











CDCLVP1212

SCAS886E - AUGUST 2009 - REVISED DECEMBER 2015

CDCLVP1212 LVPECL Output, High-Performance Clock Buffer

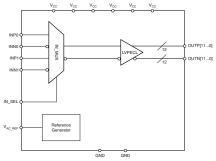
Features

- 2:12 Differential Buffer
- Selectable Clock Inputs Through Control Terminal
- Universal Inputs Accept LVPECL, LVDS, and LVCMOS/LVTTL
- 12 LVPECL Outputs
- Maximum Clock Frequency: 2 GHz
- Maximum Core Current Consumption: 88 mA
- Very Low Additive Jitter: <100 fs, rms in 10-kHz to 20-MHz Offset Range:
 - 57 fs, rms (typ) @ 122.88 MHz
 - 48 fs, rms (typ) @ 156.25 MHz
 - 30 fs, rms (typ) @ 312.5 MHz
- 2.375-V to 3.6-V Device Power Supply
- Maximum Propagation Delay: 550 ps
- Maximum Output Skew: 25 ps
- LVPECL Reference Voltage, V_{AC REF}, Available for Capacitive-Coupled Inputs
- Industrial Temperature Range: -40°C to 85°C
- ESD Protection Exceeds 2 kV (HBM)
- Supports 105°C PCB Temperature (Measured with a Thermal Pad)
- Available in 6-mm x 6-mm QFN-40 (RHA) Package

Applications

- Wireless Communications
- Telecommunications/Networking
- Medical Imaging
- Test and Measurement Equipment

Functional Block Diagram



3 Description

The CDCLVP1212 is a highly versatile, low additive jitter buffer that can generate 12 copies of LVPECL clock outputs from one of two selectable LVPECL, LVDS, or LVCMOS inputs for a variety of communication applications. It has a maximum clock frequency up to 2 GHz. The CDCLVP1212 features an on-chip multiplexer (MUX) for selecting one of two inputs that can be easily configured solely through a terminal. The overall additive control performance is less than 0.1 ps, RMS from 10 kHz to 20 MHz, and overall output skew is as low as 25 ps, making the device a perfect choice for use in demanding applications.

The CDCLVP1212 clock buffer distributes one of two selectable clock inputs (IN0, IN1) to 12 pairs of differential LVPECL clock outputs (OUT0, OUT11) with minimum skew for clock distribution. The CDCLVP1212 can accept two clock sources into an input multiplexer. The inputs can be LVPECL, LVDS, or LVCMOS/LVTTL.

The CDCLVP1212 is specifically designed for driving $50-\Omega$ transmission lines. When driving the inputs in single-ended mode, the LVPECL bias voltage (V_{AC REF}) should be applied to the unused negative input terminal. However, for high-speed performance up to 2 GHz, differential mode is strongly recommended.

The CDCLVP1212 is packaged in a small 40terminal, 6-mm x 6-mm QFN package and is characterized for operation from -40°C to 85°C.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
CDCLVP1212	QFN (40)	6.00 mm x 6.00 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.

Peak-to-Peak Voltage vs. Frequency

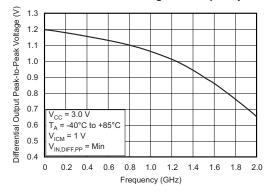




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	8.1 Test Configurations			

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	nanges from Revision D (September 2014) to Revision E	Page
•	Changed Handling Ratings to ESD Ratings; moved T _J and T _{stg} to Absolute Maximum Ratings	6
•	Added PCB temperature in Recommended Operating Conditions	6
•	Added V _{OH} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 2.375 V to 2.625	5 V <mark>7</mark>
•	Added V _{OL} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 2.375 V to 2.625	5 V <mark>7</mark>
•	Added I _{EE} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 2.375 V to 2.625	5 V <mark>8</mark>
•	Added I _{CC} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 2.375 V to 2.625	5 V 8
•	Added V _{OH} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 3 V to 3.6 V	9
•	Added V _{OH} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 3 V to 3.6 V	9
•	Added I _{EE} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 3 V to 3.6 V	10
•	Added I _{CC} specification for T _{PCB} ≤ 105°C in <i>Electrical Characteristics: LVPECL Output, at VCC</i> = 3 V to 3.6 V	10
•	Added Thermal Considerations section	23
•	Added Community Resources section	24
CI	hanges from Revision C (June 2014) to Revision D	Page
•	Added "NOTE" at the beginning of "Applications and Implementation" section	20
•	Changed JEDEC symbol to R _{eJA}	23
CI	hanges from Revision B (Augsut 2011) to Revision C	Page

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Changed data sheet flow and layout to conform with new TI standards. Added the following sections: Application and Implementation; Power Supply Recommendations; Layout, Device and Documentation Support, Mechanical,



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•	Added f_{IN} = 125 MHz, 312.5 MHz for $V_{OUT,\ DIFF,\ PP}$
•	Added Typical values, Max values, and footnotes for 122.88 MHz, 156.25 MHz, and 312.5 MHz test conditions corresponding to Random Additive Jitter in Electrical Characteristics: LVPECL Output, at V _{CC} = 3.0 V to 3.6 V
Cł	hanges from Revision A (May 2010) to Revision B Page
•	Revised description of pin 75
•	Corrected V _{IL} parameter description in <i>Electrical Characteristics</i> table for LVCMOS inputs
•	Added footnote (2) to <i>Electrical Characteristics</i> table for LVPECL Output, V _{CC} = 2.375 V to 2.625 V
•	Added footnote (2) to <i>Electrical Characteristics</i> table for LVPECL Output, V _{CC} = 2.375 V to 2.625 V
•	Changed resistor values in Figure 12
<u>•</u>	Changed resistor values in Figure 16
Cl	hanges from Original (August 2009) to Revision A Page
•	Corrected package designators in orderable device names in the Device Comparison Table4
•	Changed description of INP1, INP0 and INN1, INN0 pins in <i>Pin Descriptions</i> table
•	Changed descriptions of all output pins in <i>Pin Descriptions</i> table

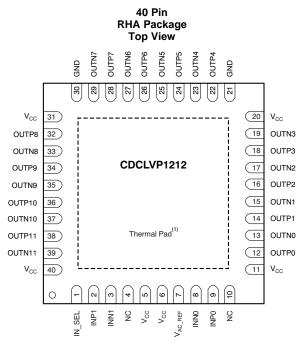


5 Device Comparison Table (1)

PACKAGED DEVICES	T _A	FEATURES		
CDCLVP1212RHAT -40°C to 85°C		40 terminal QFN (RHA) package, small tape and reel		
CDCLVP1212RHAR	-40°C to 85°C	40 terminal QFN (RHA) package, tape and reel		

⁽¹⁾ For the most current specifications and package information, see the Package Option Addendum located at the end of this data sheet or refer to our web site at www.ti.com.

6 Pin Configuration and Functions



(1) Thermal pad must be soldered to ground.

