











SN55HVD233-SP

SLLSEI2A - SEPTEMBER 2017 - REVISED DECEMBER 2017

# SN55HVD233-SP 3.3-V Radiation Hardened CAN Transceiver

#### **Features**

- QMLV (QML Class V) Radiation Hardness Assured (RHA) MIL-PRF 38535 Qualified, SMD 5962L1420901VXC
  - Single Event Latch-up (SEL) Immune to 86 MeV-cm<sup>2</sup>/mg at 125°C
  - Total Ionizing Dose (TID) immune up 50 kRad (Si) at Low Dose Rate
  - Qualified over the Military Temperature Range (-55°C to 125°C)
  - High-Performance 8-Pin Ceramic Flat Pack (HKX)
- Compatible With ISO 11898-2
- Bus Pins Fault Protection Exceeds ±16 V
- Bus Pins ESD Protection Exceeds ±16 kV HBM
- Data Rates up to 1 Mbps
- Extended –7-V to 12-V Common Mode Range
- High Input Impedance Allows for 120 Nodes
- LVTTL I/Os are 5-V Tolerant
- Adjustable Driver Transition Times for Improved Signal Quality
- Unpowered Node Does Not Disturb the Bus
- Low Current Standby Mode, 200-µA Typical
- Loopback for Diagnostic Functions
- Thermal Shutdown Protection
- Power Up and Power Down With Glitch-Free Bus Inputs and Outputs
  - High Input Impedance With Low V<sub>CC</sub>
  - Monolithic Output During Power Cycling

# 2 Applications

- Spacecraft Backplane Data Bus Communication and Control
- CAN Bus Standards Such As CANopen, DeviceNet, CAN Kingdom, ISO 11783, NMEA 2000, SAE J1939

# 3 Description

SN55HVD233-SP is used in spacecraft applications employing the controller area network (CAN) serial communication physical layer in accordance with the ISO 11898 standard. As a CAN transceiver, the device provides transmit and receive capability between the differential CAN bus and a CAN controller, with signaling rates up to 1 Mbps.

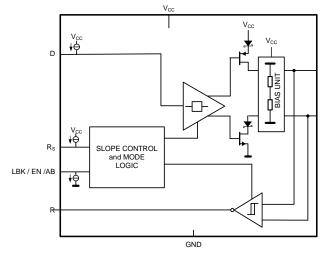
Designed for operation in especially harsh radiation environments, the SN55HVD233-SP features crosswire, overvoltage, and loss of ground protection to overtemperature (thermal protection. This device operates over a wide -7-V to 12-V common mode range. This transceiver is the interface between the host CAN controller on the microprocessor, FPGA, or ASIC, and the differential CAN bus used in satellite applications.

#### Device Information<sup>(1)</sup>

PART NUMBER	GRADE	PACKAGE
5962L1420901VXC	QMLV RHA [50 kRad]	8-lead CFP [HKX] 6.48 mm × 6.48 mm
HVD233HKX/EM <sup>(2)</sup>	Engineering Samples	8-lead CFP [HKX] 6.48 mm × 6.48 mm
SN55HVD233EVM- CVAL	Ceramic Evaluation Board	

- (1) For all available packages, see the orderable addendum at the end of the data sheet.
- These units are intended for engineering evaluation only. They are processed to a noncompliant flow. These units are not suitable for qualification, production, radiation testing or flight use. Parts are not warranted for performance over the full MIL specified temperature range of -55°C to 125°C or operating life.

## Simplified Schematic



Copyright © 2017, Texas Instruments Incorporated



# **Table of Contents**

2 Applications	
3 Description	1
4 Revision History	1
5 Description (continued)	tion 2
6 Pin Configuration and Functions	2
7 Specifications	2
7.1 Absolute Maximum Ratings	tions 2
7.1 Absolute Maximum Ratings	2
7.3 Recommended Operating Conditions 4 12.1 Layout Guidelines	2
7.4 Thermal Information	
7.5 Driver Electrical Characteristics 6 13 Device and Documentation	Support 2
7.6 Receiver Electrical Characteristics	= =
7.7 Driver Switching Characteristics	2
7.8 Receiver Switching Characteristics	2
7.9 Device Switching Characteristics	ıtion2
7.10 Typical Characteristics	2
8 Parameter Measurement Information	

# 4 Revision History

Ch	hanges from Original (September 2017) to Revision A	Page	
•	Changed the device status from Advance Information to Production Data		1



# 5 Description (continued)

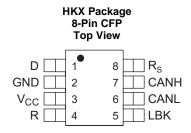
Modes: The  $R_S$ , pin 8 of the SN55HVD233-SP, provides for three modes of operation: high-speed, slope control, or low-power standby mode. The user selects the high-speed mode of operation by connecting pin 8 directly to ground, allowing the driver output transistors to switch on and off as fast as possible with no limitation on the rise and fall slope. The user can adjust the rise and fall slope by connecting a resistor to ground at pin 8, because the slope is proportional to the pin's output current. Slope control is implemented with a resistor values of 0  $\Omega$  to achieve a single ended slew rate of approximately 38 V/ $\mu$ s up to a value of 50 k $\Omega$  to achieve approximately 4 V/ $\mu$ s slew rate. For more information about slope control, refer to the *Application and Implementation* section.

The SN55HVD233-SP enters a low-current standby (listen-only) mode during which the driver is switched off and the receiver remains active if a high logic level is applied to pin 8. The local protocol controller reverses this low-current standby mode when it needs to transmit to the bus. For more information on the loopback mode, refer to the *Application Information* section.

Loopback: A logic high on the loopback LBK pin 5 of the SN55HVD233-SP places the bus output and bus input in a high-impedance state. The remaining circuit remains active and available for driver-to-receiver loopback, self-diagnostic node functions without disturbing the bus.

CAN bus states: The CAN bus has two states during powered operation of the device: dominant and recessive. A dominant bus state is when the bus is driven differentially, corresponding to a logic low on the D and R pin. A recessive bus state is when the bus is biased to  $V_{CC}$  / 2 through the high-resistance internal input resistors  $R_{IN}$  of the receiver, corresponding to a logic high on the D and R pins (see Bus States (Physical Bit Representation) and Simplified Recessive Common Mode Bias and Receiver).

# 6 Pin Configuration and Functions



**Pin Functions** 

PIN		TYPE	DESCRIPTION
NAME	NO.	ITFE	DESCRIPTION
D	1	I	CAN transmit data input (LOW for dominant and HIGH for recessive bus states), also called TXD, driver input.
GND	2	GND	Ground connection.
$V_{CC}$	3	Supply	Transceiver 3.3-V supply voltage.
R	4	0	CAN receive data output (LOW for dominant and HIGH for recessive bus states), also called RXD, receiver output.
LBK	5	I	Loopback mode input pin.
CANL	6	I/O	Low-level CAN bus line.
CANH	7	I/O	High-level CAN bus line.
RS	8	I	Mode select pin: Tie to GND = high-speed mode, Strong pullup to $V_{CC}$ = low power mode, $0$ - $\Omega$ to $50$ - $k\Omega$ pulldown to GND = slope control mode.

