 1/4 UL transceiver, and up to 64 receivers can be connected to the bus.

10.2.1.4 Receiver Fail-safe

The differential receivers of the SN65HVD1x family are fail-safe to invalid bus states caused by:

- · Open bus conditions, such as a disconnected connector
- Shorted bus conditions, such as cable damage shorting the twisted-pair together
- Idle bus conditions that occur when no driver on the bus is actively driving.

In any of these cases, the differential receiver will output a fail-safe logic High state so that the output of the receiver is not indeterminate.

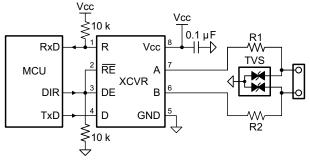
Receiver fail-safe is accomplished by offsetting the receiver thresholds, such that the input indeterminate range does not include zero volts differential. To comply with the RS-422 and RS-485 standards, the receiver output must output a High when the differential input V_{ID} is more positive than +200 mV, and must output a Low when V_{ID} is more negative than -200 mV. The receiver parameters which determine the fail-safe performance are $V_{IT(+)}$ and $V_{IT(-)}$. As shown in Section 7.6, differential signals more negative than -200 mV will always cause a Low receiver output, and differential signals more positive than +200 mV will always cause a High receiver output.

When the differential input signal is close to zero, it is still above the maximum $V_{IT(+)}$ threshold of -10 mV, and the receiver output will be High.



10.2.2 Detailed Design Procedure

To protect bus nodes against high-energy transients, the implementation of external transient protection devices is therefore necessary. Figure 10-4 shows a protection circuit against 10-kV ESD (IEC 61000-4-2), 4-kV EFT (IEC 61000-4-4), and 1-kV surge (IEC 61000-4-5) transients.



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Figure 10-4. Transient Protection Against ESD, EFT, and Surge Transients

	Table 10-2. Bill of Materials									
DEVICE	FUNCTION	ORDER NUMBER	MANUFACTURER							
XCVR	3.3-V, full-duplex RS-485 transceiver	SN65HVD1xD	TI							
R1, R2	10-Ω, pulse-proof, thick-film resistor	CRCW0603010RJNEAHP	Vishay							
TVS	Bidirectional 400-W	CDSOT23-SM712	Bourns							

Table 10-2. Bill of Materials

10.2.3 Application Curve

Figure 10-5 demonstrates operation of the SN65HVD12 at a signaling rate of 250 kbps. Two SN65HVD12 transceivers are used to transmit data through a 2,000 foot (600 m) segment of Commscope 5524 category 5e+ twisted pair cable. The bus is terminated at each end by a $100-\Omega$ resistor, matching the cable characteristic impedance.

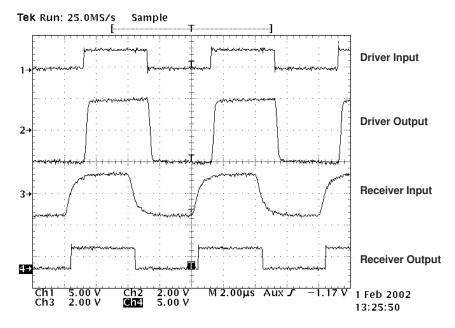


Figure 10-5. SN65HVD12 Input and Output Through 2000 Feet of Cable



11 Power Supply Recommendations

To assure reliable operation at all data rates and supply voltages, each supply must be buffered with a 100-nF ceramic capacitor located as close to the supply pins as possible. The TPS76333 linear voltage regulator is suitable for the 3.3-V supply.

12 Layout

12.1 Layout Guidelines

On-chip IEC-ESD protection is sufficient for laboratory and portable equipment, but never sufficient for EFT and surge transients occurring in industrial environments. Therefore, robust and reliable bus node design requires the use of external transient protection devices.

It is because ESD and EFT transients have a wide frequency bandwidth from approximately 3 MHz to 3 GHz, that high-frequency layout techniques must be applied during PCB design.

