





LP5910 SNVSA91F - SEPTEMBER 2015 - REVISED APRIL 2021

# LP5910 300-mA Low-Noise, Low-Io LDO

#### 1 Features

Input voltage range: 1.3 V to 3.3 V Output voltage range: 0.8 V to 2.3 V

Output current: 300 mA PSRR: 75 dB at 1 kHz

Output voltage tolerance: ±2% Low dropout: 120 mV (typical)

Very low I<sub>O</sub> (enabled, no load): 12 μA

Low output-voltage noise: 12 μV<sub>RMS</sub>

Stable with ceramic input and output capacitors

Thermal overload protection

Short-circuit protection

Reverse current protection

Automatic output discharge for fast turnoff

## 2 Applications

- Mobile phones and tablets
- Digital cameras and audio devices
- · Portable and battery-powered equipment
- · Portable medical equipment
- Virtual reality
- RF, PLL, VCO, and clock power supplies
- IP cameras

## 3 Description

The LP5910 is a low-noise LDO that can supply up to 300 mA of output current. Designed to meet the requirements of RF and analog circuits, this device provides low noise, high PSRR, low quiescent current, and superior line transient and load transient response. Using new innovative design techniques the LP5910 offers class-leading noise performance without a noise bypass capacitor and with the option for remote output capacitor placement.

The device contains a reverse current protection circuit that prevents a reverse current flow through the LDO to the IN pin when the input voltage is lower than the output voltage.

When the Enable (EN) pin is low, and the output is in an OFF state, an automatic output discharge circuit discharges the output capacitance for fast turnoff.

With its low input and low output voltage range the LP5910 is well-suited as a post DC-DC regulator (post BUCK regulator) or for single- or dual-cell applications.

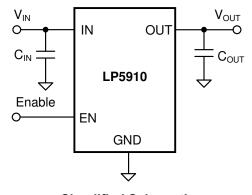
The device is designed to work with a 1-µF input and a 1-µF output ceramic capacitor. A separate noise bypass capacitor is not required.

This device is available with fixed output voltages from 0.8 V to 2.3 V in 25-mV steps. Contact Texas Instruments Sales for specific voltage option needs.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE		
LP5910	WSON (6)	2.00 mm × 2.00 mm (NOM)		
	DSBGA (4)	0.742 mm × 0.742 mm (MAX)		

For all available packages, see the orderable addendum at the end of the data sheet.



Simplified Schematic



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# **4 Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

С	hanges from Revision E (July 2017) to Revision F (April 2021)	Page
•	Updated the numbering format for tables, figures, and cross-references throughout the document	1
•	Deleted WEBENCH links from document	1
•	Deleted last bullet from Features section	1
•	Changed Dropout Voltage specifications for 1.5V ≤ V <sub>IN</sub> < 1.8V; added new rows in Dropout Voltage specifications for this voltage range	5
•	Deleted Custom Design With WEBENCH® Tools section from Detailed Design Procedure	
•	Deleted Custom Design With WEBENCH® Tools section from Documentation Support	19
С	hanges from Revision D (August 2016) to Revision E (July 2017)	Page
•	Added new package, YKA0004-C01 associated with orderables LP5910-1.1BYKAR and LP5910-1.1E added links for WEBENCH	3YKAT; 1

IN Α1

B1

ΕN

B2



# **5 Pin Configuration and Functions**

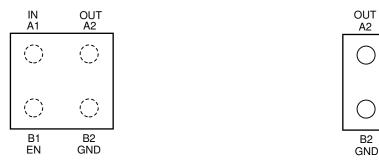


Figure 5-1. YKA Package, 4-Pin Ultra-Thin DSBGA, Figure 5-2. YKA Package, 4-Pin Ultra-Thin DSBGA, **Top View Bottom View** 

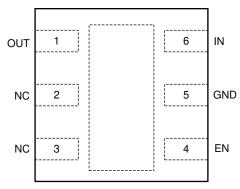


Figure 5-3. DRV Package, 6-Pin WSON With Thermal Pad, Top View

Table 5-1. Pin Functions

	PIN				
NAME	DSBGA	WSON	I/O	DESCRIPTION	
EN	B1	4	I	Enable input; disables the regulator when logic low. Enables the regulator when logic high. An internal 1-M $\Omega$ pull down resistor connects this input to ground.	
GND	B2	5	_	Common ground	
IN	A1	6	I	Voltage supply input. A 1-µF capacitor must be connected at this input.	
NC	_	2, 3	_	No internal connection. Connect to ground or leave open.	
OUT	A2	1	0	Voltage output. A 1-µF low-ESR capacitor must be connected from this pin to the GND pin. Connect this output to the load circuit.	
Exposed Pad	_	Thermal Pad	_	The exposed thermal pad on the bottom of the package must be connected to a copper area under the package on the PCB. Connect to ground potential or leave floating. Do not connect to any potential other than the same ground potential seen at device pin 5 (GND). See the <i>Power Dissipation</i> section for more information.	



## **6 Specifications**

## 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1) (2)

	MIN	MAX	UNIT
Input voltage, V <sub>IN</sub>	-0.3	3.6	V
Output voltage, V <sub>OUT</sub>	-0.3	3.6	V
Enable input voltage, V <sub>EN</sub>	-0.3	3.6	V
Continuous power dissipation <sup>(3)</sup>	Internally li	mited	W
Junction temperature, T <sub>J(MAX)</sub>		150	°C
Storage temperature, T <sub>stg</sub>	-65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltages are with respect to the GND pin.
- (3) Internal thermal shutdown circuitry protects the device from permanent damage.

## 6.2 ESD Ratings

			VALUE	UNIT
V	SD) Electrostatic discharg	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±1000	V
V <sub>(ES</sub>	SD) Liectrostatic discriary	Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±250	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

## **6.3 Recommended Operating Conditions**

over operating free-air temperature range (unless otherwise noted)(1)

	MIN	MAX	UNIT
Input voltage, V <sub>IN</sub>	1.3	3.3	V
Output voltage, V <sub>OUT</sub>	0.8	2.3	V
Enable input voltage, V <sub>EN</sub>	0	3.3	V
Output current, I <sub>OUT</sub>	0	300	mA
Junction temperature, T <sub>J</sub> <sup>(1)</sup>	-40	125	°C
Ambient temperature, T <sub>A</sub> <sup>(1)</sup>	-40	85	°C

(1) The maximum ambient temperature,  $(T_{A(MAX)})$  is a recommended value only and can vary depending on device power dissipation and  $R_{\theta JA}$ . For reliable operation, the junction temperature  $(T_J)$  must be limited to a maximum of 125°C. Ambient temperature  $(T_A)$ , thermal resistance  $(R_{\theta JA})$ ,  $V_{IN}$ ,  $V_{OUT}$ , and  $I_{OUT}$  all define  $T_J$ :  $T_J = T_A + (R_{\theta JA} \times ((V_{IN} - V_{OUT}) \times I_{OUT})$ .