Product Folder Links: SN55HVD233-SP



# 7 Specifications

## 7.1 Absolute Maximum Ratings

over operating junction temperature unless otherwise noted (1)(2)

		MIN	MAX	UNIT
$V_{CC}$	Supply voltage	-0.3	7	V
	Voltage at any bus pin (CANH or CANL)	-16	16	V
	Voltage input, transient pulse, CANH and CANL, through 100 $\Omega$ (see Figure 18)	-100	100	V
VI	Input voltage, (D, RS, LBK)	-0.5	7	V
Vo	Output voltage, (R)	-0.5	7	V
Io	Receiver output current	-10	10	mA
$T_{J}$	Operating junction temperature		150	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C

Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

# 7.2 ESD Ratings

				VALUE	UNIT
		Human had a madel (HDM) man ANCI/ECDA/JEDEC JC 004 (1)	CANH, CANL, and GND	±14000	
V <sub>(ESD)</sub>	discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	Other pins	±4000	V
		Charged-device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>		±500	

JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

# 7.3 Recommended Operating Conditions

			MIN	NOM MAX	UNIT
$V_{CC}$	Supply voltage		3	3.6	V
	Voltage at any bus pin (separat	age at any bus pin (separately or common mode)		12	V
$V_{IH}$	High-level input voltage	D, LBK	2	5.5	V
$V_{IL}$	Low-level input voltage	D, LBK	0	0.8	V
$V_{ID}$	Differential input voltage		-6	6	V
	Resistance from RS to ground for slope control		0	50	kΩ
$V_{I(RS)}$	Input voltage at RS for standby		0.75 V <sub>CC</sub>	5.5	V
	Lligh lovel output ourrent	Driver	-50		A
I <sub>OH</sub>	High-level output current	Receiver	-10		mA
	Lour lough output ourrent	Driver		50	A
I <sub>OL</sub>	Low-level output current	Receiver		10	mA
$T_{J}$	Operating junction temperature	(1)	<b>-</b> 55	125	°C

(1) Maximum junction temperature operation is allowed as long as the device maximum junction temperature is not exceeded.

Submit Documentation Feedback

Copyright © 2017, Texas Instruments Incorporated

All voltage values, except differential I/O bus voltages, are with respect to network ground pin.



#### 7.4 Thermal Information

		SN55HVD233-SP	
	THERMAL METRIC <sup>(1)(2)</sup>	HKX (CFP)	UNIT
		8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	97.1	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	21.5	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	79.1	°C/W
ΨЈТ	Junction-to-top characterization parameter	13.7	°C/W
ΨЈВ	Junction-to-board characterization parameter	73.6	°C/W
R <sub>0</sub> JC(bot)	Junction-to-case (bottom) thermal resistance	7.0	°C/W

 <sup>(1)</sup> All values except R<sub>θJC</sub> were taken on a JEDEC-51 standard High-K PCB using a nominal lead form. Differences in lead form, component density, or PCB design can affect these values.
 (2) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application

Product Folder Links: SN55HVD233-SP

report, SPRA953.



# 7.5 Driver Electrical Characteristics

The specifications shown below are valid across temperature range of –55°C to 125°C pre-radiation and 25°C post-radiation. When different, the post-radiation values are shown in a separate row specified by the corresponding RHA level (L = 50 krad).

	PARAMETER		TEST CONDITIONS	SUBGROUP <sup>(1)</sup>	MIN	TYP <sup>(2)</sup>	MAX	UNIT
	Bus output	CANH	V <sub>(D)</sub> = 0 V, V <sub>(RS)</sub> = 0 V, see Figure 12 and	[1, 2, 3]	2.4		$V_{CC}$	
$V_{O(D)}$	voltage (dominant)	CANL	Figure 13	[1, 2, 3]	0.5		1.25	V
	Bus output	CANH	$V_{(D)} = 3 \text{ V}, V_{(RS)} = 0 \text{ V}, \text{ see Figure 12 and}$			2.3		
Vo	voltage (recessive)	CANL	Figure 13			2.3		V
			V <sub>(D)</sub> = 0 V, V <sub>(RS)</sub> = 0 V, see Figure 12 and	[1, 2, 3]	1.5	2	3	
$V_{OD(D)}$	Differential outp	ut voltage	Figure 13	[1, 2, 3]	1.4			V
· OD(D)	(dominant)		$V_{(D)}$ = 0 V, $V_{(RS)}$ = 0 V, see Figure 13 and Figure 14	[1, 2, 3]	1.2	2	3	
V <sub>OD</sub>	Differential output voltage (recessive)		$V_{(D)} = 3 \text{ V}, V_{(RS)} = 0 \text{ V}, \text{ see Figure 12 and}$ Figure 13	[1, 2, 3]	-120		12	mV
OB			V <sub>(D)</sub> = 3 V, V <sub>(RS)</sub> = 0 V, no load	[1, 2, 3]	-0.5		0.05	V
V <sub>OC(pp)</sub>	Peak-to-peak common- mode output voltage		See Figure 20			1		V
I <sub>IH</sub>	High-level input current	D, LBK	V <sub>(D)</sub> = 2 V	[1, 2, 3]	-30		30	μА
I <sub>IL</sub>	Low-level input current	D, LBK	V <sub>(D)</sub> = 0.8 V	[1, 2, 3]	-30		30	μА
			V <sub>(CANH)</sub> = -7 V, CANL open, see Figure 23	[1, 2, 3]	-250			
L	Short-circuit out	nut current	V <sub>(CANH)</sub> = 12 V, CANL open, see Figure 23	[1, 2, 3]			1	mA
los	Short-circuit out	put current	$V_{(CANL)} = -7 \text{ V, CANH open, see Figure 23}$	[1, 2, 3]	-1			ША
			V <sub>(CANL)</sub> = 12 V, CANH open, see Figure 23	[1, 2, 3]			250	
Co	Output capacitance		See receiver input capacitance					
I <sub>IRS(s)</sub>	RS input current for standby		$V_{(RS)} = 0.75 V_{CC}$	[1, 2, 3]	-10			μΑ
		Standby	$V_{(RS)} = V_{CC}, V_{(D)} = V_{CC}, V_{(LBK)} = 0 \text{ V}$	[1, 2, 3]		200	600	μА
Icc	Supply current	Dominant	V <sub>(D)</sub> = 0 V, no load, V <sub>(LBK)</sub> = 0 V, RS = 0 V	[1, 2, 3]			6	mΛ
		Recessive	$V_{(D)} = V_{CC}$ , no load, $V_{(LBK)} = 0 \text{ V}$ , $V_{(RS)} = 0 \text{ V}$	[1, 2, 3]			6	mA

<sup>(1)</sup> For subgroup definitions, please see Table 1.

#### 7.6 Receiver Electrical Characteristics

The specifications shown below are valid across temperature range of  $-55^{\circ}$ C to 125°C pre-radiation and 25°C post-radiation. When different, the post-radiation values are shown in a separate row specified by the corresponding RHA level (L = 50 krad).

	PARAMETER	TEST CONDITIONS	SUBGROUP <sup>(1)</sup>	MIN	TYP <sup>(2)</sup>	MAX	UNIT
V <sub>IT+</sub>	Positive-going input threshold voltage		[1, 2, 3]		750	900	mV
V <sub>IT-</sub>	Negative-going input threshold voltage	V <sub>(LBK)</sub> = 0 V, see Table 2	[1, 2, 3]	500	650		mV
V <sub>hys</sub>	Hysteresis voltage (V <sub>IT+</sub> – V <sub>IT-</sub> )				100		mV
$V_{OH}$	High-level output voltage	I <sub>O</sub> = -4 mA, see Figure 17	[1, 2, 3]	2.4			V
$V_{OL}$	Low-level output voltage	I <sub>O</sub> = 4 mA, see Figure 17	[1, 2, 3]			0.4	V

<sup>2)</sup> All typical values are at 25°C and with a 3.3-V supply.

<sup>(1)</sup> For subgroup definitions, please see Table 1.

<sup>(2)</sup> All typical values are at 25°C and with a 3.3-V supply.



## **Receiver Electrical Characteristics (continued)**

The specifications shown below are valid across temperature range of  $-55^{\circ}$ C to 125°C pre-radiation and 25°C post-radiation. When different, the post-radiation values are shown in a separate row specified by the corresponding RHA level (L = 50 krad).

