

# ISOS510-SEP Radiation Tolerant, Current-Driven Analog Isolator With Transistor Output

#### 1 Features

- · Radiation Performance
  - Total Ionizing Dose (TID) Characterized (ELDRS-Free) up to 50krad(Si)
  - TID RLAT up to 30krad(Si)
  - Single-Event Latch-up (SEL) Immune to LET up to 43MeV-cm<sup>2</sup>/mg at 125°C
  - Single-event functional interrupt (SEFI) and single-event transient (SET) characterized LET up to 43MeV-cm<sup>2</sup>/mg
- · Space Enhanced Plastic (Space EP)
  - Meets NASA ASTM E595 Outgassing Spec
  - Military Temp Range (-55°C to 125°C)
- 1-channel diode input
- Current transfer ratio (CTR) at I<sub>F</sub> = 5mA, V<sub>CE</sub> = 5V: 100% to 155%
- High collector-emitter voltage: V<sub>CE</sub> (max) = 30V
- Robust SiO<sub>2</sub> isolation barrier
  - Isolation rating: 3750V<sub>RMS</sub>
  - Working voltage: 500V<sub>RMS</sub>, 707V<sub>PK</sub>
  - Surge capability: up to 10kV
- Response time:  $3\mu s$  (typical) at  $V_{CE}$  = 10V,  $I_{C}$  = 2mA,  $R_{L}$  = 100 $\Omega$
- Small 4-pin package (DFG)
- Safety-related certifications:
  - UL 1577 recognition
  - DIN EN IEC 60747-17 (VDE 0884-17) conformity per VDE

# 2 Applications

- Satellite electrical power systems (EPS)
- · Communications Payload
- Radar Imaging Payload

#### 3 Description

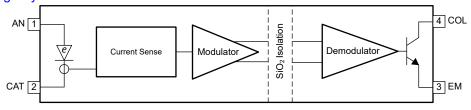
The ISOS510 radiation-tolerant device is a single-channel, current-driven, analog isolator with transistor output. The device offers significant reliability and performance advantages compared to other current-driven analog isolators, including high bandwidth, low turn-off delay, low power consumption, wider temperature ranges, flat current transfer ratio (CTR), and tight process controls resulting in small part-to-part skew. These performance advantages stay stable across radiation, temperature, and lifetime.

ISOS510 is offered in small a SOIC-4 package with 2.54mm pin pitch, supporting a  $3.75 \mathrm{kV_{RMS}}$  isolation rating . The high performance and reliability of ISOS510 enables these devices to be used in aerospace & defense applications such as feedback loops in isolated DC/DC modules, satellite propulsion power processing units, spacecraft battery management systems, and more.

#### **Package Information**

PART NUMBER	GRADE	PACKAGE (1)	PACKAGE SIZE <sup>(2)</sup>
ISOS510- SEP	SEP	DFG (SO-4) 4-pin plastic	7.0mm × 3.5mm Mass = 89.6mg

- (1) For more information, see Section 11.
- (2) The package size (length × width) is a nominal value and includes pins, where applicable.



Simplified Schematic



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# **4 Pin Configuration and Functions**

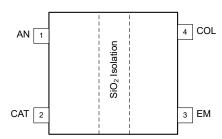


Figure 4-1. ISOS510 4-Pin SOIC (Top View)

#### **Table 4-1. Pin Functions**

	PIN	TYPE (1)	DESCRIPTION	
NO.	NAME	IIPE	DESCRIPTION	
1	AN	I	Anode connection of diode input	
2	CAT	I	Cathode connection of diode input	
3	EM	0	Emitter for transistor	
4	COL	0	Collector for transistor	

(1) I = Input, O = Output



# **5 Specifications**

## **5.1 Absolute Maximum Ratings**

over operating free-air temperature range of -55°C to +125°C (unless otherwise noted)(1)

		MIN	MAX	UNIT
I <sub>F(max)</sub>	Maximum Input forward current		15	mA
V <sub>CEO</sub>	Collector-emitter voltage		35	V
V <sub>ECO</sub>	Emitter-collector voltage		7	V
I <sub>FP</sub>	Input pulse forward current (1µs width at T <sub>A</sub> = 25°C)		1	А
V <sub>R</sub>	Input reverse voltage		7	V
Pı	Input power dissipation		86	mW
I <sub>C</sub>	Collector current		15	mA
P <sub>C</sub>	Collector power dissipation		86	mW
P <sub>T</sub>	Total power dissipation		172	mW
T <sub>A</sub>	Ambient temperature	-55	125	°C
TJ	Junction temperature	-55	150	°C