Pin Functions

	1 III T UNCTONS					
P	IN		PULL-UP/			
NAME	NUMBER	TYPE	PULLDOWN	DESCRIPTION		
V _{CC}	5, 6, 11, 20, 31, 40	Power		2.5-V to 3.3-V supplies for the device		
GND	21, 30	Ground		Device grounds		
INP0, INN0	9, 8	Input	_	Differential input pair or single-ended input. Unused input pair can be left floating.		
INP1, INN1	2, 3	Input		Redundant differential input pair or single-ended input. Unused input pair can be left floating.		
OUTP11, OUTN11	38, 39	Output		Differential LVPECL output pair no. 11. Unused output pair can be left floating.		
OUTP10, OUTN10	36, 37	Output	_	Differential LVPECL output pair no. 10. Unused output pair can be left floating.		
OUTP9, OUTN9	34, 35	Output	_	Differential LVPECL output pair no. 9. Unused output pair can be left floating.		
OUTP8, OUTN8	32, 33	Output	_	Differential LVPECL output pair no. 8. Unused output pair can be left floating.		
OUTP7, OUTN7	28, 29	Output	_	Differential LVPECL output pair no. 7. Unused output pair can be left floating.		

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Pin Functions (continued)

PIN			PULL-UP/		
NAME	NUMBER	TYPE	PULLDOWN	DESCRIPTION	
OUTP6, OUTN6	26, 27	Output	_	Differential LVPECL output pair no. 6. Unused output pair can be left floating.	
OUTP5, OUTN5	24, 25	Output	_	Differential LVPECL output pair no. 5. Unused output pair can be left floating.	
OUTP4, OUTN4	22, 23	Output	_	Differential LVPECL output pair no. 4. Unused output pair can be left floating.	
OUTP3, OUTN3	18, 19	Output	_	Differential LVPECL output pair no. 3. Unused output pair can be left floating.	
OUTP2, OUTN2	16, 17	Output	_	Differential LVPECL output pair no. 2. Unused output pair can be left floating.	
OUTP1, OUTN1	14, 15	Output	_	Differential LVPECL output pair no. 1. Unused output pair can be left floating.	
OUTP0 OUTN0	12, 13	Output	_	Differential LVPECL output pair no. 0. Unused output pair can be left floating.	
IN_SEL	1	Input	Pulldown (see Pin Characteristics)	MUX select input for input choice (see Table 1)	
V _{AC_REF}	7	Output	_	Bias voltage output for capacitive coupled inputs. Do not use V_{AC_REF} at $V_{CC} < 3.0$ V. If used, it is recommended to use a 0.1- μ F capacitor to GND on this terminal. The output current is limited to 2 mA.	
NC	4, 10	_	_	Do not connect	

Table 1. Input Selection Table

IN_SEL	ACTIVE CLOCK INPUT
0	INPO, INNO
1	INP1, INN1



7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)

		MIN	MAX	UNIT
V _{CC}	Supply voltage range ⁽²⁾	-0.5	4.6	V
V _{IN}	Input voltage range (3)	-0.5	V _{CC} + 0.5	V
V _{OUT}	Output voltage range (3)	-0.5	V _{CC} + 0.5	V
I _{IN}	Input current		20	mA
I _{OUT}	Output current		50	mA
T _A	Specified free-air temperature range (no airflow)	-40	85	°C
T _{stg}	Storage temperature range	-65	150	°C
TJ	Maximum junction temperature		125	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

			MIN	MAX	UNIT
V _(ESD) ⁽¹⁾	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽²⁾		2000	V

⁽¹⁾ Electrostatic discharge (ESD) to measure device sensitivity and immunity to damage caused by assembly line electrostatic discharges in to the device.

7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	3 1 3 (1			
		MIN	TYP	MAX	UNIT
V_{CC}	Supply voltage	2.375	2.50/3.30	3.60	V
T _A	Ambient temperature	-40		85	°C
T _{PCB}	PCB temperature (measured at thermal pad)			105	°C

7.4 Thermal Information

	THERMAL METRIC ⁽¹⁾⁽²⁾⁽³⁾		VALUE	UNIT
		0 LFM	36.1 ⁽⁴⁾	
$R_{\theta JA}$	Thermal resistance, junction-to-ambient	150 LFM	30.2 ⁽⁴⁾	°C/W
		400 LFM	28.2 (4)	
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	23.7	°C/W	
$R_{\theta JP}^{(5)}$	Thermal resistance, junction-to-pad		3.58	°C/W
ΨЈТ	Junction-to-top characterization parameter	0.5	°C/W	
ΨЈВ	Junction-to-board characterization parame	10.0	°C/W	
R _{0JC(bot)}	Junction-to-case (bottom) thermal resistan	3.8	°C/W	

- (1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report (SPRA953).
- (2) The package thermal resistance is calculated in accordance with JESD 51 and JEDEC 2S2P (high-K board).
- (3) Connected to GND with 16 thermal vias (0.3-mm diameter).
- (4) 4 x 4 vias on Pad

(5) θ_{JP} (junction-to-pad) is used for the QFN package, because the primary heat flow is from the junction to the GND pad of the QFN package.

⁽²⁾ All supply voltages must be supplied simultaneously.

⁽³⁾ The input and output negative voltage ratings may be exceeded if the input and output clamp-current ratings are observed.

⁽²⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.



7.5 Electrical Characteristics: LVCMOS Input

at $V_{CC} = 2.375$ V to 3.6 V and $T_A = -40$ °C to 85°C and $T_{PCB} \le 105$ °C (unless otherwise noted) (1)

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
f _{IN}	Input frequency			200	MHz
V_{th}	Input threshold voltage	External threshold voltage applied to complementary input	1.1	1.8	V
V_{IH}	Input high voltage		$V_{th} + 0.1$	V _{CC}	V
V_{IL}	Input low voltage		0	$V_{th} - 0.1$	V
I _{IH}	Input high current	V _{CC} = 3.6 V, V _{IH} = 3.6 V		40	μΑ
I _{IL}	Input low current	V _{CC} = 3.6 V, V _{IL} = 0 V		-40	μΑ
ΔV/ΔΤ	Input edge rate	20% to 80%	1.5		V/ns
I _{CAP}	Input capacitance			5	pF

⁽¹⁾ Figure 6 and Figure 7 show DC test setup.

7.6 Electrical Characteristics: Differential Input

at $V_{CC} = 2.375$ V to 3.6 V and $T_A = -40$ °C to 85°C and $T_{PCR} \le 105$ °C (unless otherwise noted)⁽¹⁾

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
f _{IN}	Input frequency	Clock input		2000	MHz
	Differential input peak-peak	f _{IN} ≤ 1.5 GHz	0.1	1.5	V
V _{IN, DIFF, PP}	voltage	1.5 GHz ≤ f _{IN} ≤ 2 GHz	0.2	1.5	V
V _{ICM}	Input common-mode level		1.0	V _{CC} - 0.3	V
I _{IH}	Input high current	V _{CC} = 3.6 V, V _{IH} = 3.6 V		40	μΑ
I _{IL}	Input low current	V _{CC} = 3.6 V, V _{IL} = 0 V		-40	μΑ
ΔV/ΔΤ	Input edge rate	20% to 80%	1.5		V/ns
I _{CAP}	Input capacitance			5	pF

⁽¹⁾ Figure 5 and Figure 8 show DC test setup. Figure 9 shows AC test setup.