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#### **6.4 Thermal Information**

		LP5	910		
	THERMAL METRIC <sup>(1)</sup>		DRV (WSON)	UNIT	
		4 PINS	6 PINS		
R <sub>0JA</sub> (2)	Junction-to-ambient thermal resistance, High-K	202.8	79.2 <sup>(3)</sup>	°C/W	
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	3.3	110.2	°C/W	
$R_{\theta JB}$	Junction-to-board thermal resistance	36.0	48.7	°C/W	
ΨЈТ	Junction-to-top characterization parameter	0.4	5.2	°C/W	
ΨЈВ	Junction-to-board characterization parameter	36.0	49.1	°C/W	
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	18.1	°C/W	

- (1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report.
- (2) Thermal resistance value R<sub>0JA</sub> is based on the EIA/JEDEC High-K printed circuit board defined by: *JESD51-7 High Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages*.
- (3) The PCB for the WSON/DRV package R<sub>θJA</sub> includes two (2) thermal vias under the exposed thermal pad per EIA/JEDEC JESD51-5.

## 6.5 Electrical Characteristics

 $V_{IN} = V_{OUT(NOM)} + 0.5 \text{ V}, V_{EN} = 1 \text{ V}, I_{OUT} = 1 \text{ mA}, C_{IN} = 1 \text{ } \mu\text{F}, \text{ and } C_{OUT} = 1 \text{ } \mu\text{F} \text{ (unless otherwise noted)}^{(1)} \text{ (2)} \text{ (3)}$ 

PARAMETER		TEST CONI	DITIONS	MIN	TYP	MAX	UNIT	
GENERA	L							
	Output voltage tolerance	$V_{IN} = (V_{OUT(NOM)} + 0.5 V)$ $I_{OUT} = 1 \text{ mA to } 300 \text{ mA}$	V <sub>IN</sub> = (V <sub>OUT(NOM)</sub> + 0.5 V) to 3.3 V, I <sub>OUT</sub> = 1 mA to 300 mA			2	%V <sub>OUT</sub>	
$\Delta V_{OUT}$	Line regulation	$V_{IN} = (V_{OUT(NOM)} + 0.5 V)$ $I_{OUT} = 1 \text{ mA}$	to 3.3 V,		0.01		%/V	
	Load regulation	I <sub>OUT</sub> = 1 mA to 300 mA			0.002		%/mA	
I <sub>LOAD</sub>	Load current	See <sup>(4)</sup>		0		300	mA	
1	Quiescent current <sup>(5)</sup>	V <sub>EN</sub> = 1 V, I <sub>OUT</sub> = 0 mA			12	25		
IQ	Quiescent current	V <sub>EN</sub> = 1 V, I <sub>OUT</sub> = 300 mA			230	350	μA	
I <sub>Q(SD)</sub>	Quiescent current in shutdown <sup>(5)</sup>	$V_{EN} = 0.3 \text{ V}, -40^{\circ}\text{C} \le T_{J} \le$	85°C		0.02	2		
	Output reverse current <sup>(7)</sup>	V <sub>OUT</sub> = 3.3 V, V <sub>IN</sub> = V <sub>EN</sub> =	V <sub>OUT</sub> = 3.3 V, V <sub>IN</sub> = V <sub>EN</sub> = 0 V			0	μA	
$I_{RO}$ $V_{OUT} > V_{IN}$		V <sub>OUT</sub> = 3.3 V, V <sub>IN</sub> = V <sub>EN</sub> = 1.3 V		0		50	μA	
I <sub>G</sub>	Ground current <sup>(6)</sup>	I <sub>OUT</sub> = 0 mA (V <sub>OUT</sub> = 2.3 \	/)		15		μA	
		$1.3 \text{ V} \le \text{V}_{\text{OUT}} < 1.5 \text{ V},$ $\text{I}_{\text{OUT}} = 300 \text{ mA}$	DSBGA only		200	300		
	5 (8)	$1.5 \text{ V} \le \text{V}_{\text{OUT}} < 1.8 \text{V},$ $\text{I}_{\text{OUT}} = 300 \text{ mA}$	DSBGA only		160	235		
V		$1.8 \text{ V} \le \text{V}_{\text{OUT}} \le 2.3 \text{ V},$ $\text{I}_{\text{OUT}} = 300 \text{ mA}$	DSBGA only		120	180	mV	
V <sub>DO</sub> Dropout vol	Dropout voltage <sup>(8)</sup>	$1.3 \text{ V} \le \text{V}_{\text{OUT}} < 1.5 \text{ V},$ $\text{I}_{\text{OUT}} = 300 \text{ mA}$	WSON only		245	370	IIIV	
		$1.5 \text{ V} \le \text{V}_{\text{OUT}} < 1.8 \text{V},$ $\text{I}_{\text{OUT}} = 300 \text{ mA}$	WSON only		195	270		
		$1.8 \text{ V} \le \text{V}_{\text{OUT}} \le 2.3 \text{ V},$ $\text{I}_{\text{OUT}} = 300 \text{ mA}$	WSON only		145	220		
I <sub>LIMIT</sub>	Output current limit	$V_{OUT} = V_{OUT(NOM)} - 0.1 \text{ V}$ $V_{IN} = V_{OUT(NOM)} + 0.5 \text{ V}$			450		mA	



## **6.5 Electrical Characteristics (continued)**

 $V_{IN} = V_{OUT(NOM)} + 0.5 \text{ V}, V_{EN} = 1 \text{ V}, I_{OUT} = 1 \text{ mA}, C_{IN} = 1 \text{ } \mu\text{F}, \text{ and } C_{OUT} = 1 \text{ } \mu\text{F} \text{ (unless otherwise noted)}$ 