<sup>1)</sup> Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under the operational sections of this document. If used outside the listed operational conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

#### 5.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	±2000	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>	±1000	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

## **5.3 Recommended Operating Conditions**

over operating free-air temperature range of -55°C to +125°C (unless otherwise noted)

		MIN	NOM MAX	UNIT
I <sub>F(ON)</sub>	Input ON-state forward current	0.7	10	mA
V <sub>R</sub>	Input reverse voltage		5	V
V <sub>CEO</sub>	Collector-emitter voltage	-5	30	V
T <sub>A</sub>	Ambient temperature	-55	125	°C

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# **5.4 Thermal Information**

		ISOS510-SEP	
	THERMAL METRIC(1)	DFG (SOIC)	UNIT
		4 PINS	
R <sub>0JA</sub>	Junction-to-ambient thermal resistance	283.9	°C/W
R <sub>0</sub> JC(top)	Junction-to-case (top) thermal resistance	173.1	°C/W
R <sub>0JB</sub>	Junction-to-board thermal resistance	201.4	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	125.1	°C/W
ΨЈВ	Junction-to-board characterization parameter	198.0	°C/W

(1) For more information about traditional and new thermal metrics, see Semiconductor and IC Package Thermal Metrics application note.



#### 5.5 Insulation Specifications

	PARAMETER	TEST CONDITIONS	VALUE 4-DFG	UNIT
IEC 6060	64-1			
CLR	External clearance <sup>(1)</sup>	Side 1 to side 2 distance through air	> 5	mm
CPG	External creepage <sup>(1)</sup>	Side 1 to side 2 distance across package surface	> 5	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	μm
CTI	Comparative tracking index	IEC 60112; UL 746A	>400	V
	Material Group	According to IEC 60664-1	II	
		Rated mains voltage ≤ 150 V <sub>RMS</sub>	I-IV	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 300 V <sub>RMS</sub>	I-IV	
		Rated mains voltage ≤ 600 V <sub>RMS</sub>	I-III	
DIN VDE	V 0884-11:2017 <sup>(6)</sup>			
V <sub>IORM</sub>	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	707	V <sub>PK</sub>
V <sub>IOWM</sub>	Maximum isolation working voltage	AC voltage (sine wave); time-dependent dielectric breakdown (TDDB) test	500	V <sub>RMS</sub>
1011111		DC voltage	707 5303	V <sub>DC</sub>
$V_{IOTM}$	Maximum transient isolation voltage	V <sub>TEST</sub> = V <sub>IOTM</sub> , t = 60s (qualification); V <sub>TEST</sub> = 1.2 × V <sub>IOTM</sub> , t = 1s (100% production)	5303	$V_{PK}$
V <sub>IMP</sub>	Maximum impulse voltage (2)	Tested in air, 1.2/50µs waveform per IEC 62368-1	7200	$V_{PK}$
V <sub>IOSM</sub>	Maximum surge isolation voltage <sup>(3)</sup>	V <sub>IOSM</sub> ≥ 1.3 x V <sub>IMP</sub> ; tested in oil (qualification test), 1.2/50µs waveform per IEC 62368-1	10000	$V_{PK}$
		Method a: After I/O safety test subgroup 2/3, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60s$ ; $V_{pd(m)} = 1.2 \times V_{IORM}$ , $t_m = 10s$	≤ 5	
q <sub>pd</sub>	Apparent charge <sup>(4)</sup>	Method a: After environmental tests subgroup 1, $V_{ini}$ = $V_{IOTM}$ . $t_{ini}$ = 60s; $V_{pd(m)}$ = 1.6 × $V_{IORM}$ , $t_m$ = 10s	≤ 5	рС
		Method b: At routine test (100% production) and preconditioning (type test), $V_{ini}$ = 1.2 × $V_{IOTM}$ , $t_{ini}$ = 1s; $V_{pd(m)}$ = 1.875 × $V_{IORM}$ , $t_m$ = 1s	≤ 5	
C <sub>IO</sub>	Barrier capacitance, input to output <sup>(5)</sup>	V <sub>IO</sub> = 0.4 × sin (2 πft), f = 1MHz	1	pF
		V <sub>IO</sub> = 500V, T <sub>A</sub> = 25°C	> 10 <sup>12</sup>	
R <sub>IO</sub>	Insulation resistance, input to output <sup>(5)</sup>	V <sub>IO</sub> = 500V, 100°C ≤ T <sub>A</sub> ≤ 125°C	> 10 <sup>11</sup>	Ω
		V <sub>IO</sub> = 500V at T <sub>S</sub> = 150°C	> 10 <sup>9</sup>	
	Pollution degree		2	
	Climatic category		40/125/21	
UL 1577	•			
V <sub>ISO</sub>	Withstand isolation voltage	$V_{TEST} = V_{ISO}$ , t = 60s (qualification); $V_{TEST} = 1.2 \times V_{ISO}$ , t = 1s (100% production)	3750	V <sub>RMS</sub>