7.7 Electrical Characteristics: LVPECL Output, At V_{CC} = 2.375 V to 2.625 V

 $T_A = -40$ °C to 85°C and $T_{PCB} \le 105$ °C (unless otherwise noted) ⁽¹⁾

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT	
M	Output high voltage	T _A ≤ 85°C	V _{CC} - 1.26	V _{CC} - 0.9	V	
V_{OH}	Output high voltage	T _{PCB} ≤ 105°C	V _{CC} 1.26	$V_{CC} - 0.83$	V	
M	Output low voltage	T _A ≤ 85°C	V _{CC} – 1.7	V _{CC} – 1.3	V	
V_{OL}	Output low voltage	T _{PCB} ≤ 105°C	V _{CC} – 1.7	V _{CC} – 1.25	V	
M	Differential output peak-peak	f _{IN} ≤ 2 GHz	0.5	1.35	V	
V _{OUT, DIFF, PP}	voltage	f _{IN} = 125 MHz, 312.5 MHz	1.15		V	
V_{AC_REF}	Input bias voltage (2)	I _{AC_REF} = 2 mA	V _{CC} – 1.6	V _{CC} - 1.1	V	
	Propagation delay	$V_{IN, DIFF, PP} = 0.1 V$		550	ps	
t _{PD}	Propagation delay	$V_{IN, DIFF, PP} = 0.3 V$		550	ps	
t _{SK,PP}	Part-to-part skew			150	ps	
t _{SK,O}	Output skew			25	ps	
t _{SK,P}	Pulse skew (with 50% duty cycle input)	Crossing-point-to-crossing-point distortion, f _{OUT} = 100 MHz	-50	50	ps	

⁽¹⁾ Figure 10 and Figure 11 show DC and AC test setup.

⁽²⁾ Internally generated bias voltage (V_{AC_REF}) is for 3.3 V operation only. It is recommended to apply externally generated bias voltage for V _{CC} < 3.0 V.



Electrical Characteristics: LVPECL Output, At $V_{CC} = 2.375 \text{ V}$ to 2.625 V (continued)

.,	C to 85°C and T _{PCB} ≤ 105°C (unless o PARAMETER	TEST CONDITIONS	MIN TYP	MAX	UNIT
		f _{OUT} = 100 MHz, V _{IN,SE} = V _{CC} , V _{th} = 1.25 V, 10 kHz to 20 MHz	0.11		ps, RMS
		$f_{OUT} = 100 \text{ MHz}, V_{IN,SE} = 0.9 \text{ V}, V_{th} = 1.1 \text{ V}, 10 \text{ kHz to } 20 \text{ MHz}$	0.128		ps, RMS
		f _{OUT} = 2 GHz, V _{IN,DIFF,PP} = 0.2 V, V _{ICM} = 1 V, 10 kHz to 20 MHz	0.053		ps, RMS
		$f_{OUT} = 100 \text{ MHz}, V_{IN,DIFF,PP} = 0.15 \text{ V}, \\ V_{ICM} = 1 \text{ V}, 10 \text{ kHz to } 20 \text{ MHz}$	0.093		ps, RMS
		f_{OUT} = 100 MHz, $V_{IN,DIFF,PP}$ = 1 V, V_{ICM} = 1 V, 10 kHz to 20 MHz	0.092		ps, RMS
[†] RJIT		f _{OUT} = 122.88 MHz, ⁽⁴⁾⁽⁵⁾ Square Wave, V _{IN-PP} = 1 V, 12 kHz to 20 MHz	0.057	0.088	ps, RMS
		f _{OUT} = 122.88 MHz, ⁽⁴⁾⁽⁵⁾ Square Wave, V _{IN-PP} = 1 V, 10 kHz to 20 MHz	0.057	0.088	ps, RMS
	Random additive jitter (with 50% duty cycle input) ⁽³⁾	f _{OUT} = 122.88 MHz, ⁽⁴⁾⁽⁵⁾ Square Wave, V _{IN-PP} = 1 V, 1 kHz to 40 MHz	0.086	0.121	ps, RMS
		f _{OUT} = 156.25 MHz, ⁽⁵⁾⁽⁶⁾ Square Wave, V _{IN-PP} = 1 V, 12 kHz to 20 MHz	0.048	0.071	ps, RMS
		f _{OUT} = 156.25 MHz, ⁽⁵⁾⁽⁶⁾ Square Wave, V _{IN-PP} = 1 V, 10 kHz to 20 MHz	0.048	0.071	ps, RMS
		f _{OUT} = 156.25 MHz, ⁽⁵⁾⁽⁶⁾ Square Wave, V _{IN-PP} = 1 V, 1 kHz to 40 MHz	0.068	0.097	ps, RMS
		f _{OUT} = 312.5 MHz, ⁽⁵⁾⁽⁷⁾ Square Wave, V _{IN-PP} = 1 V, 12 kHz to 20 MHz	0.030	0.048	ps, RMS
		f _{OUT} = 312.5 MHz, ⁽⁵⁾⁽⁷⁾ Square Wave, V _{IN-PP} = 1 V, 10 kHz to 20 MHz	0.030	0.048	ps, RMS
		f _{OUT} = 312.5 MHz, ⁽⁵⁾⁽⁷⁾ Square Wave, V _{IN-PP} = 1 V, 1 kHz to 40 MHz	0.045	0.068	ps, RMS
t _R /t _F	Output rise/fall time	20% to 80%		200	ps
1	Cupply internal autrent	Outputs unterminated T _A ≤ 85°C		88	A
IEE	Supply internal current	Outputs unterminated T _{PCB} ≤ 105°C		89	mA
1	Output and internal aupply aurent	All outputs terminated, 50 Ω to V_{CC} – 2 $T_A \le 85^{\circ}C$		468	mA
Icc	Output and internal supply current	internal supply current All outputs terminated, $50 \ \Omega \text{ to V}_{CC} - 2 \\ T_{PCB} \le 105 ^{\circ}\text{C}$			

⁽³⁾ Parameter is specified by characterization. Not tested in production.

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Input source: 122.88-MHz Rohde & Schwarz SMA100A Signal Generator.

Input source RMS Jitter (t_{R,JIT_IN}) and Total RMS Jitter (t_{R,JIT_OUT}) measured using Agilent E5052 Phase Noise Analyzer. Buffer device random additive jitter computed as: t_{R,JIT} = SQRT[(t_{R,JIT_OUT})² - (t_{R,JIT_IN})²].