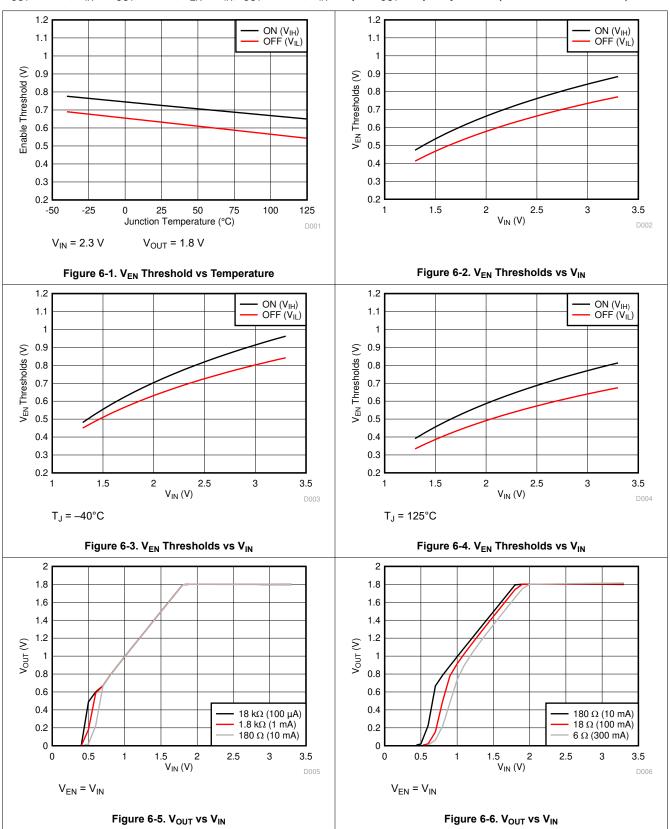
	PARAMETER	TEST CONDIT	TIONS	MIN	TYP	MAX	UNIT
		f = 100 Hz, I <sub>OUT</sub> = 20 mA, V	′ <sub>OUT</sub> ≥ 1 V		80		
		$f$ = 1 kHz, I <sub>OUT</sub> = 20 mA, V <sub>OUT</sub> $\ge$ 1 V			75		
		$f = 10 \text{ kHz}, I_{OUT} = 20 \text{ mA}, V_{OUT} \ge 1 \text{ V}$			65		
		f = 100 kHz, I <sub>OUT</sub> = 20 mA, V <sub>OUT</sub> ≥ 1 V			40		
DODD	D	$f = 2 \text{ MHz}, I_{OUT} = 20 \text{ mA}, V_{OUT}$	<sub>OUT</sub> ≥ 1 V		25		-ID
PSRR	Power supply rejection ratio <sup>(10)</sup>	f = 100 Hz, I <sub>OUT</sub> = 20 mA, 0	.8 V < V <sub>OUT</sub> < 1 V		65		dB
		$f = 1 \text{ kHz}, I_{OUT} = 20 \text{ mA}, 0.8$	3 V < V <sub>OUT</sub> < 1 V		65		
		$f = 10 \text{ kHz}, I_{OUT} = 20 \text{ mA}, 0$	.8 V < V <sub>OUT</sub> < 1 V		65		
		f = 100 kHz, I <sub>OUT</sub> = 20 mA,	0.8 V < V <sub>OUT</sub> < 1 V		40		
		$f = 2 \text{ MHz}, I_{\text{OUT}} = 20 \text{ mA}, 0.$	8 V < V <sub>OUT</sub> < 1 V		25		
_	Outrot main and the ma (10)	DW 4011- t- 4001-11-	I <sub>OUT</sub> = 1 mA		12		1 /
e <sub>N</sub>	Output noise voltage <sup>(10)</sup>	BW = 10 Hz to 100 kHz	I <sub>OUT</sub> = 300 mA		12		$\mu V_{RMS}$
_	Thermal shutdown	T <sub>J</sub> rising until output is OFF			160		80
$T_{SD}$	Thermal hysteresis	T <sub>J</sub> falling from shutdown	T <sub>J</sub> falling from shutdown		15		°C
LOGIC IN	IPUT THRESHOLDS		,				
V <sub>IL</sub>	EN low threshold (Off)	V <sub>IN</sub> = 1.3 V to 3.3 V				0.3	V
V <sub>IH</sub>	EN high threshold (On)			1			
	[N n in n man 1/9]	V <sub>EN</sub> = 3.3 V, V <sub>IN</sub> = 3.3 V			3.3		
I <sub>EN</sub>	EN pin current <sup>(9)</sup>	V <sub>EN</sub> = 0 V, V <sub>IN</sub> = 3.3 V			0.001		μA
TRANSIE	ENT CHARACTERISTICS(9)		•				
	Line transient <sup>(10)</sup>	$V_{IN}$ = ( $V_{OUT(NOM)}$ + 0.5 V) to in 30 µs $I_{OUT}$ = 1 mA	(V <sub>OUT(NOM)</sub> + 1 V)		0	1	m\/
$\Delta V_{OUT}$	Line transient	$V_{IN}$ = ( $V_{OUT(NOM)}$ + 1 V) to ( $V_{IN}$ in 30 $\mu$ s $I_{OUT}$ = 1 mA	/ <sub>OUT(NOM)</sub> + 0.5 V)	-1	0		mV
	Load transient <sup>(10)</sup>	I <sub>OUT</sub> = 1 mA to 100 mA in 10	) µs	-45			mV
	Load transient	I <sub>OUT</sub> = 100 mA to 1 mA in 10 μs				45	IIIV
	Overshoot on start-up <sup>(10)</sup>					5%	
t <sub>ON</sub>	Turnon time	From V <sub>EN</sub> > V <sub>IH</sub> to V <sub>OUT</sub> = 95	5% of V <sub>OUT(NOM)</sub>		80	200	μs
OUTPUT	DISCHARGE		-				
R <sub>AD</sub>	Output discharge pulldown resistance	V <sub>EN</sub> = 0 V, V <sub>IN</sub> = 2.3 V			160		Ω

- (1) All voltages are with respect to the device GND pin.
- (2) Minimum and maximum limits are ensured through test, design, or statistical correlation over the  $T_J$  range of  $-40^{\circ}$ C to 125°C, unless otherwise stated. Typical values represent the most likely parametric norm at  $T_A = 25^{\circ}$ C, and are provided for reference purposes only.
- (3) C<sub>IN</sub>, C<sub>OLIT</sub>: Low-ESR Surface-Mount-Ceramic Capacitors (MLCCs) used in setting electrical characteristics.
- (4) The device maintains a stable, regulated output voltage without a load current.
- (5) Quiescent current is defined here as the difference in current between the input voltage source and the load at V<sub>OUT</sub>. I<sub>Q</sub> = (I<sub>IN</sub> I<sub>OUT</sub>)
- (6) Ground current is defined here as the total current flowing to ground as a result of all input voltages applied to the device.
- (7) Output reverse current (I<sub>RO</sub>) is measured at the IN pin.
- (8) Dropout voltage is the voltage difference between the input and the output at which the output voltage drops to 100 mV below its nominal value. Dropout voltage is not a valid condition for output voltages less than 1.3 V as compliance with the minimum operating input voltage can not be ensured.
- (9) There is a 1-M $\Omega$  resistor between EN and ground on the device.
- (10) This specification is verified by design.