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves, ribs, or both on a printed circuit board are used to help increase these specifications.
- (2) Testing is carried out in air to determine the surge immunity of the package.
- (3) Testing is carried out in oil to determine the intrinsic surge immunity of the isolation barrier.
- (4) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (5) All pins on each side of the barrier tied together creating a two-pin device.
- (6) This coupler is suitable for *safe electrical insulation only* within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.



#### 5.6 Safety-Related Certifications

VDE	UL
Plan to certify according to DIN EN IEC 60747-17 (VDE 0884-17)	Plan to certify according to UL 1577 Component Recognition Program
Certificate planned	Certificate planned

#### 5.7 Safety Limiting Values

Safety limiting<sup>(1)</sup> intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
SO-4 PACKAGE (DFG)							
Is		$R_{\theta JA}$ = 283.9°C/W, $V_F$ = 1.4V, $T_J$ = 150°C, $T_A$ = 25°C			300	mA	
	Cofety limiting input summer	$R_{\theta JA}$ =283.9°C/W, $V_{CEO}$ = 30V, $T_J$ = 150°C, $T_A$ = 25°C			13.5	mA	
	Safety limiting input current	$R_{\theta JA}$ = 283.9°C/W, $V_{CEO}$ = 24V, $T_J$ =150°C, $T_A$ = 25°C			17.5	mA	
		$R_{\theta JA}$ = 283.9°C/W, $V_{CEO}$ = 15V, $T_{J}$ = 150°C, $T_{A}$ = 25°C			28	mA	
Ps	Safety limiting total power	R <sub>0JA</sub> = 283.9°C/W, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C			420	mW	
Ts	Maximum safety temperature				150	°C	

<sup>(1)</sup> The I<sub>S</sub> and P<sub>S</sub> parameters represent the safety current and safety power respectively. The maximum limits of I<sub>S</sub> and P<sub>S</sub> must not be exceeded. These limits vary with the ambient temperature, T<sub>A</sub>.

The junction-to-air thermal resistance, R<sub>BJA</sub>, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

 $T_J = T_A + R_{\theta JA} \times P$ , where P is the power dissipated in the device.

 $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$ , where  $T_{J(max)}$  is the maximum allowed junction temperature.

 $P_S = I_S \times V_I$ , where  $V_I$  is the maximum input voltage.



#### 5.8 Electrical Characteristics

over recommended operating conditions unless otherwise noted; includes RLAT at TA =  $25^{\circ}$ C if sub-group number is present for QML RHA and SEP devices

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
INPUT		·			•	
V <sub>F</sub>	Input forward voltage	I <sub>F</sub> = 5mA		1.2	1.6	V
I <sub>R</sub>	Input reverse current	V <sub>R</sub> = 5V			10	μΑ
C <sub>IN</sub>	Input capacitance	At 1MHz, $V_F = 0V$ , $T_A = 25$ °C		20		pF
OUTPUT	·	·				
C <sub>CE</sub>	Collector-emitter capacitance	1MHz, V <sub>F</sub> = 0V, T <sub>A</sub> = 25°C		15		pF
V <sub>CE(SAT)</sub>	Collector-emitter saturation voltage	I <sub>F</sub> = 10mA, I <sub>C</sub> = 1mA			0.3	V
I <sub>C_DARK</sub>	Collector dark current	V <sub>CE</sub> = 20V, I <sub>F</sub> = 0mA		-	100	nA
I <sub>EC</sub>	Reverse current	$V_{EC}$ = 5V, $I_F$ = 0mA			50	μΑ
I <sub>C_OFF</sub>	OFF_state collector current	V <sub>F</sub> = 0.7V, V <sub>CE</sub> = 30V			10	μΑ
CTR (1)	Current Transfer Ratio	I <sub>F</sub> = 2mA, V <sub>CE</sub> = 5V	80	130	180	%
CIK	Current fransier Ratio	I <sub>F</sub> = 5mA, V <sub>CE</sub> = 5V	100	120	155	%