Input source: 156.25-MHz Rohde & Schwarz SMA100A Signal Generator. (5)

Input source: 312.5-MHz Rohde & Schwarz SMA100A Signal Generator.



7.8 Electrical Characteristics: LVPECL Output, at V_{cc} = 3.0 V to 3.6 V

 $T_A = -40$ °C to 85°C and $T_{PCB} \le 105$ °C (unless otherwise noted) ⁽¹⁾

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
W	Output high voltage	T _A ≤ 85°C	V _{CC} - 1.26	V _{CC} - 0.9	V
V _{OH}	Output high voltage	T _{PCB} ≤ 105°C	V _{CC} – 1.26	V _{CC} - 0.85	V
W	Output law valtage	T _A ≤ 85°C	V _{CC} – 1.7	V _{CC} – 1.3	V
V_{OL}	Output low voltage	T _{PCB} ≤ 105°C	V _{CC} – 1.7	V _{CC} - 1.3	V
V _{OUT, DIFF, PP}	Differential output peak-peak voltage	f _{IN} ≤ 2 GHz	0.65	1.35	V
V_{AC_REF}	Input bias voltage	I _{AC_REF} = 2 mA	V _{CC} – 1.6	V _{CC} - 1.1	V
•	Propagation delay	$V_{IN, DIFF, PP} = 0.1 V$		550	ps
t _{PD}	Fropagation delay	V _{IN, DIFF, PP} = 0.3 V		550	ps
t _{SK,PP}	Part-to-part skew			150	ps
$t_{SK,O}$	Output skew			25	ps
t _{SK,P}	Pulse skew (with 50% duty cycle input)	Crossing-point-to-crossing-point distortion, f _{OUT} = 100 MHz	-50	50	ps



Electrical Characteristics: LVPECL Output, at $V_{cc} = 3.0 \text{ V}$ to 3.6 V (continued)

 $T_{\rm v} = -40^{\circ}$ C to 85°C and $T_{\rm v}$ ₅ < 105°C (unless otherwise noted) (1)

	PARAMETER	TEST CONDITIONS	MIN TYP MAX	UNIT
		f_{OUT} = 100 MHz, $V_{IN,SE}$ = V_{CC} , V_{th} = 1.65 V, 10 kHz to 20 MHz	0.101	ps, RMS
		$f_{OUT} = 100 \text{ MHz}, V_{IN,SE} = 0.9 \text{ V}, V_{th} = 1.1 \text{ V}, 10 \text{ kHz to 20 MHz}$	0.130	ps, RMS
		$f_{OUT} = 2 \text{ GHz}, V_{IN,DIFF,PP} = 0.2 \text{ V}, \\ V_{ICM} = 1 \text{ V}, 10 \text{ kHz to } 20 \text{ MHz}$	0.069	ps, RMS
	Random additive jitter (with 50% duty cycle input) ⁽²⁾	$f_{OUT} = 100 \text{ MHz}, V_{\text{IN,DIFF,PP}} = 0.15 \text{ V}, \\ V_{\text{ICM}} = 1 \text{ V}, 10 \text{ kHz to } 20 \text{ MHz}$	0.094	ps, RMS
		f_{OUT} = 100 MHz, $V_{IN,DIFF,PP}$ = 1 V, V_{ICM} = 1 V, 10 kHz to 20 MHz	0.094	ps, RMS
		f _{OUT} = 122.88 MHz, ⁽³⁾⁽⁴⁾ Square Wave, V _{IN-PP} = 1 V, 12 kHz to 20 MHz	0.057	ps, RMS
		f _{OUT} = 122.88 MHz, ⁽³⁾⁽⁴⁾ Square Wave, V _{IN-PP} = 1 V, 10 kHz to 20 MHz	0.057	ps, RMS
^t RJIT		f _{OUT} = 122.88 MHz, ⁽³⁾⁽⁴⁾ Square Wave, V _{IN-PP} = 1 V, 1 kHz to 40 MHz	0.086	ps, RMS
		f _{OUT} = 156.25 MHz, ⁽⁴⁾⁽⁵⁾ Square Wave, V _{IN-PP} = 1 V, 12 kHz to 20 MHz	0.048	ps, RMS
		f _{OUT} = 156.25 MHz, ⁽⁴⁾⁽⁵⁾ Square Wave, V _{IN-PP} = 1 V, 10 kHz to 20 MHz	0.048	ps, RMS
		f _{OUT} = 156.25 MHz, ⁽⁴⁾⁽⁵⁾ Square Wave, V _{IN-PP} = 1 V, 1 kHz to 40 MHz	0.068	ps, RMS
		f _{OUT} = 312.5 MHz, ⁽⁴⁾⁽⁶⁾ Square Wave, V _{IN-PP} = 1 V, 12 kHz to 20 MHz	0.030	ps, RMS
		f _{OUT} = 312.5 MHz, ⁽⁴⁾⁽⁶⁾ Square Wave, V _{IN-PP} = 1 V, 10 kHz to 20 MHz	0.030	ps, RMS
		f _{OUT} = 312.5 MHz, ⁽⁴⁾⁽⁶⁾ Square Wave, V _{IN-PP} = 1 V, 1 kHz to 40 MHz	0.045	ps, RMS
t _R /t _F	Output rise/fall time	20% to 80%	200	ps
I _{EE}	Supply internal current	Outputs unterminated T _A ≤ 85°C	88	mA
		T _{PCB} ≤ 105°C	89	
1	Output and internal supply current	All outputs terminated, 50 Ω to V_{CC} – 2 $T_A \le 85^{\circ}C$	468	mA
Icc	Output and internal supply current	All outputs terminated, 50Ω to $V_{CC} - 2$ $T_{PCB} \le 105^{\circ}C$		

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 ⁽²⁾ Parameter is specified by characterization. Not tested in production.
 (3) Input source: 122.88-MHz Rohde & Schwarz SMA100A Signal Generator.
 (4) Input source RMS Jitter (t_{R,JIT_IN}) and Total RMS Jitter (t_{R,JIT_OUT}) measured using Agilent E5052 Phase Noise Analyzer. Buffer device random additive jitter computed as: t_{R,JIT} = SQRT[(t_{R,JIT_OUT})² - (t_{R,JIT_IN})²].
 (5) Input source: 156.25-MHz Rohde & Schwarz SMA100A Signal Generator.

⁽⁶⁾ Input source: 312.5-MHz Rohde & Schwarz SMA100A Signal Generator.