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## **6.6 Typical Characteristics**

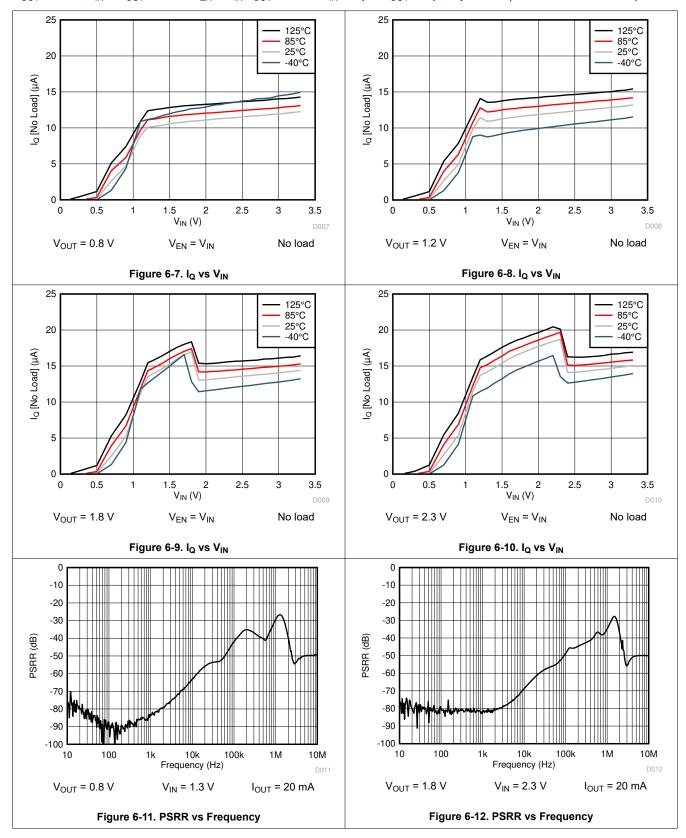
 $V_{OUT} = 1.8 \text{ V}, V_{IN} = V_{OUT} + 0.5 \text{ V}, V_{EN} = V_{IN}, I_{OUT} = 1 \text{ mA}, C_{IN} = 1 \text{ } \mu\text{F}, C_{OUT} = 1 \text{ } \mu\text{F}, T_{J} = 25 ^{\circ}\text{C} \text{ (unless otherwise noted)}$ 





## **6.6 Typical Characteristics (continued)**

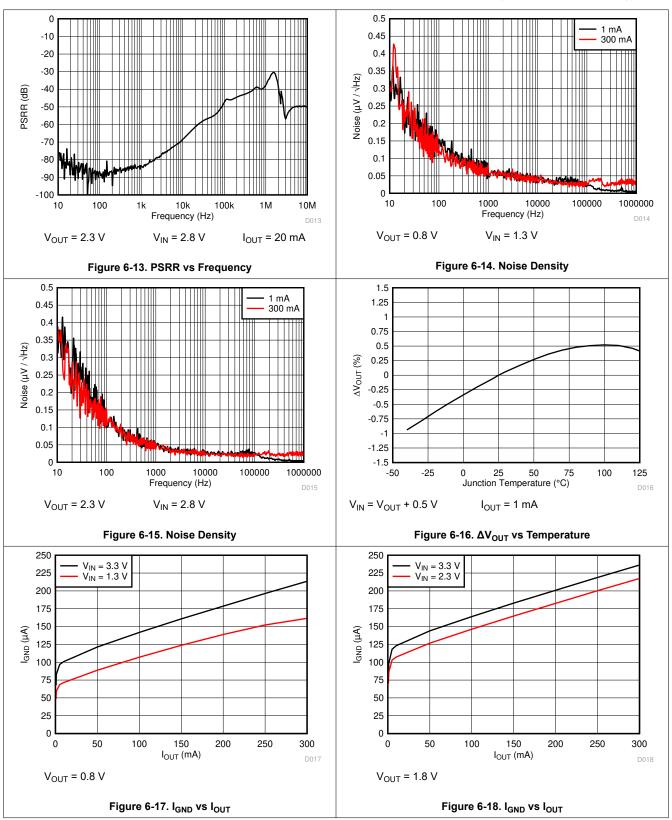
 $V_{OUT} = 1.8 \text{ V}, V_{IN} = V_{OUT} + 0.5 \text{ V}, V_{EN} = V_{IN}, I_{OUT} = 1 \text{ mA}, C_{IN} = 1 \text{ } \mu\text{F}, C_{OUT} = 1 \text{ } \mu\text{F}, T_{J} = 25 ^{\circ}\text{C} \text{ (unless otherwise noted)}$ 





## **6.6 Typical Characteristics (continued)**

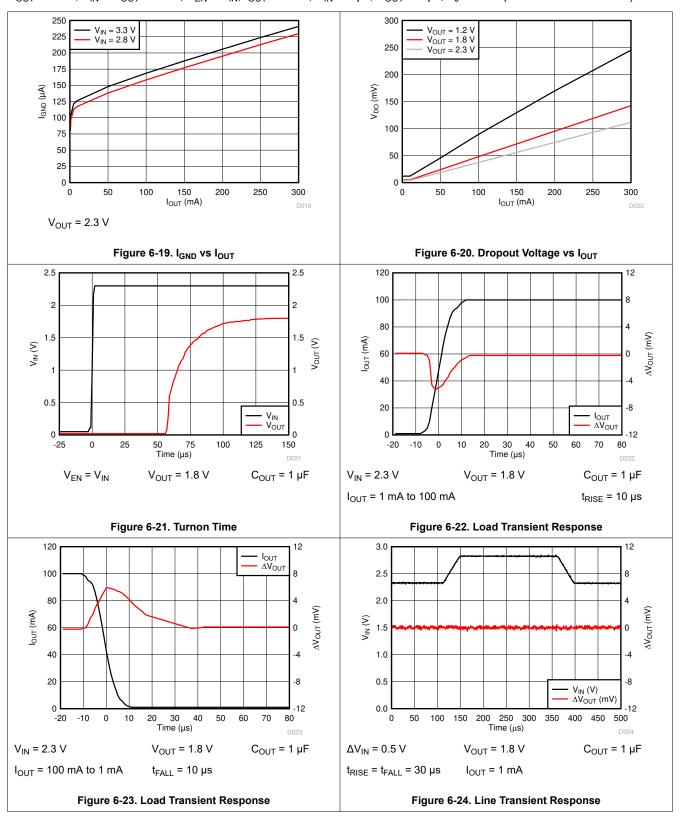
 $V_{OUT} = 1.8 \text{ V}, V_{IN} = V_{OUT} + 0.5 \text{ V}, V_{EN} = V_{IN}, I_{OUT} = 1 \text{ mA}, C_{IN} = 1 \text{ } \mu\text{F}, C_{OUT} = 1 \text{ } \mu\text{F}, T_{J} = 25 ^{\circ}\text{C} \text{ (unless otherwise noted)}$ 





## **6.6 Typical Characteristics (continued)**

 $V_{OUT}$  = 1.8 V,  $V_{IN}$  =  $V_{OUT}$  + 0.5 V,  $V_{EN}$  =  $V_{IN}$ ,  $I_{OUT}$  = 1 mA,  $C_{IN}$  = 1  $\mu$ F,  $C_{OUT}$  = 1  $\mu$ F,  $T_J$  = 25 °C (unless otherwise noted)



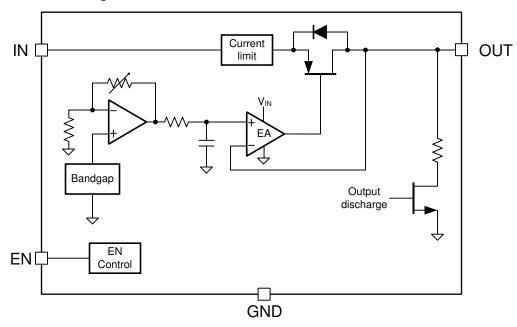


## 7 Detailed Description

#### 7.1 Overview

The LP5910 is a linear regulator capable of supplying 300-mA output current. Designed to meet the requirements of RF and analog circuits, the LP5910 device provides low noise, high PSRR, low quiescent current, and low line/load transient response figures. Using new innovative design techniques the LP5910 offers class-leading noise performance without a noise bypass capacitor and the option for remote output capacitor placement.

