











CDCLVP1102

SCAS884D - AUGUST 2009 - REVISED DECEMBER 2015

CDCLVP1102 Two-LVPECL Output, High-Performance Clock Buffer

1 Features

- 1:2 Differential Buffer
- Single Clock Input
- Universal Inputs Can Accept LVPECL, LVDS, LVCMOS/LVTTL
- Two LVPECL Outputs
- Maximum Clock Frequency: 2 GHz
- Maximum Core Current Consumption: 33 mA
- Very Low Additive Jitter: <100 fs, RMS in 10-kHz to 20-MHz Offset Range
- 2.375-V to 3.6-V Device Power Supply
- Maximum Propagation Delay: 450 ps
- Maximum Output Skew: 10 ps
- LVPECL Reference Voltage, V_{AC_REF}, Available for Capacitive-Coupled Inputs
- Industrial Temperature Range: –40°C to 85°C
- Supports 105°C PCB Temperature (Measured at Thermal Pad)
- Available in 3-mm x 3-mm QFN-16 (RGT) Package
- ESD Protection Exceeds 2 kV (HBM)

2 Applications

- Wireless Communications
- Telecommunications/Networking
- Medical Imaging
- Test and Measurement Equipment

3 Description

The CDCLVP1102 is a highly versatile, low additive jitter buffer that can generate two copies of LVPECL clock outputs from one LVPECL, LVDS, or LVCMOS input for a variety of communication applications. It has a maximum clock frequency up to 2 GHz. The overall additive jitter performance is less than 0.1 ps, RMS from 10 kHz to 20 MHz, and overall output skew is as low as 10 ps, making the device a perfect choice for use in demanding applications.

The CDCLVP1102 clock buffer distributes a single clock input (IN) to two pairs of differential LVPECL clock outputs (OUT0, OUT1) with minimum skew for clock distribution. The inputs can be LVPECL, LVDS, or LVCMOS/LVTTL.

The CDCLVP1102 is specifically designed for driving $50-\Omega$ transmission lines. When driving the inputs in single-ended mode, the LVPECL bias voltage (V_{AC_REF}) should be applied to the unused negative input pin. However, for high-speed performance up to 2 GHz, differential mode is strongly recommended.

The CDCLVP1102 is characterized for operation from –40°C to 85°C and is available in a QFN-16, 3-mm × 3-mm package.

Device Information⁽¹⁾

PART NUMBER	PACKAGE BODY SIZE (NO	
CDCLVP1102	QFN (16)	3.00 mm × 3.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Simplified Schematic

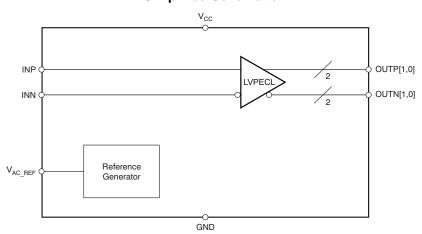




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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Ci	nanges from Revision C (January 2011) to Revision D	Page
•	Added support for 105°C thermal pad temperature	1
•	Added Device Information and Pin Configuration and Functions sections, ESD Rating table, Feature Description, Device Functional Modes, Application and Implementation, Power Supply Recommendations, Layout, Device and Documentation Support, and Mechanical, Packaging, and Orderable Information sections	1
•	Added V_{OH} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 2.375 \text{ V}$ to 2.625 V	6
•	Added V_{OL} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 2.375$ V to 2.625 V	6
•	Added I_{EE} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 2.375 \text{ V}$ to 2.625 V .	6
•	Added I_{CC} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 2.375$ V to 2.625 V .	6
•	Added V_{OH} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 3 \text{ V}$ to 3.6 V	<mark>7</mark>
•	Added V_{OL} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 3 \text{ V}$ to 3.6 V	<mark>7</mark>
•	Added I_{EE} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 3$ V to 3.6 V	<mark>7</mark>
•	Added I_{CC} specification for $T_{PCB} \le 105^{\circ}C$ in Electrical Characteristics: LVPECL Output, at $V_{CC} = 3$ V to 3.6 V	7

CI	hanges from Revision B (May 2010) to Revision C	Page
•	Revised description of pin 8	3
•	Corrected V _{IL} parameter description in <i>Electrical Characteristics</i> table for LVCMOS inputs	5
•	Added footnote (2) to <i>Electrical Characteristics</i> table for LVPECL Output, V _{CC} = 2.375 V to 2.625 V	6
•	Changed recommended resistor values in Figure 12(a)	13
•	Changed resistor values in Figure 16	15
•	Changed resistor values in Figure 17	15

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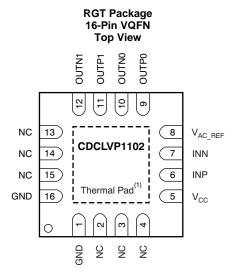


Cł	nanges from Revision A (October 2009) to Revision B	Page
•	Changed description of OUTN1 and OUTN0 pins in Pin Functions table	3

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5 Pin Configuration and Functions



(1) Thermal pad must be soldered to ground.