7.9 Pin Characteristics

	PARAMETER	MIN	TYP	MAX	UNIT
R _{PULLDOWN}	Input pulldown resistor		150		kΩ

7.10 Timing Requirements

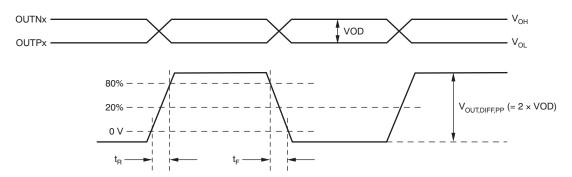
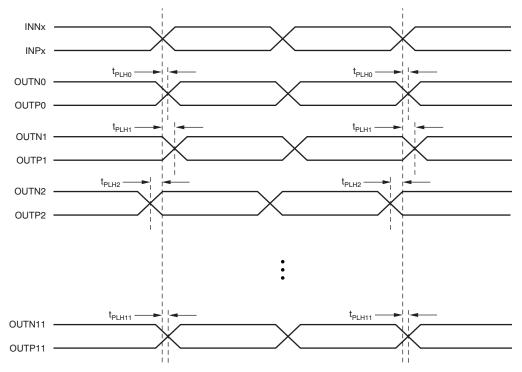


Figure 1. Output Voltage and Rise/Fall Time



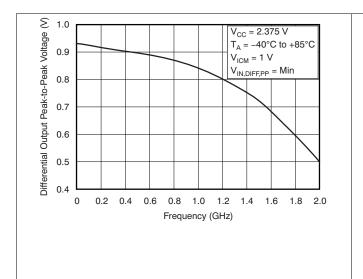
- (1) Output skew is calculated as the greater of the following: As the difference between the fastest and the slowest t_{PLHn} (n = 0, 1, 2....11), or as the difference between the fastest and the slowest t_{PHLn} (n = 0, 1, 2....11).
- (2) Part-to-part skew is calculated as the greater of the following: As the difference between the fastest and the slowest t_{PLHn} (n = 0, 1, 2....11) across multiple devices, or the difference between the fastest and the slowest t_{PHLn} (n = 0, 1, 2....11) across multiple devices.

Figure 2. Output and Part-To-Part Skew



7.11 Typical Characteristics

at $T_A = -40$ °C to 85°C (unless otherwise noted)



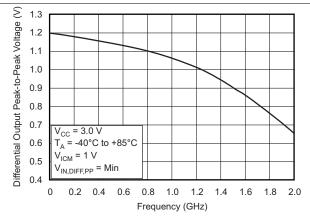


Figure 3. Differential Output Peak-To-Peak Voltage Vs Frequency

Figure 4. Differential Output Peak-To-Peak Voltage Vs Frequency

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8 Parameter Measurement Information

8.1 Test Configurations

This section describes the function of each block for the CDCLVP1212. Figure 6 through Figure 11 illustrate how the device should be set up for a variety of test configurations.

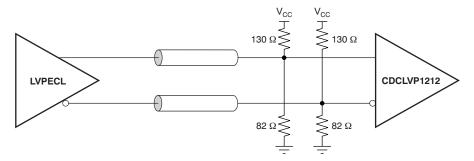


Figure 5. DC-Coupled LVPECL Input During Device Test

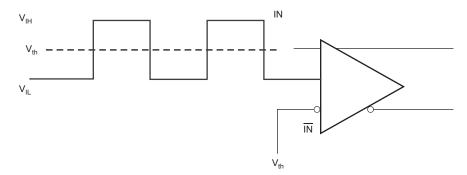


Figure 6. DC-Coupled LVCMOS Input During Device Test

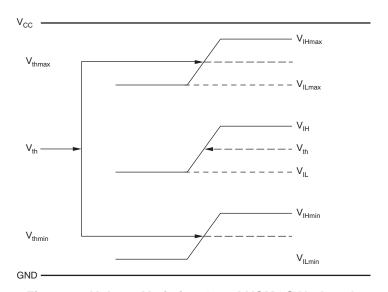


Figure 7. Voltage Variation Over LVCMOS V_{th} Levels

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Test Configurations (continued)

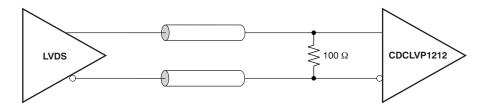


Figure 8. DC-Coupled LVDS Input During Device Test

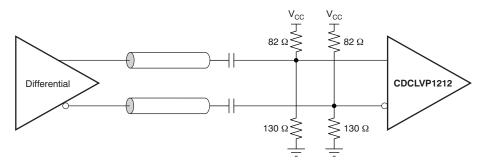


Figure 9. AC-Coupled Differential Input To Device

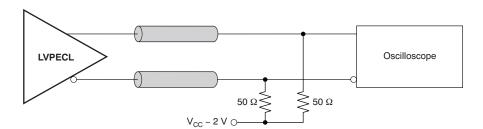


Figure 10. LVPECL Output DC Configuration During Device Test

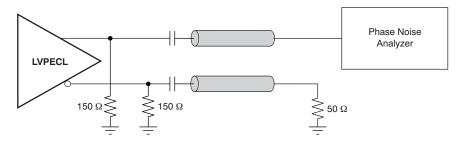


Figure 11. LVPECL Output AC Configuration During Device Test

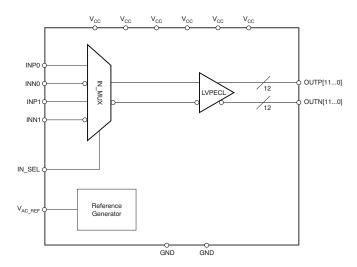


9 Detailed Description

9.1 Overview

The CDCLVP1212 uses an open emitter follower stage for its LVPECL outputs. Therefore, proper output biasing and termination are required to ensure correct operation of the device and to maximize output signal integrity. The proper termination for LVPECL outputs is a 50 Ω to (VCC -2) V, but this DC voltage is not readily available on PCB. Therefore, a Thevenin equivalent circuit is worked out for the LVPECL termination in both direct-coupled (DC) and AC-coupled configurations. These configurations are shown in Figure 12 (a and b) for VCC = 2.5 V and Figure 13 (a and b) for VCC = 3.3 V, respectively. It is recommended to place all resistive components close to either the driver end or the receiver end. If the supply voltage for the driver and receiver is different, AC coupling is required.

9.2 Functional Block Diagram



9.3 Feature Description

The CDCLVP1212 is a low additive jitter universal to LVPECL fan out buffer with 2 selectable inputs. The small package, low output skew, and low additive jitter make for a flexible device in demanding applications.

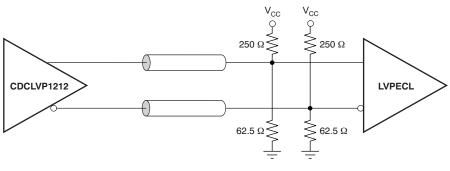
9.4 Device Functional Modes

The two inputs of the CDCLVP1212 are internally muxed together and can be selected via the control pin. Unused inputs and outputs can be left floating to reduce overall component cost. Both AC and DC coupling schemes can be used with the CDCLVP1212 to provide greater system flexibility.