## 7.2 Functional Block Diagram



## 7.3 Feature Description

## 7.3.1 No-Load Stability

The LP5910 remains stable and in regulation with no external load.

### 7.3.2 Thermal Overload Protection

The LP5910 contains a thermal shutdown protection circuit to turn off the output current when excessive heat is dissipated in the LDO. Thermal shutdown occurs when the thermal junction temperature  $(T_J)$  of the main pass-FET exceeds 160°C (typical). Thermal shutdown hysteresis assures that the LDO again resets (turns on) when the temperature falls to 145°C (typical).

#### 7.3.3 Short-Circuit Protection

The LP5910 contains internal current limit which reduces output current to a safe value if the output is overloaded or shorted. Depending upon the value of  $V_{IN}$ , thermal limiting may also become active as the average power dissipated causes the die temperature to increase to the limit value (about 160°C). The hysteresis of the thermal shutdown circuitry can result in a *cyclic* behavior on the output as the die temperature heats and cools.

#### 7.3.4 Output Automatic Discharge

The LP5910 output employs an internal  $160-\Omega$  (typical) pulldown resistance to discharge the output when the EN pin is low, and the device is disabled.

### 7.3.5 Reverse Current Protection

The device contains a reverse current protection circuit that prevents a backward current flowing through the LDO from the OUT pin to the IN pin.



## 7.4 Device Functional Modes

### 7.4.1 Enable (EN)

The LP5910 may be switched to the ON or OFF state by logic input at the EN pin. A logic-high voltage on the EN pin turns the device to the ON state. A logic-low voltage on the EN pin turns the device to the OFF state. If the application does not require the shutdown feature, the EN pin must be tied to VIN to keep the regulator output permanently in the ON state when power is applied

To ensure proper operation, the signal source used to drive the EN input must be able to swing above and below the specified turnon or turnoff voltage thresholds listed in the *Electrical Characteristics* section under  $V_{II}$  and  $V_{IH}$ .

A 1-M $\Omega$  pulldown resistor ties the EN input to ground. If the EN pin is left open, the internal 1-M $\Omega$  pulldown resistor ensures that the device is turned into an OFF state by default.

When the EN pin is low, and the output is in an OFF state, the output activates an internal pulldown resistance to discharge the output capacitance for fast turnoff.

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## 8 Applications and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

## 8.1 Application Information

The LP5910 is designed to meet the requirements of RF and analog circuits, by providing low noise, high PSRR, low quiescent current, and low line or load transient response figures. The device offers excellent noise performance without the need for a noise bypass capacitor and is stable with input and output capacitors with a value of 1  $\mu$ F. The LP5910 delivers this performance in an industry-standard DSBGA package which, for this device, is specified with a  $T_J$  of  $-40^{\circ}$ C to  $+125^{\circ}$ C.

## 8.2 Typical Application

Figure 8-1 shows the typical application circuit for the LP5910. Input and output capacitances may need to be increased above 1-µF minimum for some applications.

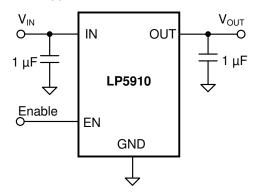


Figure 8-1. LP5910 Typical Application

#### 8.2.1 Design Requirements

For typical LP5910 applications, use the parameters listed in Table 8-1.

**Table 8-1. Design Parameters** 

DESIGN PARAMETER	EXAMPLE VALUE
Input voltage	1.3 V to 3.3 V
Output voltage	0.8 V to 2.3 V
Output current	300 mA
Output capacitor range	1 μF to 10 μF

#### 8.2.2 Detailed Design Procedure

#### 8.2.2.1 External Capacitors

Like most low-dropout regulators, the LP5910 requires external capacitors for regulator stability. The device is specifically designed for portable applications requiring minimum board space and smallest components. These capacitors must be correctly selected for good performance.

## 8.2.2.2 Input Capacitor

An input capacitor is required for stability. It is recommended that a 1-µF capacitor be connected from the LP5910 IN pin to ground. (This capacitance value may be increased without limit.) The input capacitor must be

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located a distance of not more than 1 cm from the IN pin and returned to a clean analog ground. Any good quality ceramic, tantalum, or film capacitor may be used at the input.

#### **Note**

Tantalum capacitors can suffer catastrophic failures due to surge current when connected to a low-impedance source of power (like a battery or a very large capacitor). If a tantalum capacitor is used at the input, it must be guaranteed by the manufacturer to have a surge current rating sufficient for the application. There are no requirements for the equivalent series resistance (ESR) on the input capacitor, but tolerance and temperature coefficient must be considered when selecting the capacitor to ensure the capacitance remains 1  $\mu$ F ±30% over the entire operating temperature range.

#### 8.2.2.3 Output Capacitor

For capacitance values in the range of 1  $\mu F$  to 4.7  $\mu F$ , ceramic capacitors are the smallest, least expensive and have the lowest ESR values, thus making them best for eliminating high frequency noise. The ESR of a typical 1- $\mu F$  ceramic capacitor is in the range of 20 m $\Omega$  to 40 m $\Omega$ , which easily meets the ESR requirement for stability for the LP5910. The temperature performance of ceramic capacitors varies by type. Most large value ceramic capacitors (  $\geq$  2.2  $\mu F$ ) are manufactured with Z5U or Y5V temperature characteristics, which results in the capacitance dropping by more than 50% as the temperature goes from 25°C to 85°C.

A better choice for temperature coefficient in a ceramic capacitor is X7R. This type of capacitor is the most stable and holds the capacitance within  $\pm 15\%$  over the temperature range. Tantalum capacitors are less desirable than ceramic for use as output capacitors because they are more expensive when comparing equivalent capacitance and voltage ratings in the 1- $\mu$ F to 4.7- $\mu$ F range.

## 8.2.2.4 Capacitor Characteristics

The LP5910 is designed to work with ceramic capacitors on the input and output to take advantage of the benefits they offer. For capacitance values in the range of 1  $\mu$ F to 10  $\mu$ F, ceramic capacitors are the smallest, least expensive and have the lowest ESR values, thus making them best for eliminating high frequency noise. The ESR of a typical 1- $\mu$ F ceramic capacitor is in the range of 20 m $\Omega$  to 40 m $\Omega$ , which easily meets the ESR requirement for stability for the LP5910.