Pin Functions

PI	N	TVDE	DECODINE		
NAME NUMBER		TYPE	DESCRIPTION		
GND	1, 16	Ground	Device ground		
INP, INN	6, 7	Input	Differential input pair or single-ended input		
OUTP1, OUTN1	11, 12	Output	Differential LVPECL output pair no. 1. Unused output pair can be left floating.		
OUTP0 OUTN0	9, 10	Output	Differential LVPECL output pair no. 0. Unused output pair can be left floating.		
V _{AC_REF}	8	Output	Bias voltage output for capacitive-coupled input pair no. 0. Do not use V_{AC_REF} at $V_{CC} < 3.0$ V. If used, it is recommended to use a 0.1- μ F capacitor to GND on this pin. The output current is limited to 2 mA.		
V _{CC}	5	Power	2.5-V/3.3-V supply for the device		
NC	2, 3, 4, 13, 14, 15	_	Do not connect		



6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
V _{CC}	Supply voltage range ⁽²⁾	-0.5	4.6	V
V _{IN}	Input voltage range (3)	-0.5	V _{CC} + 0.5	V
V _{OUT}	Output voltage range (3)	-0.5	V _{CC} + 0.5	V
I _{IN}	Input current		20	mA
I _{OUT}	Output current		50	mA
T _A	Specified free-air temperature (no airflow)	-40	85	°C
TJ	Maximum junction temperature		125	°C
T _{stg}	Storage temperature	-65	150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

		VALUE	UNIT
V _(ESD) Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	٧

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_{CC}	Supply voltage	2.375	2.50/3.30	3.60	V
T_A	Ambient temperature	-40		85	°C
T _{PCB}	PCB temperature (measured at thermal pad)			105	°C

6.4 Thermal Information

		CDCLVP1102		
	THERMAL METRIC ⁽¹⁾⁽²⁾⁽³⁾	RGT (VQFN)	UNIT	
			16 PINS	
		0 LFM ⁽⁴⁾	51.8	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	150 LFM ⁽⁴⁾	45 °C/	°C/W
	4	400 LFM ⁽⁴⁾	40.8	
R _{θJC(to}	Junction-to-case (top) thermal resistance		61.3	°C/W
$R_{\theta JB}^{(5)}$	Junction-to-board thermal resistance		21.7	°C/W
ΨЈТ	Junction-to-top characterization parameter		1.5	°C/W
ψ_{JB}	Junction-to-board characterization parameter		21.7	°C/W
$R_{\theta JC(b}$ ot)	Junction-to-case (bottom) thermal resistance		6.3	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report (SPRA953).

⁽²⁾ All supply voltages must be supplied simultaneously.

⁽³⁾ The input and output negative voltage ratings may be exceeded if the input and output clamp-current ratings are observed.

⁽²⁾ The package thermal resistance is calculated in accordance with JESD 51 and JEDEC 2S2P (high-K board).

⁽³⁾ Connected to GND with four thermal vias (0.3-mm diameter).

^{4) 2} x 2 vias on Pad

⁽⁵⁾ θ_{JP} (junction-to-pad) is used for the QFN package, because the primary heat flow is from the junction to the GND pad of the QFN package.



6.5 Electrical Characteristics: LVCMOS Input

at $V_{CC} = 2.375$ V to 3.6 V and $T_A = -40$ °C to 85°C and $T_{PCB} \le 105$ °C (unless otherwise noted)⁽¹⁾

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
f _{IN}	Input frequency			200	MHz
V_{th}	Input threshold voltage	External threshold voltage applied to complementary input	1.1	1.8	٧
V _{IH}	Input high voltage		V _{th} + 0.1	V _{CC}	V
V_{IL}	Input low voltage		0	$V_{th} - 0.1$	V
I _{IH}	Input high current	$V_{CC} = 3.6 \text{ V}, V_{IH} = 3.6 \text{ V}$		40	μΑ
I _{IL}	Input low current	$V_{CC} = 3.6 \text{ V}, V_{IL} = 0 \text{ V}$		-40	μΑ
ΔV/ΔΤ	Input edge rate	20% to 80%	1.5		V/ns
I _{CAP}	Input capacitance			5	pF

⁽¹⁾ Figure 3 and Figure 4 show DC test setup.