9.4.1 LVPECL Output Termination

The CDCLVP1212 is an open emitter for LVPECL outputs. Therefore, proper biasing and termination are required to ensure correct operation of the device and to minimize signal integrity. The proper termination for LVPECL outputs is a 50 Ω to (V_{CC} –2) V, but this DC voltage is not readily available on PCB. Therefore, a Thevenin equivalent circuit is worked out for the LVPECL termination in both direct-coupled (DC) and AC-coupled configurations. These configurations are shown in Figure 12 (a and b) for V_{CC} = 2.5 V and Figure 13 (a and b) for V_{CC} = 3.3 V, respectively. It is recommended to place all resistive components close to either the driver end or the receiver end. If the supply voltage for the driver and receiver is different, AC coupling is required.





(a) Output DC Termination

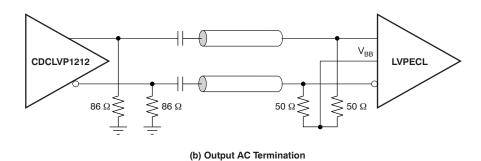
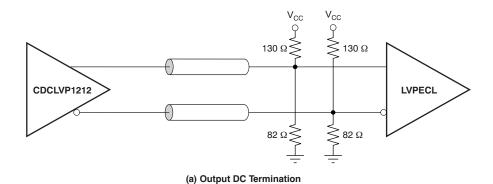


Figure 12. LVPECL Output DC and AC Termination For $V_{CC} = 2.5 \text{ V}$





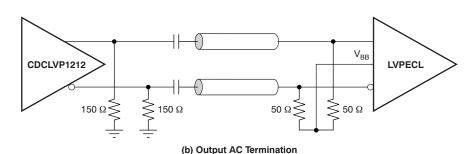


Figure 13. LVPECL Output DC and AC Termination For $V_{CC} = 3.3 \text{ V}$

9.4.2 Input Termination

The CDCLVP1212 inputs can be interfaced with LVPECL, LVDS, or LVCMOS drivers. Figure 14 illustrates how to DC couple an LVCMOS input to the CDCLVP1212. The series resistance (R_S) should be placed close to the LVCMOS driver; its value is calculated as the difference between the transmission line impedance and the driver output impedance.

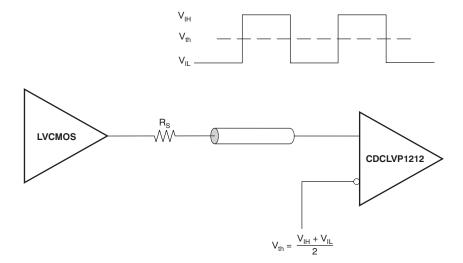


Figure 14. DC-Coupled LVCMOS Input to CDCLVP1212

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Figure 15 shows how to DC couple LVDS inputs to the CDCLVP1212. Figure 16 and Figure 17 describe the method of DC coupling LVPECL inputs to the CDCLVP1212 for $V_{CC} = 2.5 \text{ V}$ and $V_{CC} = 3.3 \text{ V}$, respectively.

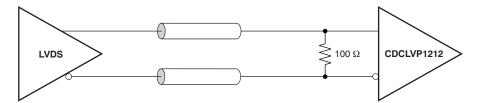


Figure 15. DC-Coupled LVDS Inputs to CDCLVP1212

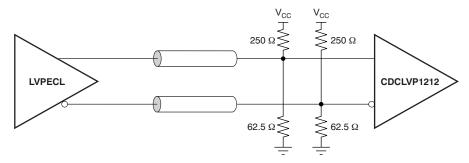


Figure 16. DC-Coupled LVPECL Inputs to CDCLVP1212 ($V_{CC} = 2.5 \text{ V}$)

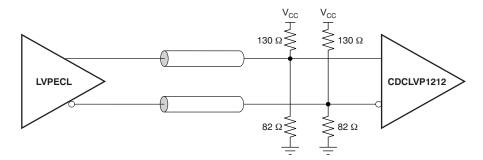


Figure 17. DC-Coupled LVPECL Inputs to CDCLVP1212 ($V_{CC} = 3.3 \text{ V}$)



Figure 18 and Figure 19 show the technique of AC coupling differential inputs to the CDCLVP1212 for V_{CC} = 2.5 V and V_{CC} = 3.3 V, respectively. It is recommended to place all resistive components close to either the driver end or the receiver end. If the supply voltages of the driver and receiver are different, AC coupling is required.

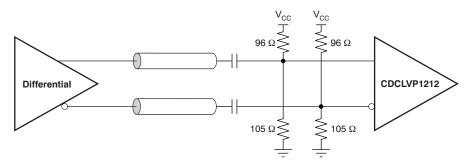


Figure 18. AC-Coupled Differential Inputs to CDCLVP1212 ($V_{CC} = 2.5 \text{ V}$)

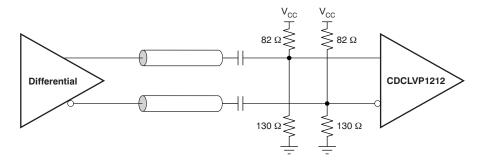


Figure 19. AC-Coupled Differential Inputs to CDCLVP1212 ($V_{CC} = 3.3 \text{ V}$)



10 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

The CDCLVP1212 is a low additive jitter LVPECL fanout buffer that can generate four copies of two selectable LVPECL, LVDS, or LVCMOS inputs. The CDCLVP1212 can accept reference clock frequencies up to 2 GHz while providing low output skew.

10.2 Typical Application

10.2.1 Fanout Buffer for Line Card Application

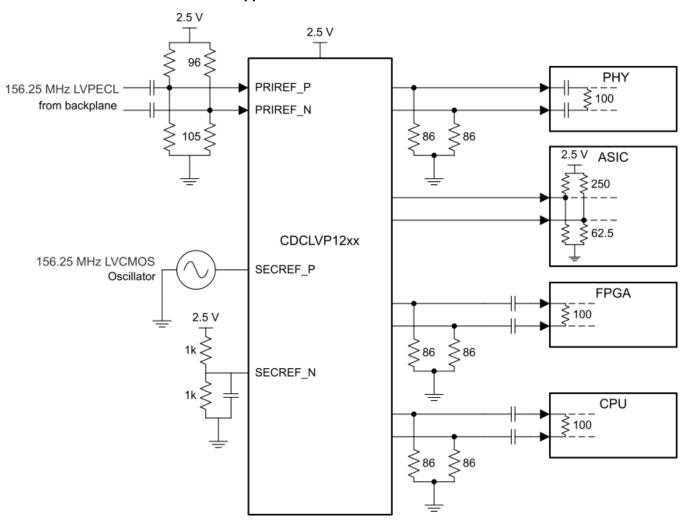


Figure 20. CDCLVP1212 Block Diagram

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Typical Application (continued)

10.2.1.1 Design Requirements

The CDCLVP1212 shown in Figure 20 is configured to be able to select two inputs, a 156.25-MHz LVPECL clock from the backplane, or a secondary 156.25-MHz LVCMOS 2.5-V oscillator. Either signal can be then fanned out to desired devices, as shown.