	PARAME	TER	TEST CON	DITIONS	SUBGROUP <sup>(1)</sup>	MIN	TYP <sup>(2)</sup> M	IAX	UNIT
			$V_{(CANH)}$ or $V_{(CANL)} = 12 \text{ V}$		[1, 2, 3]	150	ţ	500	
	Puo input our	ont	$V_{(CANH)}$ or $V_{(CANL)} = 12 \text{ V}$ , $V_{CC} = 0 \text{ V}$	Other bus pin = 0 V, $V_{(D)} = 3 V$ ,	[1, 2, 3]	150	(	600	^
I <sub>I</sub>	Bus input curr	ent	CANH or CANL = $-7 \text{ V}$	$V_{(LBK)} = 0 V,$	[1, 2, 3]	-610	_	100	μΑ
			CANH or CANL = -7 V, V <sub>CC</sub> = 0 V	V <sub>(RS)</sub> = 0 V	[1, 2, 3]	-450	_	100	
Cı	Input capacitance (CANH or CANL)		Pin-to-ground, $V_1 = 0.4 \sin(4E6\pi t) + 0.5 \text{ V},$ $V_{(D)} = 3 \text{ V}, V_{(LBK)} = 0 \text{ V}$				40		pF
C <sub>ID</sub>	Differential input capacitance		Pin-to-pin, $V_1 = 0.4 \sin(4E6 V_{(D)} = 3 V, V_{(LBK)} = 0 V$	πt) + 0.5 V,			20		pF
$R_{\text{ID}}$	Differential inp	out resistance			[4, 5, 6]	40	,	105	kΩ
R <sub>IN</sub>	Input resistance (CANH or CANL)		$V_{(D)} = 3 \text{ V}, V_{(LBK)} = 0 \text{ V}$		[4, 5, 6]	20		55	kΩ
I <sub>CC</sub>		Standby	$V_{(RS)} = V_{CC}, V_{(D)} = V_{CC}, V_{(D)}$	<sub>_BK)</sub> = 0 V	[1, 2, 3]		200	600	μΑ
	Supply current	Dominant	$V_{(D)} = 0 \text{ V, no load, } V_{(RS)} =$	0 V, V <sub>(LBK)</sub> = 0 V	[1, 2, 3]			6	mA
		Recessive	$V_{(D)} = V_{CC}$ , no load, $V_{(RS)} =$	= 0 V, V <sub>(LBK)</sub> = 0 V	[1, 2, 3]			6	mA

# 7.7 Driver Switching Characteristics

The specifications shown below are valid across temperature range of –55°C to 125°C pre-radiation and 25°C post-radiation. When different, the post-radiation values are shown in a separate row specified by the corresponding RHA level (L = 50 krad).

	PARAMETER	TEST CONDITIONS	SUBGROUP <sup>(1)</sup>	MIN TYP <sup>(2)</sup>	MAX	UNIT	
		V <sub>(RS)</sub> = 0 V, see Figure 15	[9, 10, 11]	35	85		
t <sub>PLH</sub>	Propagation delay time,  low-to-high-level output	RS with 10 $k\Omega$ to ground, see Figure 15	[9, 10, 11]	70	125	ns	
	ion to high lovel output	RS with 50 $k\Omega$ to ground, see Figure 15	[9, 10, 11]	500	870		
		V <sub>(RS)</sub> = 0 V, see Figure 15	[9, 10, 11]	70	120		
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output	RS with 10 k $\Omega$ to ground, see Figure 15	[9, 10, 11]	130	180	ns	
	riigir to low level output	RS with 50 k $\Omega$ to ground, see Figure 15	[9, 10, 11]	870	1200		
		V <sub>(RS)</sub> = 0 V, see Figure 15		35			
t <sub>sk(p)</sub>	Pulse skew ( t <sub>PHL</sub> - t <sub>PLH</sub>  )	RS with 10 kΩ to ground, see Figure 15		60		ns	
		RS with 50 kΩ to ground, see Figure 15		370			
t <sub>r</sub>	Differential output signal rise time	V OV see Figure 45	[9, 10, 11]	20	70	ns	
t <sub>f</sub>	Differential output signal fall time	V <sub>(RS)</sub> = 0 V, see Figure 15	[9, 10, 11]	20	70	ns	
t <sub>r</sub>	Differential output signal rise time	D0 with 40 L0 to second and Figure 45	[9, 10, 11]	30	135	ns	
t <sub>f</sub>	Differential output signal fall time	RS with 10 kΩ to ground, see Figure 15	[9, 10, 11]	30	135	ns	
t <sub>r</sub>	Differential output signal rise time	DO with 50 LO to second and Figure 45	[9, 10, 11]	350	1400	ns	
t <sub>f</sub>	Differential output signal fall time	RS with 50 kΩ to ground, see Figure 15	[9, 10, 11]	350	1400	ns	
t <sub>en(s)</sub>	Enable time from standby to dominant	See Figure 19	[9, 10, 11]	0.6	1.5	μS	

<sup>(1)</sup> For subgroup definitions, please see Table 1.

Copyright © 2017, Texas Instruments Incorporated

<sup>(2)</sup> All typical values are at 25°C and with a 3.3-V supply.



# 7.8 Receiver Switching Characteristics

The specifications shown below are valid across temperature range of –55°C to 125°C pre-radiation and 25°C post-radiation. When different, the post-radiation values are shown in a separate row specified by the corresponding RHA level (L = 50 krad).

	PARAMETER	TEST CONDITIONS	SUBGROUP <sup>(1)</sup>	MIN TYP <sup>(2)</sup>	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output		[9, 10, 11]	35	105	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output	h-to-low-level See Figure 17		35	105	ns
t <sub>sk(p)</sub>	Pulse skew ( t <sub>PHL</sub> - t <sub>PLH</sub>  )	Coo rigaro ri		7		ns
t <sub>r</sub>	Output signal rise time			2		ns
t <sub>f</sub>	Output signal fall time			2		ns

<sup>(1)</sup> For subgroup definitions, please see Table 1.

# 7.9 Device Switching Characteristics

The specifications shown below are valid across temperature range of  $-55^{\circ}$ C to 125°C pre-radiation and 25°C post-radiation. When different, the post-radiation values are shown in a separate row specified by the corresponding RHA level (L = 50 krad).

	PARAMETER	TEST CONDITIONS	SUBGROUP <sup>(1)</sup>	MIN TYP <sup>(2)</sup>	MAX	UNIT
t <sub>(LBK)</sub>	Loopback delay, driver input to receiver output	See Figure 22		7.5		ns
		V <sub>(RS)</sub> at 0 V, see Figure 21	[9, 10, 11]	70 150		
	Total loop delay, driver input to receiver output, recessive to dominant	$V_{(RS)}$ with 10 k $\Omega$ to ground, see Figure 21	[9, 10, 11]	105	225	ns
		$V_{(RS)}$ with 50 k $\Omega$ to ground, see Figure 21	[9, 10, 11]	500	600	
		V <sub>(RS)</sub> at 0 V, See Figure 21	[9, 10, 11]	70	150	
t <sub>(loop2)</sub>	Total loop delay, driver input to receiver output, dominant to recessive	$V_{(RS)}$ with 10 k $\Omega$ to ground, see Figure 21	[9, 10, 11]	105	225	ns
		$V_{(RS)}$ with 50 k $\Omega$ to ground, see Figure 21	[9, 10, 11]	500	600	

<sup>(1)</sup> For subgroup definitions, please see Table 1.