6.6 Electrical Characteristics: Differential Input

at V_{CC} = 2.375 V to 3.6 V and T_A = -40°C to 85°C and $T_{PCB} \le 105$ °C (unless otherwise noted)⁽¹⁾

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
f _{IN}	Input frequency	Clock input		2000	MHz
	Differential inner and analysis land	f _{IN} ≤ 1.5 GHz	0.1	1.5	V
V _{IN, DIFF, PP}	Differential input peak-peak voltage	1.5 GHz ≤ f _{IN} ≤ 2 GHz	0.2	1.5	V
V _{ICM}	Input common-mode level		1	V _{CC} - 0.3	V
I _{IH}	Input high current	$V_{CC} = 3.6 \text{ V}, V_{IH} = 3.6 \text{ V}$		40	μΑ
I _{IL}	Input low current	$V_{CC} = 3.6 \text{ V}, V_{IL} = 0 \text{ V}$		-40	μΑ
ΔV/ΔΤ	Input edge rate	20% to 80%	1.5		V/ns
I _{CAP}	Input capacitance			5	pF

⁽¹⁾ Figure 5 and Figure 6 show DC test setup. Figure 7 shows AC test setup.

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6.7 Electrical Characteristics: LVPECL Output

at $V_{CC} = 2.375$ V to 2.625 V and $T_A = -40$ °C to 85°C and $T_{PCB} \le 105$ °C (unless otherwise noted)⁽¹⁾

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
	Output high coaltains	T _A ≤ 85°C	V _{CC} - 1.26	V _{CC} - 0.9	V
V _{OH}	Output high voltage	T _{PCB} ≤ 105°C	V _{CC} - 1.26	V _{CC} - 0.83	V
V	Outrout law valtage	T _A ≤ 85°C	V _{CC} - 1.7	V _{CC} – 1.3	V
V_{OL}	Output low voltage	T _{PCB} ≤ 105°C	V _{CC} - 1.7	V _{CC} – 1.25	V
V _{OUT, DIFF, PP}	Differential output peak-peak voltage	f _{IN} ≤ 2 GHz	0.5	1.35	V
V _{AC_REF}	Input bias voltage (2)	I _{AC_REF} = 2 mA	V _{CC} – 1.6	V _{CC} - 1.1	V
+	Propagation delay	V _{IN, DIFF, PP} = 0.1 V		450	ps
t _{PD}	Fropagation delay	$V_{IN, DIFF, PP} = 0.3 \text{ V}$		450	ps
t _{SK,PP}	Part-to-part skew			100	ps
t _{SK,O}	Output skew			10	ps
t _{SK,P}	Pulse skew (with 50% duty cycle input)	Crossing-point-to-crossing-point distortion, f _{OUT} = 100 MHz	-50	50	ps
		$f_{OUT} = 100 \text{ MHz}, V_{IN,SE} = V_{CC}, V_{th} = 1.25 \text{ V}, 10 \text{ kHz to 20 MHz}$		0.089	ps, RMS
		f _{OUT} = 100 MHz, V _{IN,SE} = 0.9 V, V _{th} = 1.1 V, 10 kHz to 20 MHz		0.093	ps, RMS
t _{RJIT}	Random additive jitter (with 50% duty cycle input)		0.037	ps, RMS	
		f _{OUT} = 100 MHz, V _{IN,DIFF,PP} = 0.15 V, V _{ICM} = 1 V, 10 kHz to 20 MHz		0.094	ps, RMS
		f _{OUT} = 100 MHz, V _{IN,DIFF,PP} = 1 V, V _{ICM} = 1 V, 10 kHz to 20 MHz		0.091	ps, RMS
t _R /t _F	Output rise/fall time	20% to 80%		200	ps
	Our about a series of	Outputs unterminated, T _A ≤ 85°C		33	mA
I _{EE}	Supply internal current	Outputs unterminated, T _{PCB} ≤ 105°C		450 450 100 10 50 0.089 0.093 0.037 0.094 0.091	mA
1	Output and internal cumply current	All outputs terminated, 50 Ω to $V_{CC}-2$, $T_A \le 85^{\circ}C$		100	mA
Icc	Output and internal supply current	All outputs terminated, 50 Ω to $V_{CC}-2$, $T_{PCB} \le 105^{\circ}C$		105	mA

 ⁽¹⁾ Figure 8 and Figure 9 show DC and AC test setup.
 (2) Internally generated bias voltage (V_{AC_REF}) is for 3.3-V operation only. It is recommended to apply externally generated bias voltage for V_{CC} < 3.0 V.