The configuration example is driving 4 LVPECL receivers in a line card application with the following properties:

- The PHY device has internal AC coupling and appropriate termination and biasing. The CDCLVP1212 will need to be provided with 86-Ω emitter resistors near the driver for proper operation.
- The ASIC is capable of DC coupling with a 2.5-V LVPECL driver such as the CDCLVP1212. This ASIC features internal termination so no additional components are needed.
- The FPGA requires external AC coupling but has internal termination. Again, 86-Ω emitter resistors are placed near the CDCLVP1212 and 0.1-uF capacitors are placed to provide AC coupling. Similarly, the CPU is internally terminated and requires external AC coupling capacitors.

10.2.1.2 Detailed Design Procedure

Refer to *Input Termination* for proper input terminations, dependent on single ended or differential inputs.

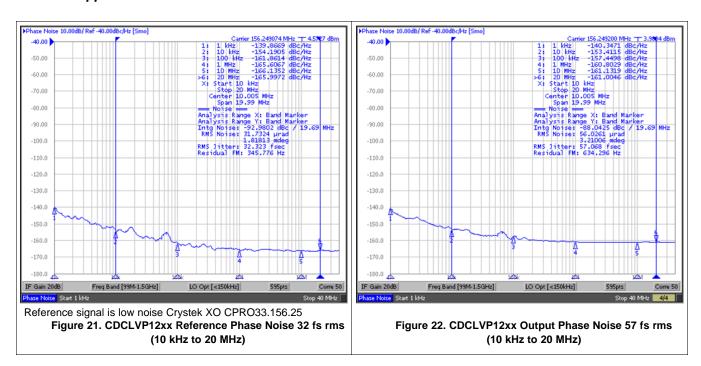
Refer to LVPECL Output Termination for output termination schemes depending on the receiver application.

Unused outputs can be left floating.

In this example, the PHY, ASIC, and FPGA/CPU require different schemes. Power supply filtering and bypassing is critical for low noise applications.

See *Power Supply Recommendations* for recommended filtering techniques. A reference layout is provided on the CDCLVP1212 Evaluation Module at SCAU036.

10.2.1.3 Application Curves



The low additive noise of the CDCLVP12xx can be shown in this line card application. The low noise 156.25-MHz XO with 32-fs RMS jitter drives the CDCLVP12xx, resulting in 57-fs RMS when integrated from 10 kHz to 20 MHz. The resultant additive jitter is a low 47-fs RMS for this configuration.

Product Folder Links: CDCLVP1212



11 Power Supply Recommendations

High-performance clock buffers are sensitive to noise on the power supply, which can dramatically increase the additive jitter of the buffer. Thus, it is essential to reduce noise from the system power supply, especially when jitter or phase noise is very critical to applications.

Filter capacitors are used to eliminate the low-frequency noise from the power supply, where the bypass capacitors provide the very low impedance path for high-frequency noise and guard the power-supply system against the induced fluctuations. These bypass capacitors also provide instantaneous current surges as required by the device and should have low equivalent series resistance (ESR). To properly use the bypass capacitors, they must be placed very close to the power-supply terminals and laid out with short loops to minimize inductance. It is recommended to add as many high-frequency (for example, 0.1 μ F) bypass capacitors as there are supply terminals in the package. It is recommended, but not required, to insert a ferrite bead between the board power supply and the chip power supply that isolates the high-frequency switching noises generated by the clock driver; these beads prevent the switching noise from leaking into the board supply. It is imperative to choose an appropriate ferrite bead with very low DC resistance in order to provide adequate isolation between the board supply and the chip supply, as well as to maintain a voltage at the supply terminals that is greater than the minimum voltage required for proper operation.

Figure 23 illustrates this recommended power-supply decoupling method.

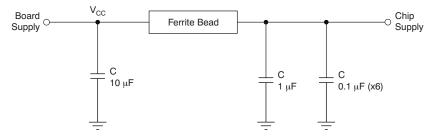


Figure 23. Power-Supply Decoupling

11.1 Thermal Management

Power consumption of the CDCLVP1212 can be high enough to require attention to thermal management. For reliability and performance reasons, the die temperature should be limited to a maximum of 125°C. That is, as an estimate, ambient temperature (T_A) plus device power consumption times $R_{\theta,JA}$ should not exceed 125°C.

The device package has an exposed pad that provides the primary heat removal path to the printed circuit board (PCB). To maximize the heat dissipation from the package, a thermal landing pattern including multiple vias to a ground plane must be incorporated into the PCB within the footprint of the package. The exposed pad must be soldered down to ensure adequate heat conduction out of the package. Figure 24 shows a recommended land and via pattern.



12 Layout

12.1 Layout Guidelines

Power consumption of the CDCLVP1212 can be high enough to require attention to thermal management. For reliability and performance reasons, the die temperature should be limited to a maximum of 125°C. That is, as an estimate, ambient temperature (TA) plus device power consumption times should not exceed 125°C.

The device package has an exposed pad that provides the primary heat removal path to the printed circuit board (PCB). To maximize the heat dissipation from the package, a thermal landing pattern including multiple vias to a ground plane must be incorporated into the PCB within the footprint of the package. The exposed pad must be soldered down to ensure adequate heat conduction out of the package. Figure 24 shows a recommended land and via pattern.

12.2 Layout Example

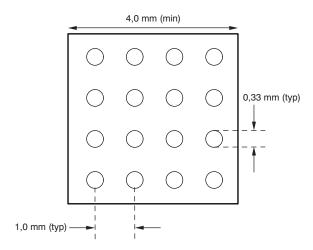


Figure 24. Recommended PCB Layout

12.3 Thermal Considerations

The CDCLVP1212 supports high temperatures on the printed circuit board (PCB) measured at the thermal pad. The system designer needs to ensure that the maximum junction temperature is not exceeded. Ψ_{jb} can allow the system designer to measure the board temperature with a fine gauge thermocouple and back calculate the junction temperature using Equation 1. Note that Ψ_{jb} is close to R θ JB as 75% to 95% of the heat of a device is dissipated by the PCB. For further information, refer to SPRA953 and SLUA566.

$$T_{\text{junction}} = T_{\text{PCB}} + (\Psi_{\text{jb}} \times \text{Power}) \tag{1}$$

Example:

Calculation of the junction-lead temperature with a 4-layer JEDEC test board using four thermal vias:

 $T_{PCB} = 105^{\circ}C$

 $\Psi_{ib} = 10.0$ °C/W

Power_{inclTerm} = I_{max} x V_{max} = 516 mA x 3.6 V = 1857.6 mW (max power consumption including termination resistors)

 $Power_{exclTerm} = 1404.6 \text{ mW}$ (max power consumption excluding termination resistors; see SLYT127 for further details)

 $\Delta T_{Junction} = \Psi_{jb} \times Power_{exclTerm} = 10.0^{\circ}C/W \times 1404.6 \text{ mW} = 14.05^{\circ}C$

 $T_{Junction} = \Delta T_{Junction} + T_{Chassis} = 14.05^{\circ}C + 105^{\circ}C = 119^{\circ}C$ (the maximum junction temperature of 125°C is not violated)

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13 Device and Documentation Support

13.1 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

13.2 Trademarks

E2E is a trademark of Texas Instruments.