(1) CTR (%) =  $(I_C / I_F) \times 100\%$ 

# **5.9 Switching Characteristics**

All specifications are at T<sub>A</sub> = 25°C unless otherwise noted

	PARAMETER	TEST CONDITIONS	MIN TYP	MAX	UNIT
AC					
t <sub>r</sub>	Rise time, see Figure 6-2 and Figure 6-3	$V_{CC} = 10V, I_C = 2mA, R_L = 100\Omega, C_L = 50pF$	3.2		μs
t <sub>f</sub>	Fall time, see Figure 6-2 and Figure 6-3	$V_{CC}$ = 10V, $I_{C}$ = 2mA, $R_{L}$ = 100 $\Omega$ , $C_{L}$ = 50pF	4.0		μs
	Turn on time, see Figure 6-2 and Figure	$V_{CC}$ = 10V, $I_{C}$ = 2mA, $R_{L}$ = 100 $\Omega$ , $C_{L}$ = 50pF	5.7		μs
T <sub>ON</sub>	6-3	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		μs	
		$V_{CC}$ =5V, $R_L$ =1.9k $\Omega$ , $I_F$ =16mA, $C_L$ =50pF	0.62		μs
_	Turn off time, see Figure 6-2 and Figure	, , , , , , , , , , , , , , , , , , , ,	3.6		μs
T <sub>OFF</sub>	6-3	$V_{CC}$ =5V, $R_L$ =4.7k $\Omega$ , $I_F$ =1.6mA, $C_L$ =50pF	8		μs
		$V_{CC}$ =5V, $R_L$ =1.9k $\Omega$ , $I_F$ =16mA, $C_L$ =50pF	10		μs
ts	Storage time; time required for the output waveform to change from 0% (100%) to 10% (90%) when input is turned on and back off, see Figure 6-2 and Figure 6-3	$V_{CC} = 5V$ , $I_F = 1.6$ mA, $R_L = 4.7$ k $\Omega$		21	μs
BW	Bandwidth, see Figure 6-3 and Figure 6-4	$2k\bar{\Omega}$ V <sub>CC</sub> = 5V, $\bar{R}_{LOAD}$ = 100 $\Omega$ , C <sub>L</sub> =	680		kHz



# **6 Parameter Measurement Information**

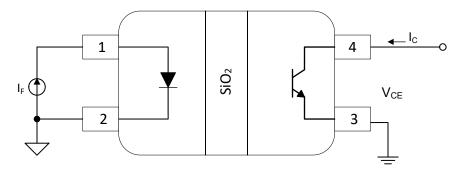


Figure 6-1. ISOS510 Test Circuit for CTR

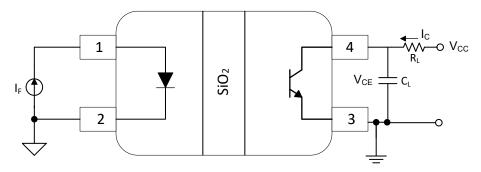


Figure 6-2. ISOS510 Test Circuit for Switching Timing

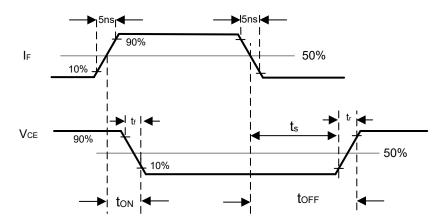


Figure 6-3. ISOS510 Switching Timing Waveforms

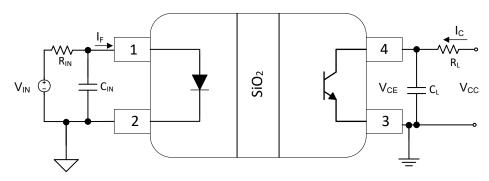


Figure 6-4. ISOS510 Test Circuit for Bandwidth



# 7 Detailed Description

#### 7.1 Overview

The functional block diagram of ISOS510 device is shown in Section 7.2. The input signal is transmitted across the isolation barrier using an on-off keying (OOK) modulation scheme. The transmitter sends a high-frequency carrier across the barrier that contains information on how much current is flowing through the input pins. The receiver demodulates the signal after advanced signal conditioning and drives the transistor in the output stage. This device also incorporates advanced circuit techniques to maximize bandwidth and minimize radiated emissions. Figure 7-2 shows conceptual details of how the OOK scheme works.