6.8 Electrical Characteristics: LVPECL Output

at $V_{CC} = 3.0 \text{ V}$ to 3.6 V and $T_A = -40 ^{\circ}\text{C}$ to $85 ^{\circ}\text{C}$ and $T_{PCB} \le 105 ^{\circ}\text{C}$ (unless otherwise noted) $^{(1)}$

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT	
	0	T _A ≤ 85°C	V _{CC} - 1.26	V _{CC} - 0.9	V	
V_{OH}	Output high voltage	T _{PCB} ≤ 105°C	V _{CC} - 1.26	V _{CC} - 0.85	V	
	Outrot laws laws	T _A ≤ 85°C	V _{CC} - 1.7	V _{CC} - 1.3	V	
V_{OL}	Output low voltage	T _{PCB} ≤ 105°C	V _{CC} - 1.7	V _{CC} - 1.3	V	
V _{OUT, DIFF, PP}	Differential output peak-peak voltage	f _{IN} ≤ 2 GHz	0.65	1.35	V	
V _{AC_REF}	Input bias voltage	I _{AC_REF} = 2 mA	V _{CC} - 1.6	V _{CC} - 1.1	V	
	Drangation dalay	V _{IN, DIFF, PP} = 0.1 V		450	ps	
t _{PD}	Propagation delay	V _{IN, DIFF, PP} = 0.3 V		450	ps	
t _{SK,PP}	Part-to-part skew			100	ps	
t _{SK,O}	Output skew			10	ps	
t _{SK,P}	Pulse skew (with 50% duty cycle input)	Crossing-point-to-crossing-point distortion, f _{OUT} = 100 MHz	-50	50	ps	
		$f_{OUT} = 100 \text{ MHz}, V_{IN,SE} = V_{CC}, V_{th} = 1.65 \text{ V}, 10 \text{ kHz to 20 MHz}$		0.081		
		f _{OUT} = 100 MHz, V _{IN,SE} = 0.9 V, V _{th} = 1.1 V, 10 kHz to 20 MHz		0.097		
t _{RJIT}	Random additive jitter (with 50%	Random additive jitter (with 50% duty cycle input)	0.050	ps, RMS		
	daty cycle input)		0.098			
		f _{OUT} = 100 MHz, V _{IN,DIFF,PP} = 1 V, V _{ICM} = 1 V, 10 kHz to 20 MHz	0.095			
t _R /t _F	Output rise/fall time	20% to 80%		200	ps	
	Completing and account	Outputs unterminated, $T_A \le 85^{\circ}C$		33	mA	
I _{EE}	Supply internal current	Outputs unterminated, T _{PCB} ≤ 105°C		34	mA	
	Output and internal aunaly aurest	All outputs terminated, 50 Ω to $V_{CC}-2$, $T_A \le 85^{\circ}C$	buts terminated, 50 Ω to 2,		mA	
I _{cc}	Output and internal supply current	All outputs terminated, 50 Ω to $V_{CC}-2$, $T_{PCB} \le 105^{\circ}C$		10		

⁽¹⁾ Figure 8 and Figure 9 show DC and AC test setup.

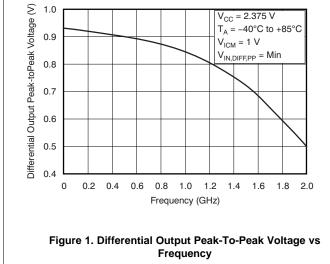
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6.9 Typical Characteristics

at $T_A = -40$ °C to 85°C (unless otherwise noted)



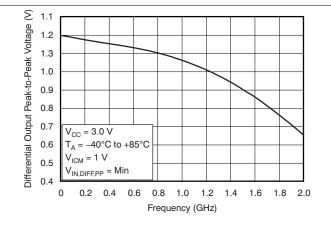


Figure 2. Differential Output Peak-To-Peak Voltage vs Frequency



7 Parameter Measurement Information

7.1 Test Configurations

This section describes the function of each block for the CDCLVP1102. Figure 3 through Figure 9 illustrate how the device should be setup for a variety of test configurations.