Table 1. Quality Conformance Inspection<sup>(1)</sup>

SUBGROUP	DESCRIPTION	TEMPERATURE (°C)
1	Static tests at	25
2	Static tests at	125
3	Static tests at	-55
4	Dynamic tests at	25
5	Dynamic tests at	125
6	Dynamic tests at	-55
7	Functional tests at	25
8A	Functional tests at	125
8B	Functional tests at	-55
9	Switching tests at	25
10	Switching tests at	125
11	Switching tests at	<b>-</b> 55

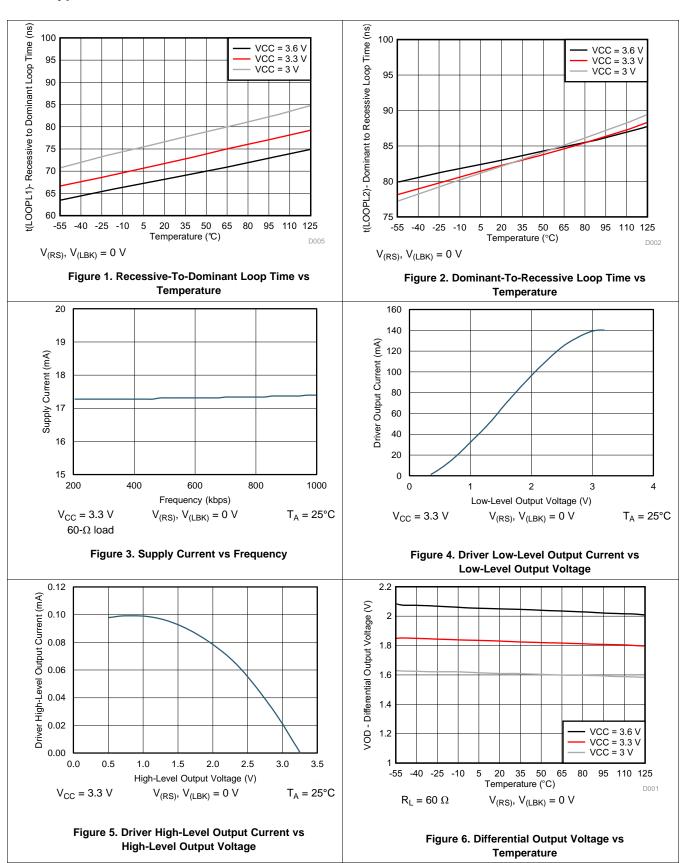
<sup>(1)</sup> MIL-STD-883, Method 5005 - Group A

<sup>(2)</sup> All typical values are at 25°C and with a 3.3-V supply.

<sup>(2)</sup> All typical values are at 25°C and with a 3.3-V supply.



# 7.10 Typical Characteristics



# TEXAS INSTRUMENTS

# **Typical Characteristics (continued)**

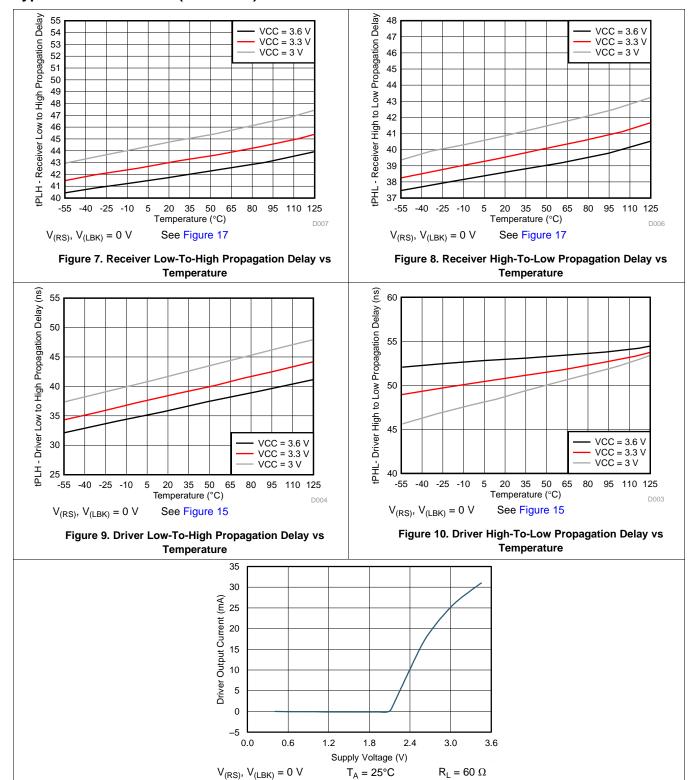


Figure 11. Driver Output Current vs Supply Voltage



# 8 Parameter Measurement Information

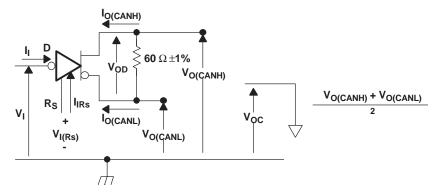


Figure 12. Driver Voltage, Current, and Test Definition

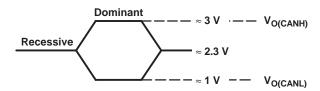


Figure 13. Bus Logic State Voltage Definitions

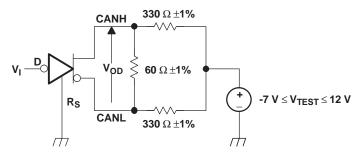
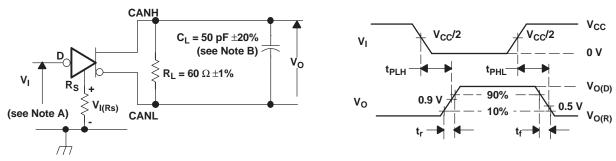


Figure 14. Driver V<sub>OD</sub>



- A. The input pulse is supplied by a generator having the following characteristics:
  - Pulse repetition rate (PRR) ≤125 kHz, 50% duty cycle
  - t<sub>r</sub> ≤ 6 ns
  - t<sub>f</sub> ≤ 6 ns
  - $Z_O = 50 \Omega$
- B.  $C_L$  includes fixture and instrumentation capacitance.

Figure 15. Driver Test Circuit and Voltage Waveforms

Copyright © 2017, Texas Instruments Incorporated



# **Parameter Measurement Information (continued)**

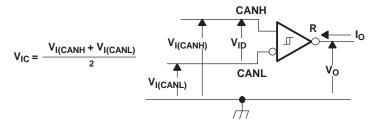
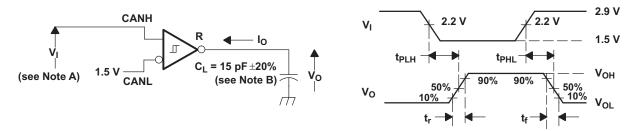


Figure 16. Receiver Voltage and Current Definitions

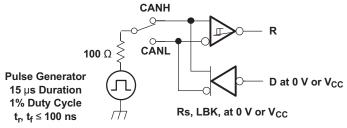


- A. The input pulse is supplied by a generator having the following characteristics:
  - PRR ≤125 kHz, 50% duty cycle
  - t<sub>r</sub> ≤ 6 ns
  - t<sub>f</sub> ≤ 6 ns
  - $Z_{\rm O} = 50~\Omega$
- B. C<sub>L</sub> includes fixture and instrumentation capacitance.

Figure 17. Receiver Test Circuit and Voltage Waveforms

**INPUT** OUTPUT **MEASURED** R **V<sub>CANH</sub>** VCANL |V<sub>ID</sub>| -7 V -6.1 V L 900 mV 12 V 11.1 V L 900 mV  $V_{OL}$ -1 V –7 V 6 V 6 V 12 V 6 V L -7 V Н -6.5 V 500 mV 11.5 V Н 500 mV 12 V 6 V -7 V -1 V Н  $V_{OH}$ 6 V 12 V Н 6 V Χ Open Open Н

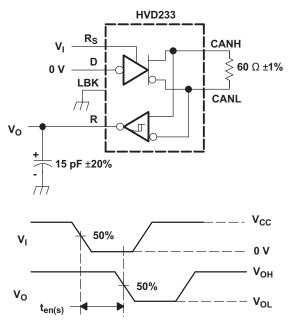
Table 2. Differential Input Voltage Threshold Test



NOTE: This test is conducted to test survivability only. Data stability at the R output is not specified.