#### 7.2 Functional Block Diagram

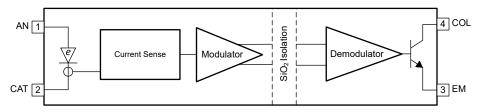


Figure 7-1. Conceptual Block Diagram of an Opto-emulator ISOS510

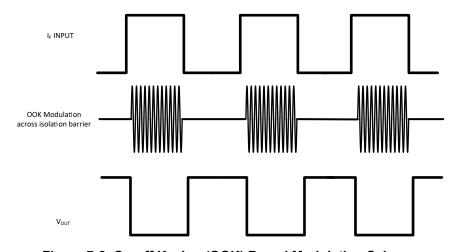


Figure 7-2. On-off Keying (OOK) Based Modulation Scheme

#### 7.3 Feature Description

The ISOS510 device can isolate both analog and digital signals due to the current-driven input stage. When supplied an input current ( $I_F$ ) of at least 0.7mA to the AN pin, this powers the internal modulator which includes the modulator and current sense blocks. The  $I_F$  is sensed and converted into a high frequency carrier that is passed across the isolation barrier. Once demodulated on the secondary side of the device, this is used to bias the output transistor which tries to sink current from an external source into the COL pin ( $I_C$ ) proportional to the  $I_F$  value. The ratio of  $I_C$  to  $I_F$  is the current transfer ratio (CTR).

#### 7.4 Device Functional Modes

#### 7.4.1 Active Mode

If the external source or circuit connected to the COL pin can supply enough current to satisfy the CTR of the device for a given  $I_F$  value, then the device is considered in "active mode". This is how analog signals can be transmitted through this device.

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#### 7.4.2 Saturation Mode

If the external source or circuit connected to the COL pin cannot supply enough current to satisfy the CTR of the device for a given  $I_F$  value (For example,  $I_C$  = 1mA when  $I_F$  = 10mA), then the output transistor saturates and goes into a low impedance state. The device is considered in "saturation mode". This is how digital signals can be transmitted through this device.

# 8 Application and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

#### 8.1 Application Information

ISOS510 is commonly used in the feedback control loops of isolated power supplies. This device is used to solve the problem of feeding back current while isolating the primary and secondary domains to regulate the output voltage.

In power supplies, the output voltage is isolated from main input voltage using a transformer (for example: flyback converter). For analog power supply units, the controller IC is typically on the primary side of the transformer. For closed loop control, measuring the output voltage on the secondary side and feeding the voltage back to the controller on the primary is necessary. The most common method of achieving this design is using an isolator such as ISOS510, error amplifier (commonly TL1431-SP), and a voltage comparator to form a feedback loop across the isolation barrier

Figure 8-1 illustrates a typical isolated power supply. In this implementation, the output voltage is sensed by an error amplifier using the resistor divider (R1 and R2). Depending on the voltage level that the error amplifier senses, the TL1431-SP can drive the current of the ISOS510 higher or lower which is then compared to a voltage reference. The information is passed across the isolation barrier through ISOS510 to the primary side, where the PWM control circuit modulates the power stage to regulate the output voltage. The TL1431-SP and ISOS510 play an important role for stable feedback and control loop.

The ISOS510 devices enable improvements in transient response, reliability, and stability as compared to commonly used optocoupler as the CTR is stable over wide temperature range providing a small, low-cost, highly reliable, and easy-to-design implementation.



#### 8.1.1 Typical Application

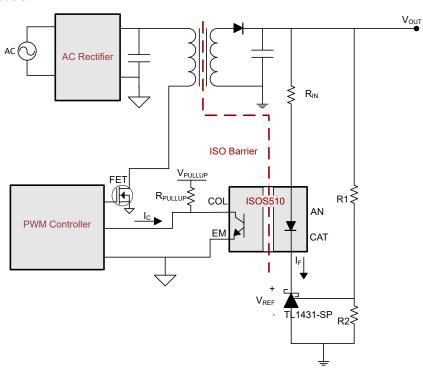


Figure 8-1. Typical Isolated Power Supply Application Using ISOS510

#### 8.1.1.1 Design Requirements

To design with ISOS510, use the parameters listed in Table 8-1.