- 1. Place the protection circuitry close to the bus connector to prevent noise transients from entering the board.
- 2. Use V_{CC} and ground planes to provide low-inductance. Note that high-frequency currents follow the path of least inductance and not the path of least impedance.
- 3. Design the protection components into the direction of the signal path. Do not force the transient currents to divert from the signal path to reach the protection device.
- 4. Apply 100-nF to 220-nF bypass capacitors as close as possible to the V_{CC} pins of transceiver, UART, and controller ICs on the board.
- 5. Use at least two vias for V_{CC} and ground connections of bypass capacitors and protection devices to minimize effective via-inductance.
- 6. Use $1-k\Omega$ to $10-k\Omega$ pull-up or pull-down resistors to enable lines to limit noise currents in these lines during transient events.
- 7. Insert pulse-proof series resistors into the A and B bus lines if the TVS clamping voltage is higher than the specified maximum voltage of the transceiver bus terminals. These resistors limit the residual clamping current into the transceiver and prevent it from latching up.
- 8. While pure TVS protection is sufficient for surge transients up to 1 kV, higher transients require metal-oxide varistors (MOVs) which reduce the transients to a few hundred volts of clamping voltage, and transient blocking units (TBUs) that limit transient current to less than 1 mA.



12.2 Layout Example

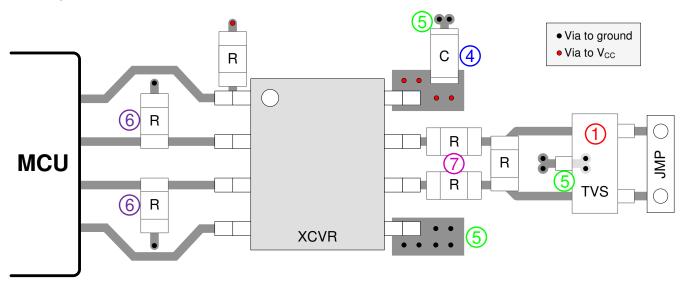


Figure 12-1. SN65HVD1x Layout Example

12.3 Thermal Considerations

12.3.1 Thermal Characteristics of IC Packages

R_{0JA} (Junction-to-Ambient Thermal Resistance) is defined as the difference in junction temperature to ambient temperature divided by the operating power.

 $R_{\theta,JA}$ is not a constant and is a strong function of:

- the PCB design (50% variation)
- altitude (20% variation)
- device power (5% variation)

 $R_{\theta JA}$ can be used to compare the thermal performance of packages when specific test conditions are defined and used. Standardized testing includes specification of PCB construction, test chamber volume, sensor locations, and the thermal characteristics of holding fixtures. $R_{\theta JA}$ is often misused when it is used to calculate junction temperatures for other installations.

TI uses two test PCBs as defined by JEDEC specifications. The low-k board gives average in-use condition thermal performance, and it consists of a single copper trace layer 25 mm long and 2-oz thick. The high-k board gives best case in-use condition, and it consists of two 1-oz buried power planes with a single copper trace layer 25 mm long and 2-oz thick. A 4% to 50% difference in $R_{\theta,JA}$ can be measured between these two test cards.

 $R_{ text{ ilegal}JC}$ (Junction-to-Case Thermal Resistance) is defined as the difference in junction temperature to case divided by the operating power. It is measured by putting the mounted package up against a copper block cold plate to force heat to flow from the die, through the mold compound into the copper block.

 $R_{\theta JC}$ is a useful thermal characteristic when a heat sink is applied to package. It is not a useful characteristic to predict junction temperature, because it provides pessimistic numbers if the case temperature is measured in a nonstandard system and junction temperatures are backed out. It can be used with $R_{\theta JB}$ in 1-dimensional thermal simulation of a package system.

 $R_{\theta JB}$ (Junction-to-Board Thermal Resistance) is defined as the difference in the junction temperature and the PCB temperature at the center of the package (closest to the die) when the PCB is clamped in a cold-plate structure. $R_{\theta JB}$ is only defined for the high-k test card.

 $R_{\theta JB}$ provides an overall thermal resistance between the die and the PCB. It includes a bit of the PCB thermal resistance (especially for BGAs with thermal balls) and can be used for simple 1-dimensional network analysis of package system, see Figure 12-2.



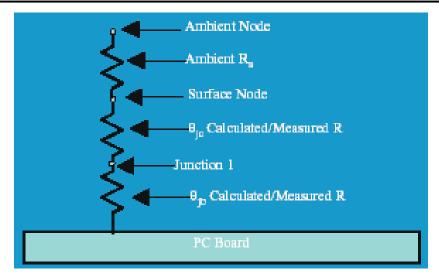


Figure 12-2. PCB Thermal Resistances



13 Device and Documentation Support

13.1 Device Support

13.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to order now.