A better choice for temperature coefficient in a ceramic capacitor is X7R. This type of capacitor is the most stable and holds the capacitance within  $\pm 15\%$  over the temperature range. Tantalum capacitors are less desirable than ceramic for use as output capacitors because they are more expensive when comparing equivalent capacitance and voltage ratings in the  $1-\mu F$  to  $10-\mu F$  range.

Another important consideration is that tantalum capacitors have higher ESR values than equivalent size ceramics. This means that while it may be possible to find a tantalum capacitor with an ESR value within the stable range, it would have to be larger in capacitance (which means bigger and more costly) than a ceramic capacitor with the same ESR value. Also, the ESR of a typical tantalum increases about 2:1 as the temperature goes from 25°C down to -40°C, so some guard band must be allowed.

## 8.2.2.5 Remote Capacitor Operation

The LP5910 requires at least a 1-µF capacitor at the OUT pin, but there is no strict requirements about the location of the capacitor in regards to the pin. In practical designs the output capacitor may be located up to 10 cm away from the LDO. This means that there is no need to have a special capacitor close to the OUT pin if there is already respective capacitors in the system (like a capacitor at the input of supplied part). The remote capacitor feature helps user to minimize the number of capacitors in the system.

As a good design practice, keep the wiring parasitic inductance at a minimum, using as wide as possible traces from the LDO output to the capacitors, keeping the LDO output trace layer as close as possible to ground layer and avoiding vias on the path. If there is a need to use vias, implement as many vias as possible between the connection layers. It is recommended to keep parasitic wiring inductance less than 35 nH. For the applications with fast load transients, an input capacitor is recommended, equal to or larger to the sum of the capacitance at the output node, for the best load-transient performance.

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## 8.2.2.6 No-Load Stability

The LP5910 remains stable, and in regulation, with no external load.

#### 8.2.2.7 Enable Control

The LP5910 may be switched to an ON or OFF state by a logic input at the EN pin. A voltage on this pin greater than  $V_{IH}$  turns the device on, while a voltage less than  $V_{IL}$  turns the device off.

When the EN pin is low, the regulator output is off and the device typically consumes less than 1  $\mu$ A. Additionally, an output pulldown circuit is activated which ensures that any charge stored on  $C_{OUT}$  is discharged to ground.

If the application does not require the use of the shutdown feature, the EN pin can be tied directly to the IN pin to keep the regulator output permanently on.

An internal 1-M $\Omega$  pulldown resistor ties the EN input to ground, ensuring that the device remains off if the EN pin is left open circuit. To ensure proper operation, the signal source used to drive the EN pin must be able to swing above and below the specified turn-on/off voltage thresholds listed in the *Electrical Characteristics* under V<sub>IL</sub> and V<sub>IH</sub>.

**Table 8-2. Recommended Output Capacitor Specification** 

PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
Output capacitor, C <sub>OUT</sub>	Capacitance for stability	0.7	1	10	μF
	ESR	5		500	mΩ

### 8.2.2.8 Power Dissipation

Knowing the device power dissipation and proper sizing of the thermal plane connected to the tab or pad is critical to ensuring reliable operation. Device power dissipation depends on input voltage, output voltage, and load conditions and can be calculated with Equation 1.

$$P_{D(MAX)} = (V_{IN(MAX)} - V_{OUT}) \times I_{OUT(MAX)}$$
(1)

Power dissipation can be minimized, and greater efficiency can be achieved, by using the lowest available voltage drop option that would still be greater than the dropout voltage (V<sub>DO</sub>). However, keep in mind that higher voltage drops result in better dynamic (that is, PSRR and transient) performance.

On the WSON (DRV) package, the primary conduction path for heat is through the exposed power pad to the PCB. To ensure the device does not overheat, connect the exposed pad, through thermal vias, to an internal ground plane with an appropriate amount of copper PCB area.

On the DSBGA (YKA) package, the primary conduction path for heat is through the four bumps to the PCB.

The maximum allowable junction temperature  $(T_{J(MAX)})$  determines maximum power dissipation allowed  $(P_{D(MAX)})$  for the device package.

Power dissipation and junction temperature are most often related by the junction-to-ambient thermal resistance  $(R_{\theta JA})$  of the combined PCB and device package and the temperature of the ambient air  $(T_A)$ , according to Equation 2 or Equation 3:

$$T_{J(MAX)} = T_{A(MAX)} + (R_{\theta JA} \times P_{D(MAX)}) \tag{2}$$

$$P_{D(MAX)} = (T_{J(MAX)} - T_{A(MAX)}) / R_{\theta JA}$$
(3)

Unfortunately, this  $R_{\theta JA}$  is highly dependent on the heat-spreading capability of the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The  $R_{\theta JA}$  recorded in *Thermal Information* is determined by the specific EIA/JEDEC JESD51-7 standard for PCB and copper-spreading area, and is to be used only as a relative measure of package thermal performance. For a well-designed thermal layout,  $R_{\theta JA}$  is actually the sum of the package junction-to-case (bottom) thermal resistance ( $R_{\theta JCbot}$ ) plus the thermal resistance contribution by the PCB copper area acting as a heat sink.



## 8.2.2.9 Estimating Junction Temperature

The EIA/JEDEC standard recommends the use of psi ( $\Psi$ ) thermal characteristics to estimate the junction temperatures of surface mount devices on a typical PCB board application. These characteristics are not true thermal resistance values, but rather package specific thermal characteristics that offer practical and relative means of estimating junction temperatures. These psi metrics are determined to be significantly independent of copper-spreading area. The key thermal characteristics ( $\Psi_{JT}$  and  $\Psi_{JB}$ ) are given in *Thermal Information* and are used in accordance with Equation 4 or Equation 5.

$$T_{J(MAX)} = T_{TOP} + (\Psi_{JT} \times P_{D(MAX)}) \tag{4}$$

#### where

- P<sub>D(MAX)</sub> is explained in Equation 1.
- T<sub>TOP</sub> is the temperature measured at the center-top of the device package.

$$T_{J(MAX)} = T_{BOARD} + (\Psi_{JB} \times P_{D(MAX)})$$
 (5)

#### where

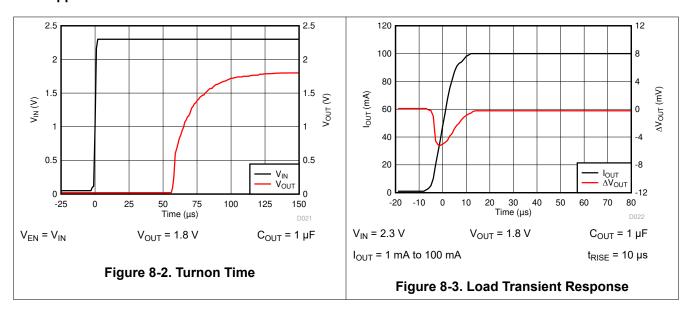
- P<sub>D(MAX)</sub> is explained in Equation 1.
- T<sub>BOARD</sub> is the PCB surface temperature measured 1-mm from the device package and centered on the package edge.