**Table 8-1. Design Parameters** 

PARAMETER	VALUE							
Input forward current range, I <sub>F</sub>	0.7mA (min), 10mA (max)							
Current transfer ratio at I <sub>F</sub> = 5mA, CTR	100% to 155%							
Collector current tolerance, I <sub>C</sub>	30mA (max)							
Collector-emitter voltage (saturation), V <sub>CE(SAT)</sub>	0.3V (max)							
Input forward voltage, V <sub>F</sub>	1.2V (typ)							

#### 8.1.1.2 Detailed Design Procedure

This section presents the design procedure for using the ISOS510 device. External components must be selected to operate ISOS510 within the *Recommended Operating Conditions*. The following recommendations on component selection focus on the design of a typical feedback control loop for an isolated flyback converter.

When using an isolator in a feedback control loop for an isolated power supply, many variables can affect how to properly use the isolator, including the output voltage of the power supply and the type of controller the feedback signal is being sent to. For this example, assume that the output voltage of this power supply,  $V_{OUT}$ , is 5V, and the PWM controller being used has an integrated error amplifier with a COMP pin that acts as the output of this amplifier.

#### 8.1.1.2.1 Sizing R<sub>PULLUP</sub>

The transistor output of ISOS510 operates in active, saturation, reverse, and cut-off regions, just like a regular transistor. To verify that the output does not get damaged when the output is saturated, the minimum value of  $R_{PULLUP}$  can be calculated for a given pull-up voltage,  $V_{PULLUP}$ , in Equation 1:

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$$R_{\text{PULLUP}} > \frac{V_{\text{PULLUP}} - V_{\text{CE(SAT)}}}{I_{\text{C(MAX)}}} \tag{1}$$

For the example of a feedback loop application, we can calculate the minimum required value for  $R_{PULLUP}$  for a given  $V_{PULLUP}$  of 10V, the maximum output voltage of the error amplifier ( $V_{COMP(MAX)}$ ) of 2.5V, and the maximum output current of the error amplifier is internally clamped at 1.6mA. The equation to calculate  $R_{PULLUP}$  is shown in Equation 2:

$$R_{PULLUP} > \frac{V_{PULLUP} - V_{COMP(MAX)}}{I_{COMP(CLAMP)}} = \frac{10V - 2.5V}{1.6mA} = 4.66k\Omega$$
 (2)

#### 8.1.1.2.2 Sizing R<sub>IN</sub>

The input side of ISOS510 is current-driven. To limit the amount of current flowing into the AN pin, placing a series resistor, R<sub>IN</sub>, in series with the input as shown in Figure 8-1 is recommended.

Depending on how the ISOS510 device is being used, the value of  $R_{\text{IN}}$  can vary quite a bit. However, at a high level, to make sure the input does not get damaged, the minimum value of  $R_{\text{IN}}$  can be calculated for a given input voltage,  $V_{\text{IN}}$ , in Equation 3:

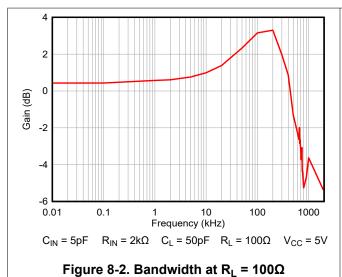
$$R_{\rm IN} > \frac{V_{\rm IN} - V_{\rm F}}{I_{\rm C(MAX)}} \tag{3}$$

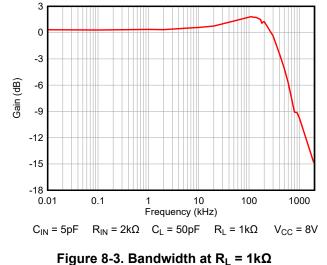
However, in the use case of a feedback loop,  $R_{IN}$  directly affects the mid-band gain of the loop. Assuming that the TL1431-SP has been configured to give a reference voltage,  $V_{REF}$ , of 2.5V and  $R_{PULLUP}$  is  $5k\Omega$ , Equation 4 is used to calculate the maximum value of  $R_{IN}$  verifying that the  $V_{COMP}$  voltage on the primary side can be pulled to the saturation voltage of the ISOS510,  $V_{CE(SAT)}$ .

$$R_{\text{IN}} < \frac{(V_{\text{OUT}} - V_{\text{REF}} - V_{\text{F}}) \times R_{\text{PULLUP}} \times CTR_{\text{MIN}}}{V_{\text{PULLUP}} - V_{\text{CE}(\text{SAT})}} = \frac{(5V - 2.5V - 1.2V) \times 5k\Omega \times 100\%}{10V - 0.3V} = 670\Omega \tag{4}$$