Table 13-1. Related Links

PARTS	PRODUCT FOLDER	ORDER NOW	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
SN65HVD10	Click here	Click here	Click here	Click here	Click here
SN65HVD11	Click here	Click here	Click here	Click here	Click here
SN65HVD12	Click here	Click here	Click here	Click here	Click here
SN75HVD10	Click here	Click here	Click here	Click here	Click here
SN75HVD11	Click here	Click here	Click here	Click here	Click here
SN75HVD12	Click here	Click here	Click here	Click here	Click here

13.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

13.4 Support Resources

TI E2E[™] support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the guick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

13.5 Trademarks

TI E2E[™] is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

13.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

13.7 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
SN65HVD10D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	VP10
SN65HVD10DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP10
SN65HVD10DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP10
SN65HVD10DRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP10
SN65HVD10P	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD10
SN65HVD10P.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD10
SN65HVD10QD	Obsolete	Production	SOIC (D) 8	-	=	Call TI	Call TI	-40 to 125	VP10Q
SN65HVD10QDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP10Q
SN65HVD10QDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP10Q
SN65HVD10QDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP10Q
SN65HVD10QDRG4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP10Q
SN65HVD11D	Obsolete	Production	SOIC (D) 8	-	=	Call TI	Call TI	-40 to 85	VP11
SN65HVD11DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP11
SN65HVD11DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP11
SN65HVD11DRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP11
SN65HVD11P	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD11
SN65HVD11P.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD11
SN65HVD11QD	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 125	VP11Q
SN65HVD11QDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP11Q
SN65HVD11QDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP11Q
SN65HVD11QDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP11Q
SN65HVD11QDRG4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	VP11Q
SN65HVD12D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	VP12
SN65HVD12DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP12
SN65HVD12DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP12
SN65HVD12DRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP12
SN65HVD12P	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD12
SN65HVD12P.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD12
SN75HVD10D	Obsolete	Production	SOIC (D) 8	-	=	Call TI	Call TI	0 to 70	VN10



0 to 70

0 to 70

0 to 70

9-Nov-2025

75HVD12

75HVD12

75HVD12



SN75HVD12P

SN75HVD12P.A

SN75HVD12PE4

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Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking
	, ,					(4)	(5)		, ,
SN75HVD10DR	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	0 to 70	VN10
SN75HVD10P	Obsolete	Production	PDIP (P) 8	-	-	Call TI	Call TI	0 to 70	75HVD10
SN75HVD11D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	0 to 70	VN11
SN75HVD12D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	0 to 70	VN12
SN75HVD12DR	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	0 to 70	VN12

Yes

Yes

Yes

NIPDAU

NIPDAU

NIPDAU

N/A for Pkg Type

N/A for Pkg Type

N/A for Pkg Type

50 | TUBE

50 | TUBE

50 | TUBE

Active

Active

Active

Production

Production

Production

PDIP (P) | 8

PDIP (P) | 8

PDIP (P) | 8

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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⁽¹⁾ Status: For more details on status, see our product life cycle.

⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

PACKAGE OPTION ADDENDUM

www.ti.com 9-Nov-2025

OTHER QUALIFIED VERSIONS OF SN65HVD10, SN65HVD11, SN65HVD12:

● Enhanced Product : SN65HVD10-EP, SN65HVD12-EP

NOTE: Qualified Version Definitions:

• Enhanced Product - Supports Defense, Aerospace and Medical Applications

PACKAGE MATERIALS INFORMATION

www.ti.com 6-Aug-2025

TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SN65HVD10DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD10QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD10QDRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD11DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD11QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD11QDRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD12DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



www.ti.com 6-Aug-2025



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
SN65HVD10DR	SOIC	D	8	2500	340.5	336.1	25.0
SN65HVD10QDR	SOIC	D	8	2500	340.5	336.1	25.0
SN65HVD10QDRG4	SOIC	D	8	2500	353.0	353.0	32.0
SN65HVD11DR	SOIC	D	8	2500	340.5	336.1	25.0
SN65HVD11QDR	SOIC	D	8	2500	353.0	353.0	32.0
SN65HVD11QDRG4	SOIC	D	8	2500	353.0	353.0	32.0
SN65HVD12DR	SOIC	D	8	2500	340.5	336.1	25.0

PACKAGE MATERIALS INFORMATION

www.ti.com 6-Aug-2025

TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
SN65HVD10P	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD10P.A	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD11P	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD11P.A	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD12P	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD12P.A	Р	PDIP	8	50	506	13.97	11230	4.32
SN75HVD12P	Р	PDIP	8	50	506	13.97	11230	4.32
SN75HVD12P.A	Р	PDIP	8	50	506	13.97	11230	4.32
SN75HVD12PE4	Р	PDIP	8	50	506	13.97	11230	4.32



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



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