All other trademarks are the property of their respective owners.

13.3 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

13.4 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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www.ti.com 10-Nov-2025

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
CDCLVP1212RHAR	Active	Production	VQFN (RHA) 40	2500 LARGE T&R	Yes	NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CDCLVP 1212
CDCLVP1212RHAR.B	Active	Production	VQFN (RHA) 40	2500 LARGE T&R	Yes	NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CDCLVP 1212
CDCLVP1212RHAT	Active	Production	VQFN (RHA) 40	250 SMALL T&R	Yes	NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CDCLVP 1212
CDCLVP1212RHAT.B	Active	Production	VQFN (RHA) 40	250 SMALL T&R	Yes	NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CDCLVP 1212

⁽¹⁾ Status: For more details on status, see our product life cycle.

- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.



PACKAGE OPTION ADDENDUM

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	` '	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CDCLVP1212RHAR	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2

PACKAGE MATERIALS INFORMATION

www.ti.com 5-Dec-2023



*All dimensions are nominal

	Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)	
ı	CDCLVP1212RHAR	VQFN	RHA	40	2500	350.0	350.0	43.0	

6 x 6, 0.5 mm pitch

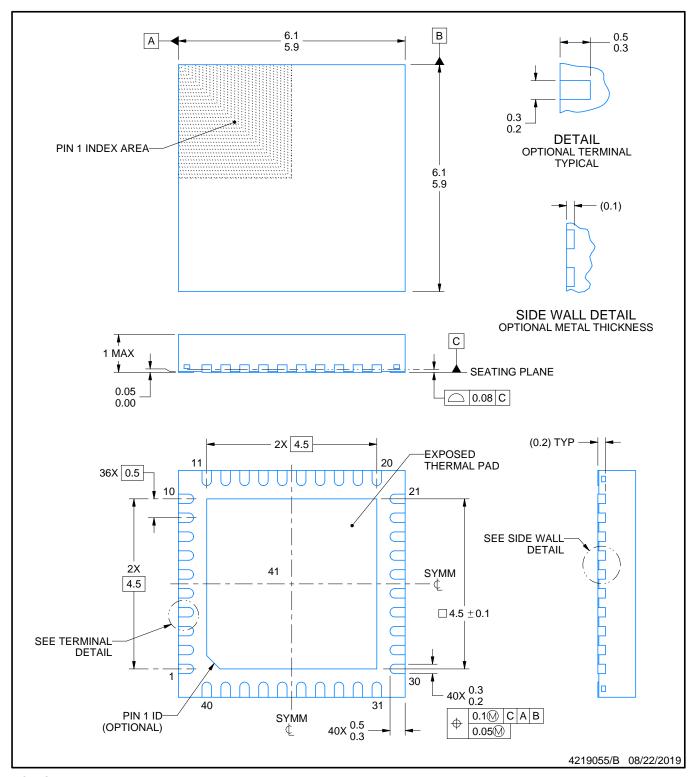
PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





PLASTIC QUAD FLATPACK - NO LEAD

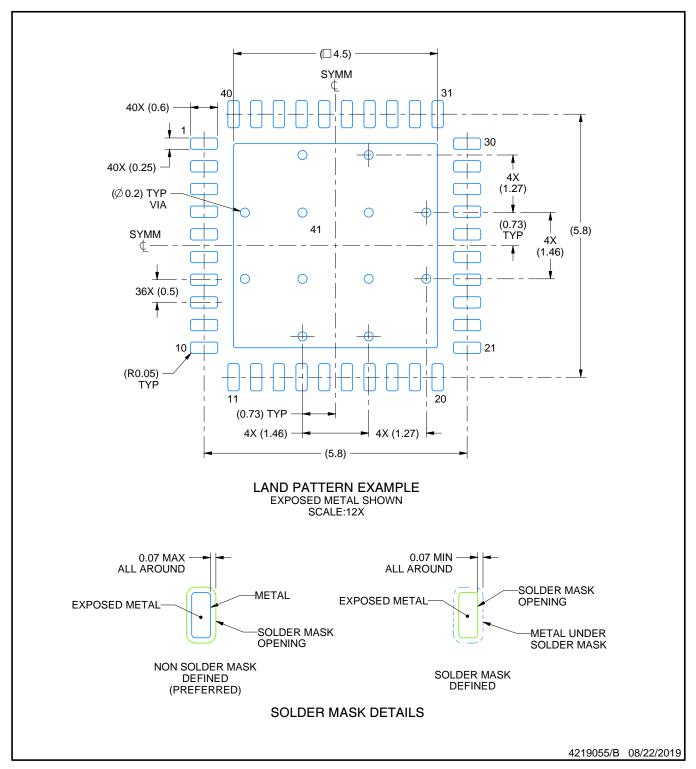


NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK - NO LEAD

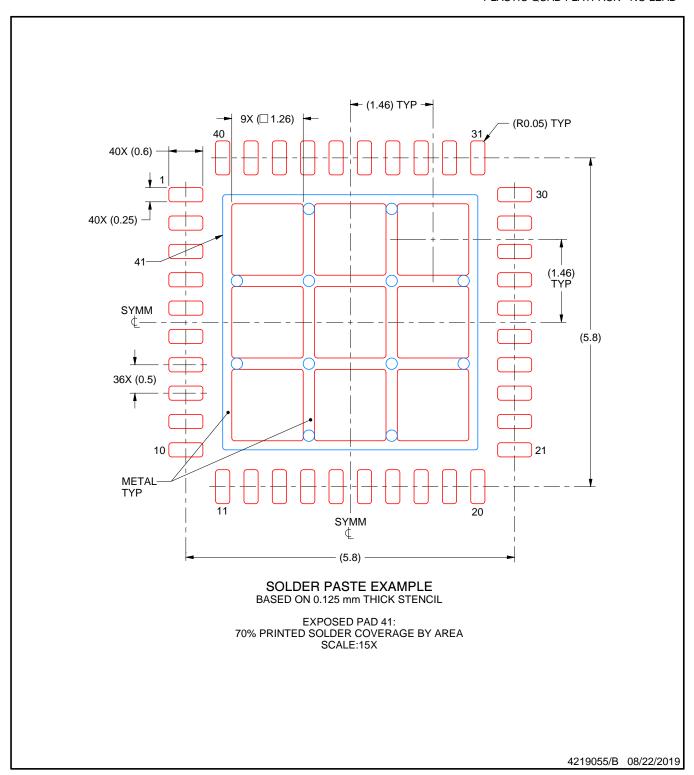


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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Last updated 10/2025