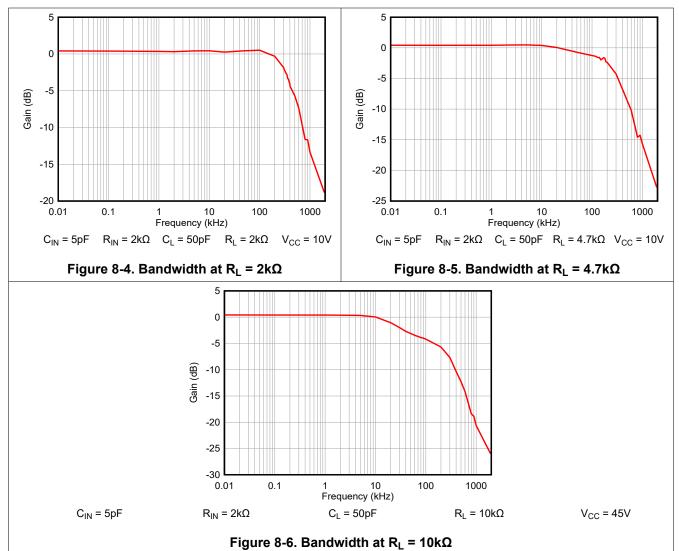
#### 8.1.1.3 Application Curves

The following curves shows ISOS510 bandwidth performance over different loading conditions where  $V_{IN} = 5V_{DC} + 2V_{PK}$ . See Figure 6-4 for setup details.





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# 8.2 Power Supply Recommendations

ISOS510 does not require a dedicated power supply to operate since there is no supply pin. Take care to not violate recommended I/O specifications, such as the minimum I<sub>F</sub>, for proper device functionality.

#### 8.3 Layout

#### 8.3.1 Layout Guidelines

- The device connections to ground must be tied to the PCB ground plane using a direct connection or two vias to help minimize inductance.
- The connections of capacitors and other components to the PCB ground plane must use a direct connection or two vias for minimum inductance.

#### 8.3.2 Layout Example

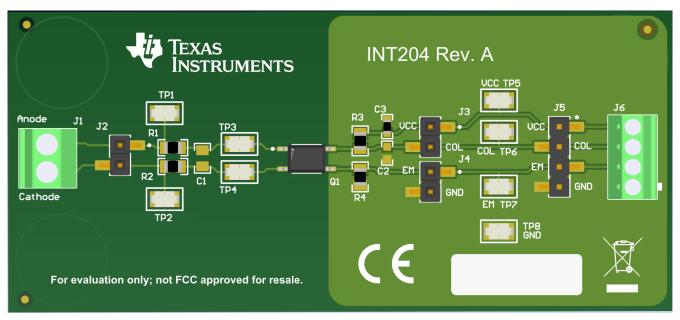


Figure 8-7. Layout Example of ISOS510 With a Single Layer Board



# 9 Device and Documentation Support

## 9.1 Documentation Support

#### 9.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, Isolation Glossary, application note
- Texas Instruments, Introduction to Opto-Emulators, application note

# 9.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

# 9.3 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 9.4 Trademarks

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# 9.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

# 9.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

#### 10 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

DATE	REVISION	NOTES
August 2025	*	Initial Release

# 11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

Submit Document Feedback

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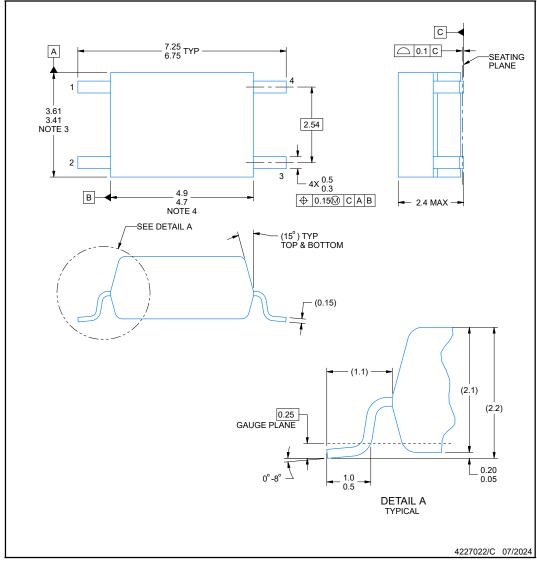
# **DFG0004A**



#### **PACKAGE OUTLINE**

# SOIC - 2.4 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



#### NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
   This drawing is subject to change without notice.
- This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 mm per side.
   This dimension does not include interlead flash.