Figure 18. Test Circuit, Transient Overvoltage Test



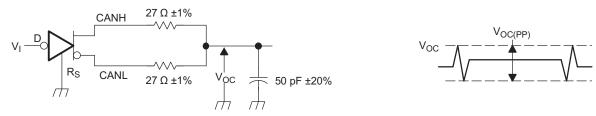


Copyright © 2017, Texas Instruments Incorporated

NOTE: All V<sub>I</sub> input pulses are supplied by a generator having the following characteristics:

- t<sub>r</sub> or t<sub>f</sub> ≤ 6 ns
- PRR = 125 kHz, 50% duty cycle

Figure 19.  $T_{en(s)}$  Test Circuit and Voltage Waveforms



NOTE: All V<sub>I</sub> input pulses are supplied by a generator having the following characteristics:

- $t_r$  or  $t_f \le 6$  ns
- PRR = 125 kHz, 50% duty cycle

Figure 20. V<sub>OC(pp)</sub> Test Circuit and Voltage Waveforms

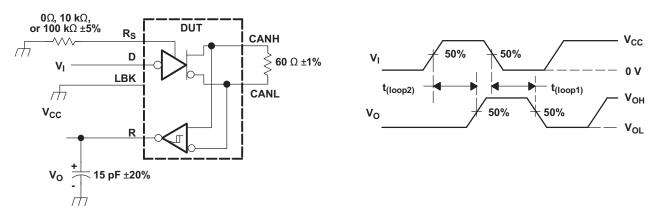
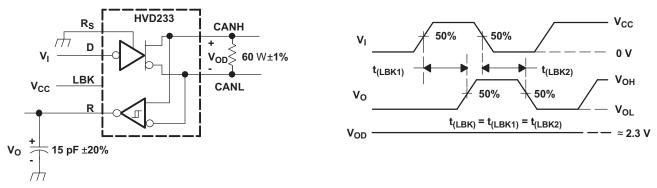


Figure 21. T<sub>(loop)</sub> Test Circuit and Voltage Waveforms





Copyright © 2017, Texas Instruments Incorporated

Figure 22. T<sub>(LBK)</sub> Test Circuit and Voltage Waveforms

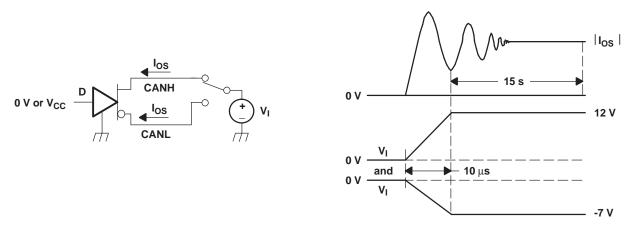
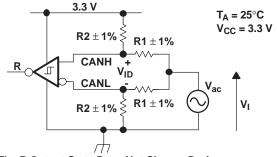


Figure 23. I<sub>OS</sub> Test Circuit and Waveforms



The R Output State Does Not Change During Application of the Input Waveform.

		V <sub>ID</sub>	R1	R2		
		500 mV	50 Ω	<b>280</b> Ω		
		900 mV	50 Ω	130 Ω	ľ	
Vı	$\cap$		$\bigvee$			12 V -7 V

NOTE: All input pulses are supplied by a generator with  $f \le 1.5$  MHz.

Figure 24. Common-Mode Voltage Rejection



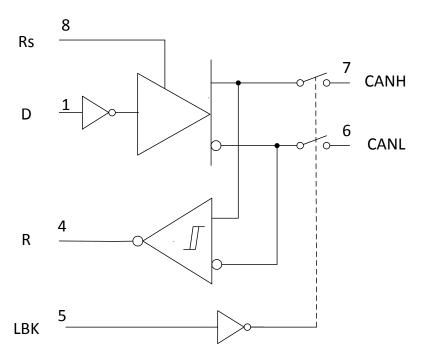
# 9 Detailed Description

#### 9.1 Overview

The SN55HVD233-SP is used in applications employing the CAN serial communication physical layer in accordance with the ISO 11898 standard. As a CAN transceiver, the device provides transmit and receive capability between the differential CAN bus and a CAN controller, with signaling rates up to 1 Mbps.

Designed for operation in especially harsh environments, the SN55HVD233-SP features cross-wire, overvoltage, and loss of ground protection to ±16 V, overtemperature (thermal shutdown) protection, and common-mode transient protection of ±100 V. This device operates over a wide –7-V to 12-V common mode range. This transceiver is the interface between the host CAN controller on the microprocessor, FPGA, or ASIC, and the differential CAN bus used in satellite applications.

#### 9.2 Functional Block Diagram



Copyright © 2017, Texas Instruments Incorporated

## 9.3 Feature Description

#### 9.3.1 Modes

The  $R_s$ , pin 8 of the SN55HVD233-SP, provides for three modes of operation: high-speed, slope control, or low-power standby mode. The user selects the high-speed mode of operation by connecting pin 8 directly to ground, allowing the driver output transistors to switch on and off as fast as possible with no limitation on the rise and fall slope. The user can adjust the rise and fall slope by connecting a resistor to ground at pin 8, because the slope is proportional to the pin's output current. Slope control is implemented with a resistor values of 0  $\Omega$  to achieve a single ended slew rate of approximately 38 V/ $\mu$ s up to a value of 50 k $\Omega$  to achieve approximately 4 V/ $\mu$ s slew rate. For more information about slope control, refer to *Application and Implementation*.

The SN55HVD233-SP enters a low-current standby (listen-only) mode during which the driver is switched off and the receiver remains active if a high logic level is applied to pin 8. The local protocol controller reverses this low-current standby mode when it needs to transmit to the bus.

Copyright © 2017, Texas Instruments Incorporated



# 9.3.2 Loopback

A logic high on the loopback LBK pin 5 of the SN55HVD233-SP places the bus output and bus input in a high-impedance state. The remaining circuit remains active and available for driver-to-receiver loopback, self-diagnostic node functions without disturbing the bus. For more information on the loopback mode, refer to the *Application Information*.



#### 9.3.3 CAN Bus States

The CAN bus has two states during powered operation of the device: dominant and recessive. A dominant bus state is when the bus is driven differentially, corresponding to a logic low on the D and R pin. A recessive bus state is when the bus is biased to V<sub>CC</sub> / 2 through the high-resistance internal input resistors R<sub>IN</sub> of the receiver, corresponding to a logic high on the D and R pins (see Figure 25 and Figure 26).

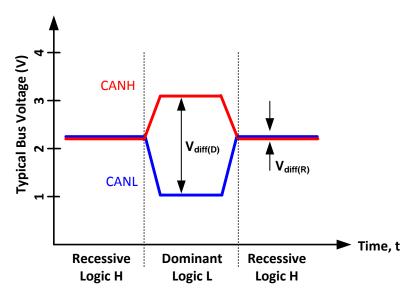


Figure 25. Bus States (Physical Bit Representation)

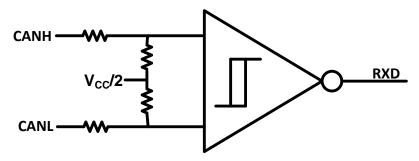


Figure 26. Simplified Recessive Common Mode Bias and Receiver

#### 9.3.4 ISO 11898 Compliance of SN55HVD233-SP

#### 9.3.4.1 Introduction

Many users value the low-power consumption of operating their CAN transceivers from a 3.3-V supply. However, some users are concerned about the interoperability with 5-V supplied transceivers on the same bus. This report analyzes this situation to address those concerns.

## 9.3.4.2 Differential Signal

CAN is a differential bus where complementary signals are sent over two wires and the voltage difference between the two wires defines the logical state of the bus. The differential CAN receiver monitors this voltage difference and outputs the bus state with a single-ended output signal.

Copyright © 2017, Texas Instruments Incorporated Submit Documentation Feedback



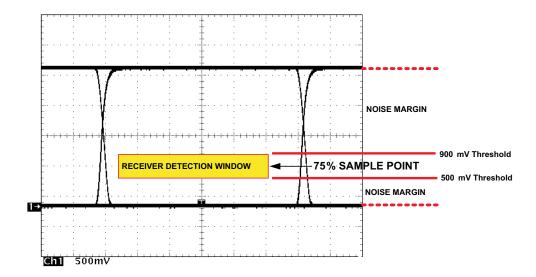


Figure 27. Typical SN55HVD233-SP Differential Output Voltage Waveform

The CAN driver creates the difference in voltage between CANH and CANL in the dominant state. The dominant differential output of the SN55HVD233-SP is greater than 1.5 V and less than 3 V across a  $60-\Omega$  load. The minimum required by ISO 11898 is 1.5 V and maximum is 3 V. These are the same limiting values for 5-V supplied CAN transceivers. The bus termination resistors drive the recessive bus state and not the CAN driver.