For more information about the thermal characteristics  $\Psi_{JT}$  and  $\Psi_{JB}$ , see the *Semiconductor and IC Package Thermal Metrics* application report, available for download at www.ti.com.

For more information about measuring T<sub>TOP</sub> and T<sub>BOARD</sub>, see the *Using New Thermal Metrics* application report, available for download at www.ti.com.

For more information about the EIA/JEDEC JESD51 PCB used for validating R<sub>0JA</sub>, see the *Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs* application report, available for download at www.ti.com.

#### 8.2.3 Application Curves





# 9 Power Supply Recommendations

This device is designed to operate from an input supply voltage range of 1.3 V to 3.3 V. The input supply must be well regulated and free of spurious noise. To ensure that the LP5910 output voltage is well regulated and dynamic performance is optimum, the input supply must be at least  $V_{OUT}$  + 0.5 V.



## 10 Layout

## 10.1 Layout Guidelines

The dynamic performance of the LP5910 is dependant on the layout of the PCB. PCB layout practices that are adequate for typical LDOs may degrade the PSRR, noise, or transient performance of the LP5910.

Best performance is achieved by placing  $C_{\text{IN}}$  and  $C_{\text{OUT}}$  on the same side of the PCB as the LP5910 device, and as close as is practical to the package. The ground connections for  $C_{\text{IN}}$  and  $C_{\text{OUT}}$  must be back to the LP5910 GND pin using as wide and as short of a copper trace as is practical.

Avoid connections using long trace lengths, narrow trace widths, and/or connections through vias. These add parasitic inductances and resistance that results in inferior performance especially during transient conditions.

## 10.1.1 DSBGA Mounting

The DSBGA package requires specific mounting techniques, which are detailed in the *DSBGA Wafer Level Chip Scale Package* application note. For best results during assembly, alignment ordinals on the PC board may be used to facilitate placement of the DSBGA device.

## 10.1.2 DSBGA Light Sensitivity

Exposing the DSBGA device to direct light may cause incorrect operation of the device. High intensity light sources such as halogen lamps can affect electrical performance if they are situated in close proximity to the device. The wavelengths that have the most detrimental effect are reds and infra-reds, which means that the fluorescent lighting used inside most buildings has little effect on performance.

## 10.2 Layout Examples

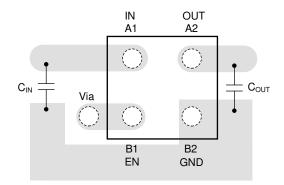


Figure 10-1. LP5910 Typical DSBGA Layout

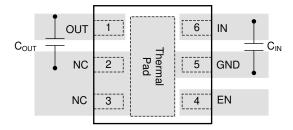


Figure 10-2. LP5910 Typical WSON Layout

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# 11 Device and Documentation Support

## 11.1 Documentation Support

#### 11.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, AN-1112 DSBGA Wafer Level Chip Scale Package application note
- Texas Instruments, Semiconductor and IC Package Thermal Metrics application report
- Texas Instruments, Using New Thermal Metrics application report
- Texas Instruments, Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs application report

## 11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

## 11.3 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 11.4 Trademarks

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All trademarks are the property of their respective owners.

#### 11.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 11.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

## Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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## **PACKAGING INFORMATION**

Orderable part number	Status (1)	Material type	Package   Pins	Package qty   Carrier	<b>RoHS</b> (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
LP5910-0.9YKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	D
LP5910-0.9YKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	D
LP5910-1.0DRVR	Active	Production	WSON (DRV)   6	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59A
LP5910-1.0DRVR.B	Active	Production	WSON (DRV)   6	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59A
LP5910-1.0YKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	А
LP5910-1.0YKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	А
LP5910-1.1BYKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	T
LP5910-1.1BYKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	T
LP5910-1.1BYKAT	Active	Production	DSBGA (YKA)   4	250   SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	T
LP5910-1.1BYKAT.B	Active	Production	DSBGA (YKA)   4	250   SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	Т
LP5910-1.1YKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	E
LP5910-1.1YKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	E
LP5910-1.2YKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	В
LP5910-1.2YKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	В
LP5910-1.725YKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	N
LP5910-1.725YKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	N
LP5910-1.825YKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	0
LP5910-1.825YKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	0
LP5910-1.825YKAT	Active	Production	DSBGA (YKA)   4	250   SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	0
LP5910-1.825YKAT.B	Active	Production	DSBGA (YKA)   4	250   SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	0
LP5910-1.8DRVR	Active	Production	WSON (DRV)   6	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59C
LP5910-1.8DRVR.B	Active	Production	WSON (DRV)   6	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59C
LP5910-1.8DRVT	Active	Production	WSON (DRV)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59C
LP5910-1.8DRVT.B	Active	Production	WSON (DRV)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59C
LP5910-1.8DRVTG4	Active	Production	WSON (DRV)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59C
LP5910-1.8DRVTG4.B	Active	Production	WSON (DRV)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	59C
LP5910-1.8YKAR	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	С
LP5910-1.8YKAR.B	Active	Production	DSBGA (YKA)   4	3000   LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	С
LP5910-1.8YKAT	Active	Production	DSBGA (YKA)   4	250   SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	С



## PACKAGE OPTION ADDENDUM

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Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
						(4)	(5)		
LP5910-1.8YKAT.B	Active	Production	DSBGA (YKA)   4	250   SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 125	С

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.



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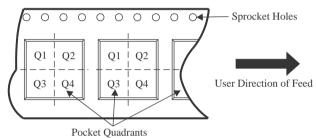
## TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

## QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LP5910-0.9YKAR	DSBGA	YKA	4	3000	180.0	8.4	0.8	0.8	0.47	2.0	8.0	Q1
LP5910-1.0DRVR	WSON	DRV	6	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
LP5910-1.0YKAR	DSBGA	YKA	4	3000	180.0	8.4	0.8	0.8	0.47	2.0	8.0	Q1
LP5910-1.1BYKAR	DSBGA	YKA	4	3000	180.0	8.4	0.8	0.8	0.47	4.0	8.0	Q1
LP5910-1.1BYKAT	DSBGA	YKA	4	250	180.0	8.4	0.8	0.8	0.47	4.0	8.0	Q1
LP5910-1.1YKAR	DSBGA	YKA	4	3000	180.0	8.4	0.8	0.8	0.47	4.0	8.0	Q1
LP5910-1.2YKAR	DSBGA	YKA	4	3000	180.0	8.4	0.8	0.8	0.47	2.0	8.0	Q1
LP5910-1.725YKAR	DSBGA	YKA	4	3000	180.0	8.4	8.0	0.8	0.47	4.0	8.0	Q1
LP5910-1.825YKAR	DSBGA	YKA	4	3000	180.0	8.4	8.0	0.8	0.47	4.0	8.0	Q1
LP5910-1.825YKAT	DSBGA	YKA	4	250	180.0	8.4	8.0	0.8	0.47	4.0	8.0	Q1
LP5910-1.8DRVR	WSON	DRV	6	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
LP5910-1.8DRVT	WSON	DRV	6	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
LP5910-1.8DRVTG4	WSON	DRV	6	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
LP5910-1.8YKAR	DSBGA	YKA	4	3000	180.0	8.4	8.0	0.8	0.47	4.0	8.0	Q1
LP5910-1.8YKAT	DSBGA	YKA	4	250	180.0	8.4	0.8	0.8	0.47	4.0	8.0	Q1