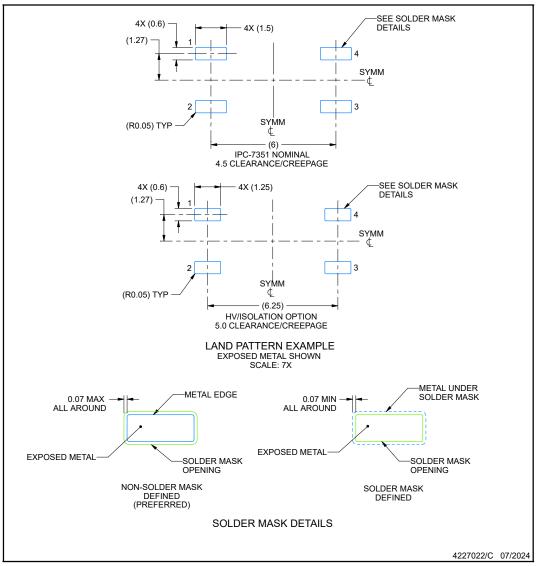


#### **EXAMPLE BOARD LAYOUT**

# **DFG0004A**

#### SOIC - 2.4 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



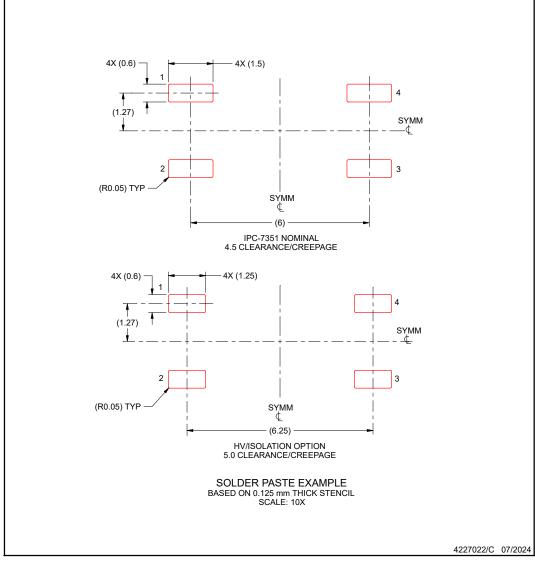


#### **EXAMPLE STENCIL DESIGN**

# **DFG0004A**

#### SOIC - 2.4 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



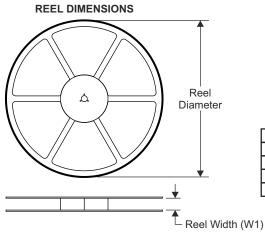
NOTES: (continued)

- 7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.8. Board assembly site may have different recommendations for stencil design.





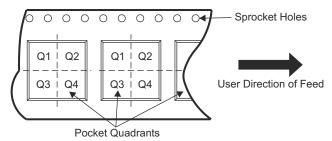
# 11.1 Tape and Reel Information



# **TAPE DIMENSIONS** Ф B0 ▼ Ф

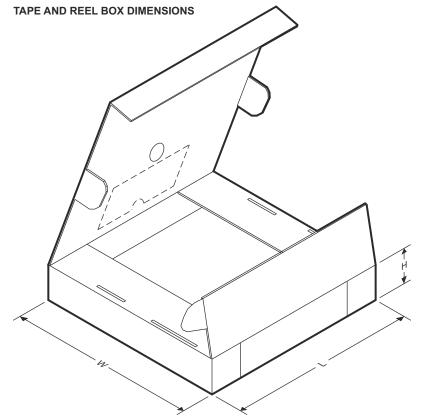
A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
XISOS510DFGTSEP	SO-4	DFG	4	250	330.0	12.4	8.0	3.8	2.7	12.0	12.0	Q1





Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
XISOS510DFGTSEP	SO-4	DFG	4	250	356.0	356.0	35.0

www.ti.com 28-Aug-2025

#### PACKAGING INFORMATION

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
						(4)	(5)		
XISOS510DFGTSEP	Active	Preproduction	SOIC (DFG)   4	250   SMALL T&R	-	Call TI	Call TI	-55 to 125	

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

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