A CAN receiver is required to output a recessive state with less than 500 mV and a dominant state with more than 900 mV difference voltage on its bus inputs. The CAN receiver must do this with common-mode input voltages from –2 V to 7 V. The SN55HVD233-SP receiver meets these same input specifications as 5-V supplied receivers.

#### 9.3.4.2.1 Common-Mode Signal

A common-mode signal is an average voltage of the two signal wires that the differential receiver rejects. The common-mode signal comes from the CAN driver, ground noise, and coupled bus noise. The supply voltage of the CAN transceiver has nothing to do with noise. The SN55HVD233-SP driver lowers the common-mode output in a dominant bit by a couple hundred millivolts from that of most 5-V drivers. While this does not fully comply with ISO 11898, this small variation in the driver common-mode output is rejected by differential receivers and does not effect data, signal noise margins, or error rates.

#### 9.3.4.3 Interoperability of 3.3-V CAN in 5-V CAN Systems

The 3.3-V supplied CAN transceivers are electrically interchangeable with 5-V CAN transceivers. The differential output is the same. The recessive common mode output is the same. The dominant common mode output voltage is a couple hundred millivolts lower than 5 V supplied drivers, while the receivers exhibit identical specifications as 5-V devices.

To help ensure the widest interoperability possible, the SN55HVD233-SP successfully passed the internationally recognized GIFT ICT conformance and interoperability testing for CAN transceivers. Electrical interoperability does not always assure interchangeability, however. Most implementers of CAN buses recognize that ISO 11898 does not sufficiently specify the electrical layer and that strict standard compliance alone does not ensure full interchangeability. Interchangeability is ensured with thorough equipment testing.



#### 9.3.5 Thermal Shutdown

If the junction temperature of the device exceeds the thermal shutdown threshold, the device turns off the CAN driver circuits thus blocking the D pin to bus transmission path. The shutdown condition is cleared when the junction temperature drops below the thermal shutdown temperature of the device. The CAN bus pins are high-impedance biased to recessive level during a thermal shutdown, and the receiver-to-R pin path remains operational.

# 9.4 Device Functional Modes

Table 3. Driver I/O

DRIVER <sup>(1)</sup>									
	INPUTS		OUTPUTS						
D	LBK	RS	CANH	CANL	BUS STATE				
X	X	> 0.75 V <sub>CC</sub>	Z	Z	Recessive				
L	L or open	< 0.22.1/	Н	L	Dominant				
H or open	X	≤ 0.33 V <sub>CC</sub>	Z	Z	Recessive				
X	Н	≤ 0.33 V <sub>CC</sub>	Z	Z	Recessive				

(1) H = High level; L = Low level; Z = High impedance; X = Irrelevant

Table 4. Receiver I/O

RECEIVER <sup>(1)</sup>							
	OUTPUT						
BUS STATE	$V_{ID} = V_{(CANH)} - V_{(CANL)}$	D	R				
Dominant	V <sub>ID</sub> ≥ 0.9 V	X	L				
Recessive	V <sub>ID</sub> ≤ 0.5 V or open	H or open	Н				
?	$0.5 \text{ V} < \text{V}_{\text{ID}} < 0.9 \text{ V}$	H or open	?				
Dominant	V <sub>ID</sub> ≥ 0.9 V	Χ	L				
Recessive	V <sub>ID</sub> ≤ 0.5 V or open	Н	Н				
Recessive	V <sub>ID</sub> ≤ 0.5 V or open	L	L				
?	$0.5 \text{ V} < \text{V}_{\text{ID}} < 0.9 \text{ V}$	L	L				

(1) H = High level; L = Low level; Z = High impedance; X = Irrelevant; ? = Indeterminate

Copyright © 2017, Texas Instruments Incorporated



# 10 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

#### **10.1** Application Information

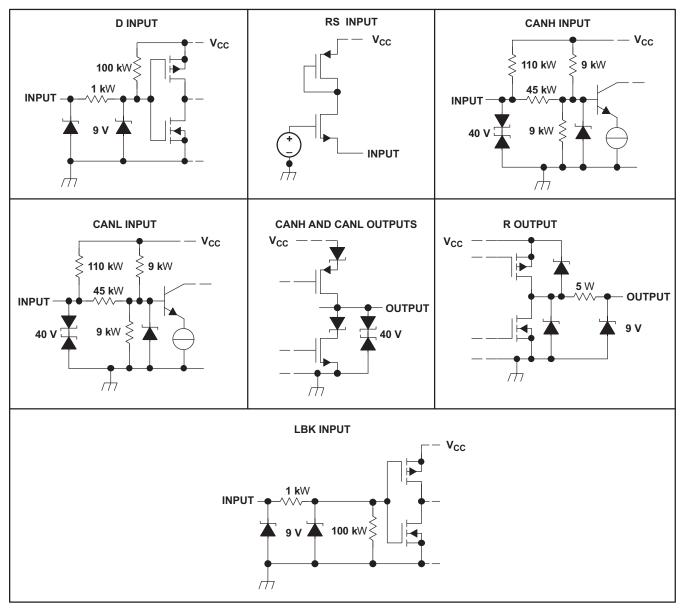
# 10.1.1 Diagnostic Loopback

The diagnostic loopback or internal loopback function of the SN55HVD233-SP is enabled with a high-level input on pin 7, LBK. This mode disables the driver output while keeping the bus pins biased to the recessive state. This mode also redirects the D data input (transmit data) through logic to the received data output (R), thus creating an internal loopback of the transmit-to-receive data path. This mimics the loopback that occurs normally with a CAN transceiver because the receiver loops back the driven output to the R (receive data) pin. This mode allows the host microprocessor to input and read back a bit sequence or CAN messages to perform diagnostic routines without disturbing the CAN bus. Figure 33 shows a typical CAN bus application.

If the LBK pin is not used, it may be tied to ground (GND). However, it is pulled low internally (defaults to a low-level input) and may be left open if not in use.



# **Application Information (continued)**

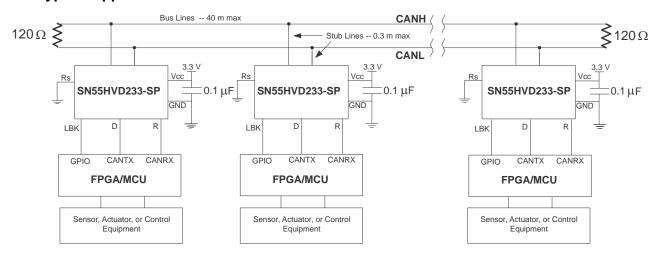


Copyright © 2017, Texas Instruments Incorporated

Figure 28. Equivalent Input and Output Schematic Diagrams



#### 10.2 Typical Application



Copyright © 2017, Texas Instruments Incorporated

Figure 29. Typical Application Schematic

#### 10.2.1 Design Requirements

The High-Speed ISO 11898 Standard specifications are given for a maximum signaling rate of 1 Mbps with a bus length of 40 m and a maximum of 30 nodes. It also recommends a maximum unterminated stub length of 0.3 m. The cable is specified to be a shielded or unshielded twisted-pair with a 120- $\Omega$  characteristic impedance (ZO). The standard defines a single line of twisted-pair cable with the network topology as shown in Figure 29. It is terminated at both ends with 120- $\Omega$  resistors, which match the characteristic impedance of the line to prevent signal reflections. According to ISO 11898, placing RL on a node should be avoided because the bus lines lose termination if the node is disconnected from the bus.