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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LP5910-0.9YKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.0DRVR	WSON	DRV	6	3000	182.0	182.0	20.0
LP5910-1.0YKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.1BYKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.1BYKAT	DSBGA	YKA	4	250	182.0	182.0	20.0
LP5910-1.1YKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.2YKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.725YKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.825YKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.825YKAT	DSBGA	YKA	4	250	182.0	182.0	20.0
LP5910-1.8DRVR	WSON	DRV	6	3000	182.0	182.0	20.0
LP5910-1.8DRVT	WSON	DRV	6	250	182.0	182.0	20.0
LP5910-1.8DRVTG4	WSON	DRV	6	250	182.0	182.0	20.0
LP5910-1.8YKAR	DSBGA	YKA	4	3000	182.0	182.0	20.0
LP5910-1.8YKAT	DSBGA	YKA	4	250	182.0	182.0	20.0



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

4206925/F





PLASTIC SMALL OUTLINE - NO LEAD



## NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature
- number SLUA271 (www.ti.com/lit/slua271).

  5. Vias are optional depending on application, refer to device data sheet. If some or all are implemented, recommended via locations are shown.



PLASTIC SMALL OUTLINE - NO LEAD



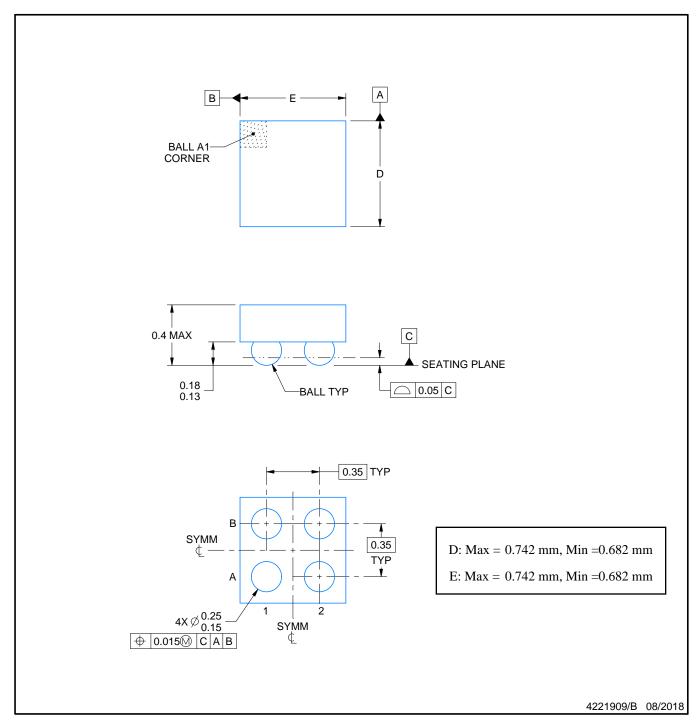
NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.





DIE SIZE BALL GRID ARRAY

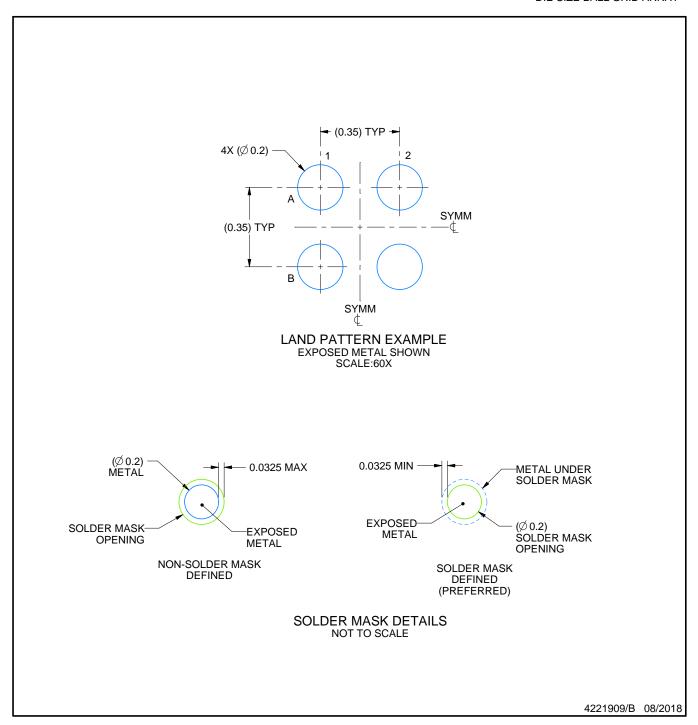


## NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.



DIE SIZE BALL GRID ARRAY

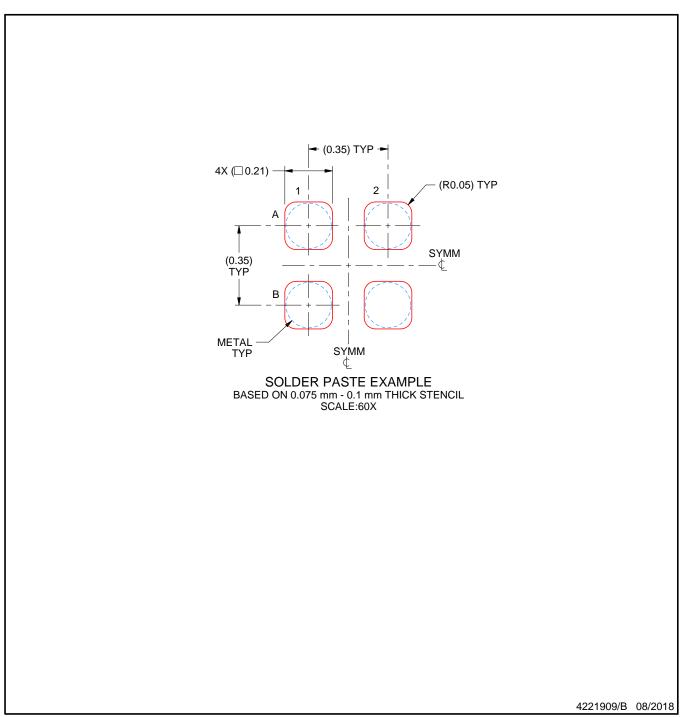


NOTES: (continued)

3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints. For more information, see Texas Instruments literature number SNVA009 (www.ti.com/lit/snva009).



DIE SIZE BALL GRID ARRAY



NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.



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