#### 10.2.2 Detailed Design Procedure

BUS LENGTH (m)

40

1

100

0.5

200

0.25

500

0.1

1000

0.05

Table 5. Suggested Cable Length vs Signaling Rate

Basically, the maximum bus length is determined by, or rather is a trade-off with the selected signaling rate as listed in Table 5.

A signaling rate decreases as transmission distance increases. While steady-state losses may become a factor at the longest transmission distances, the major factors limiting signaling rate as distance is increased are time varying. Cable bandwidth limitations, which degrade the signal transition time and introduce inter-symbol interference (ISI), are primary factors reducing the achievable signaling rate when transmission distance is increased.

For a CAN bus, the signaling rate is also determined from the total system delay – down and back between the two most distant nodes of a system and the sum of the delays into and out of the nodes on a bus with the typical 5 ns/m prop delay of a twisted-pair cable. Also, consideration must be given the signal amplitude loss due to resistance of the cable and the input resistance of the transceivers. Under strict analysis, skin effects, proximity to other circuitry, dielectric loss, and radiation loss effects all act to influence the primary line parameters and degrade the signal.

A conservative rule of thumb for bus lengths over 100 m is derived from the product of the signaling rate in Mbps and the bus length in meters, which should be less than or equal to 50.



Signaling Rate (Mbps)  $\times$  Bus Length (m)  $\leq$  50. Operation at extreme temperatures should employ additional conservatism.

#### 10.2.2.1 Slope Control

Adjust the rise and fall slope of the SN55HVD233-SP driver output by connecting a resistor from the RS (pin 8) to ground (GND), or to a low-level input voltage as shown in Figure 30.

The slope of the driver output signal is proportional to the pin's output current. This slope control is implemented with an external resistor value ranging from 0  $\Omega$  to achieve a  $\approx$ 38 V/ $\mu$ s single ended slew rate, and up to 50 k $\Omega$  to achieve a  $\approx$ 4 V/ $\mu$ s slew rate as displayed in Figure 31. Figure 32 shows typical driver output waveforms with slope control.

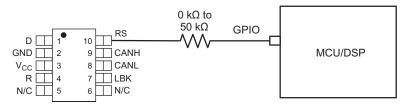
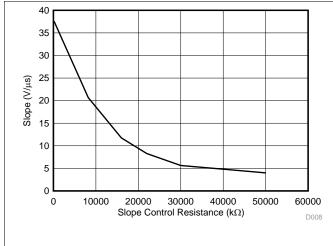


Figure 30. Slope Control/Standby Connection to a DSP

#### 10.2.2.2 Standby

If a high-level input (>  $0.75~V_{CC}$ ) is applied to RS (pin 8), the circuit enters a low-current, listen-only standby mode during which the driver is switched off and the receiver remains active. The local controller can reverse this low-power standby mode when the rising edge of a dominant state (bus differential voltage > 900-mV typical) occurs on the bus.

# 10.2.3 Application Curves





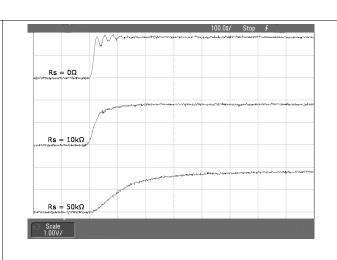


Figure 32. Typical SN55HVD233-SP 250-Kbps Output Pulse Waveforms With Slope Control



# 11 Power Supply Recommendations

TI recommends to have localized capacitive decoupling near device VCC pin to GND. Values of 4.7  $\mu$ F at VCC pin and 10  $\mu$ F, 1  $\mu$ F, and 0.1  $\mu$ F at supply have tested well on evaluation modules.

## 12 Layout

## 12.1 Layout Guidelines

Minimize stub length from node insertion to bus.

#### 12.1.1 Bus Loading, Length, and Number of Nodes

The ISO11898 standard specifies up to 1-Mbps data rate, maximum bus length of 40 m, maximum drop line (stub) length of 0.3 m, and a maximum of 30 nodes. However, with careful network design, the system may have longer cables, longer stub lengths, and many more nodes to a bus. Many CAN organizations and standards have scaled the use of CAN for applications outside the original ISO11898 standard. They made system level trade-offs for data rate, cable length, and parasitic loading of the bus. Examples of some of these specifications are ARINC825, CANopen, CAN Kingdom, DeviceNet, and NMEA200.

A high number of nodes requires a transceiver with high input impedance and wide common mode range such as the SN55HVD233-SP CAN. ISO11898-2 specifies the driver differential output with a  $60-\Omega$  load (two  $120-\Omega$  termination resistors in parallel), and the differential output must be greater than 1.5 V. The SN55HVD233-SP is specified to meet the 1.5-V requirement with a  $60-\Omega$  load, and additionally specified with a differential output voltage minimum of 1.2 V across a common mode range of -2 to 7 V through a  $330-\Omega$  coupling network. This network represents the bus loading of 120 SN55HVD233-SP transceivers based on their minimum differential input resistance of 40 k $\Omega$ . Therefore, the SN55HVD233-SP supports up to 120 transceivers on a single bus segment with margin to the 1.2-V minimum differential input voltage requirement at each node. For CAN network design, margin must be given for signal loss across the system and cabling, parasitic loadings, network imbalances, ground offsets, and signal integrity; thus, a practical maximum number of nodes may be lower. Bus length may also be extended beyond the original ISO11898 standard of 40 m by careful system design and data rate tradeoffs. For example, CANopen network design guidelines allow the network to be up to 1 km with changes in the termination resistance, cabling, less than 64 nodes, and significantly lowered data rate.

This flexibility in CAN network design is one of the key strengths of the various extensions and additional standards that have been built on the original ISO11898 CAN standard. Using this flexibility requires good network design.

#### 12.1.2 CAN Termination

The ISO11898 standard specifies the interconnect to be a twisted pair cable (shielded or unshielded) with  $120-\Omega$  characteristic impedance ( $Z_0$ ). Use resistors equal to the characteristic impedance of the line to terminate both ends of the cable to prevent signal reflections. Keep unterminated drop lines (stubs) connecting nodes to the bus as short as possible to minimize signal reflections. The termination may be on the cable or in a node, but if nodes may be removed from the bus, the termination must be carefully placed so that it is not removed from the bus.

Product Folder Links: SN55HVD233-SP



# **Layout Guidelines (continued)**

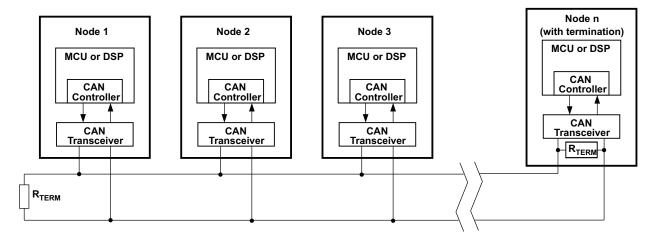


Figure 33. Typical CAN Bus

Termination is typically a  $120-\Omega$  resistor at each end of the bus. If filtering and stabilization of the common mode voltage of the bus is desired, then the user may use split termination (see Figure 34). Split termination uses two  $60-\Omega$  resistors with a capacitor in the middle of these resistors to ground. Split termination improves the electromagnetic emissions behavior of the network by eliminating fluctuations in the bus common mode voltages at the start and end of message transmissions.

Take care with the power ratings of the termination resistors used, especially for the worst-case condition (if a system power supply is shorted across the termination resistance to ground). In most cases, under the worst-case condition, much higher current passes through the termination resistance than the CAN transceiver's current limit.

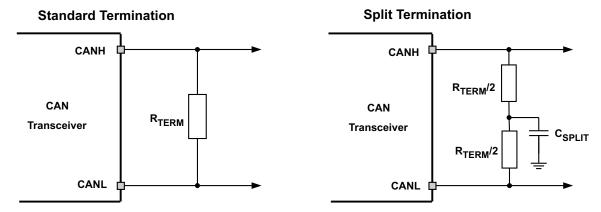


Figure 34. CAN Bus Termination Concepts

Copyright © 2017, Texas Instruments Incorporated



# 12.2 Layout Example

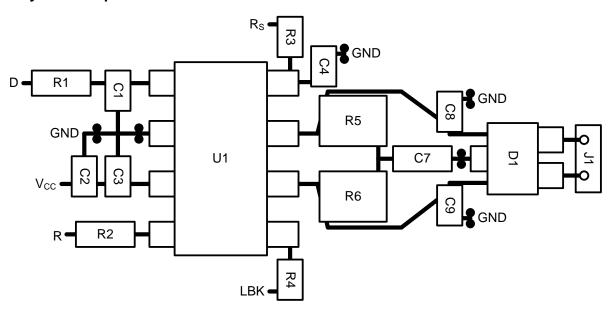


Figure 35. Board Layout Example



# 13 Device and Documentation Support

#### 13.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 13.2 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 13.3 Trademarks

E2E is a trademark of Texas Instruments.

All other trademarks are the property of their respective owners.

#### 13.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

#### 13.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

Copyright © 2017, Texas Instruments Incorporated

www.ti.com 10-Nov-2025

#### PACKAGING INFORMATION

Orderable part number	Status (1)	Material type	Package   Pins	Package qty   Carrier	<b>RoHS</b> (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
5962L1420901VXC	Active	Production	CFP (HKX)   8	25   TUBE	Yes	NIAU	N/A for Pkg Type	-55 to 125	L1420901VXC HVD233-SP
5962L1420901VXC.A	Active	Production	CFP (HKX)   8	25   TUBE	Yes	NIAU	N/A for Pkg Type	-55 to 125	L1420901VXC HVD233-SP
HVD233HKX/EM	Active	Production	CFP (HKX)   8	25   TUBE	Yes	NIAU	N/A for Pkg Type	25 to 25	HVD233HKX/EM EVAL ONLY
HVD233HKX/EM.A	Active	Production	CFP (HKX)   8	25   TUBE	Yes	NIAU	N/A for Pkg Type	25 to 25	HVD233HKX/EM EVAL ONLY

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

# PACKAGE OPTION ADDENDUM

www.ti.com 10-Nov-2025

# OTHER QUALIFIED VERSIONS OF SN55HVD233-SP:

◆ Catalog : SN55HVD233-SEP

NOTE: Qualified Version Definitions:

Catalog - TI's standard catalog product

# **PACKAGE MATERIALS INFORMATION**

www.ti.com 23-May-2025

# **TUBE**

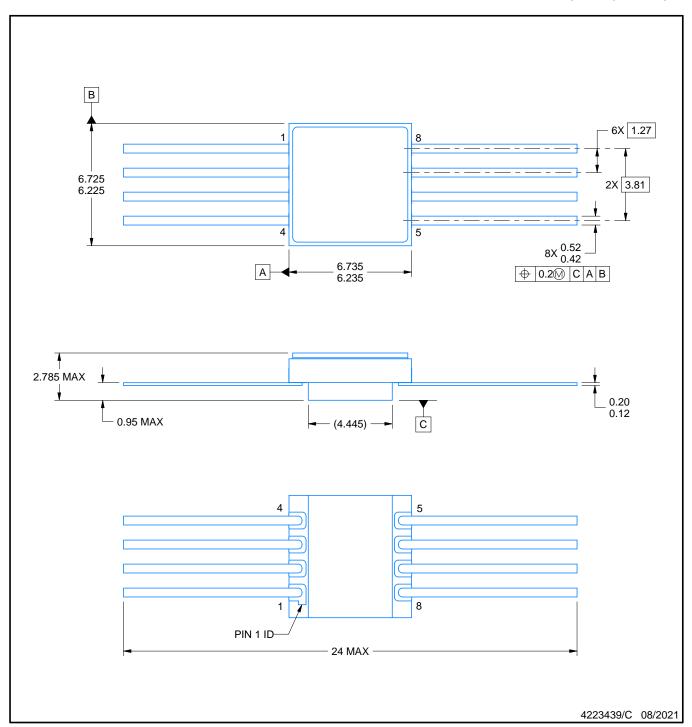


#### \*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
5962L1420901VXC	HKX	CFP (HSL)	8	25	506.98	26.16	6220	NA
5962L1420901VXC.A	HKX	CFP (HSL)	8	25	506.98	26.16	6220	NA
HVD233HKX/EM	HKX	CFP (HSL)	8	25	506.98	26.16	6220	NA
HVD233HKX/EM.A	HKX	CFP (HSL)	8	25	506.98	26.16	6220	NA



CERAMIC FLATPACK



#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  This drawing is subject to change without notice.

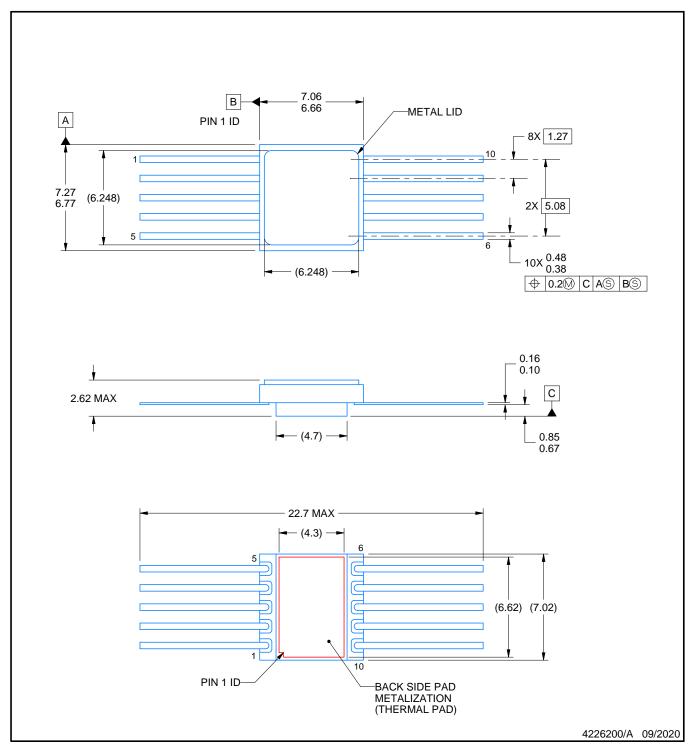
  his package is hermetically sealed with a metal lid.

  his package is plated.





CERAMIC DUAL FLATPACK

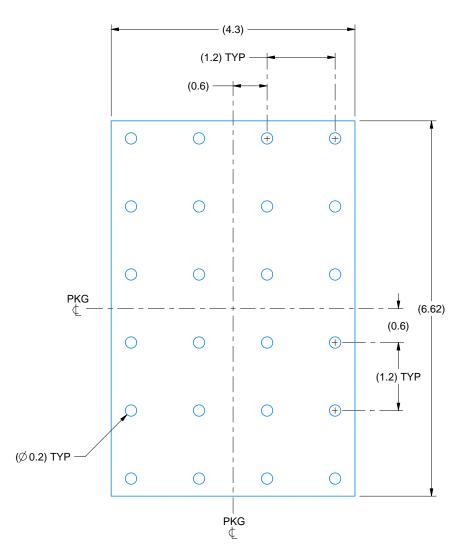


#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- This drawing is subject to change without notice.
   This package is hermetically sealed with a metal lid.
- 4. The terminals are gold plated.
- 5. This drawing does not comply with MIL STD 1835. Do not use this package for compliant product.6. Metal lid is connected to back side pad metalization.



CERAMIC DUAL FLATPACK



HEATSINK LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:15X

#### IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you fully indemnify TI and its representatives against any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to TI's Terms of Sale, TI's General Quality Guidelines, or other applicable terms available either on ti.com or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products. Unless TI explicitly designates a product as custom or customer-specified, TI products are standard, catalog, general purpose devices.

TI objects to and rejects any additional or different terms you may propose.

Copyright © 2025, Texas Instruments Incorporated

Last updated 10/2025