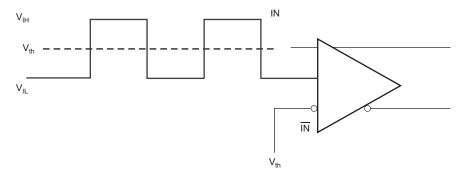


Figure 3. DC-Coupled LVCMOS Input During Device Test

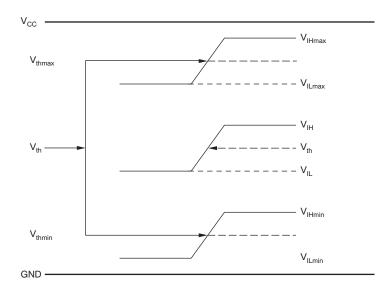


Figure 4. V_{th} Variation Over LVCMOS Levels

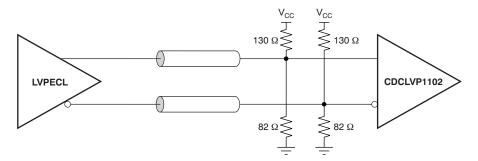


Figure 5. DC-Coupled LVPECL Input During Device Test



Test Configurations (continued)

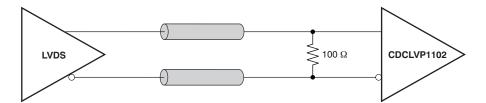


Figure 6. DC-Coupled LVDS Input During Device Test

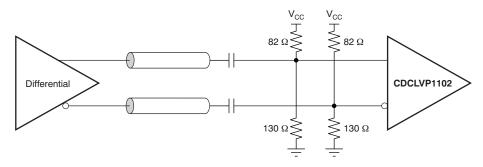


Figure 7. AC-Coupled Differential Input to Device

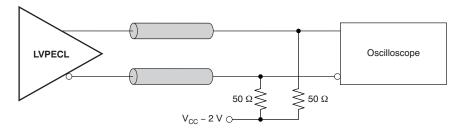


Figure 8. LVPECL Output DC Configuration During Device Test

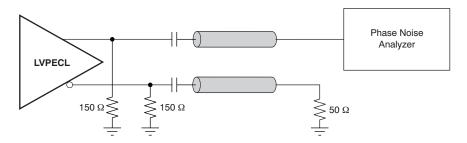


Figure 9. LVPECL Output AC Configuration During Device Test



Test Configurations (continued)

Figure 10 shows the output voltage and rise/fall time. Output and part-to-part skew are shown in Figure 11.

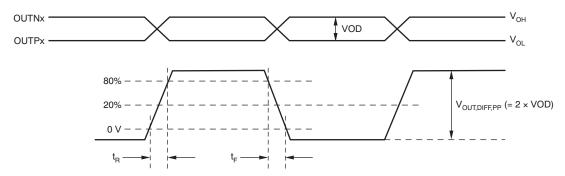
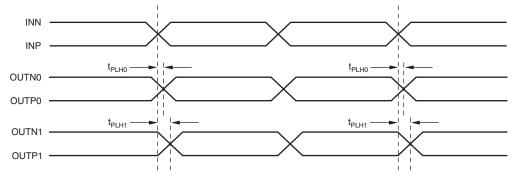


Figure 10. Output Voltage and Rise/Fall Time



- (1) Output skew is calculated as the greater of the following: As the difference between the fastest and the slowest t_{PLHn} (n = 0, 1), or as the difference between the fastest and the slowest t_{PHLn} (n = 0, 1).
- (2) Part-to-part skew is calculated as the greater of the following: As the difference between the fastest and the slowest t_{PLHn} (n = 0, 1) across multiple devices, or the difference between the fastest and the slowest t_{PHLn} (n = 0, 1) across multiple devices.

Figure 11. Output and Part-to-Part Skew

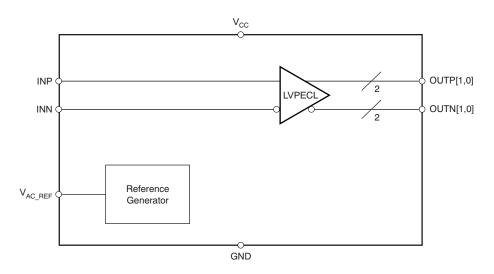


8 Detailed Description

8.1 Overview

The CDCLVP1102 is an open emitter for LVPECL outputs. Therefore, proper biasing and termination are required to ensure correct operation of the device and to minimize signal integrity. The proper termination for LVPECL outputs is 50 Ω to (V_{CC} –2) V, but this DC voltage is not readily available on PCB. Therefore, a Thevenin equivalent circuit is worked out for the LVPECL termination in both direct-coupled (DC) and AC coupled configurations. These configurations are shown in Figure 12 for V_{CC} = 2.5 V and Figure 13 for V_{CC} = 3.3 V, respectively. It is recommended to place all resistive components close to either the driver end or the receiver end. If the supply voltage for the driver and receiver is different, AC coupling is required.

8.2 Functional Block Diagram



8.3 Feature Description

The CDCLVP1102 is a low additive jitter universal to LVPECL fan out buffer with 2 outputs and 1 input. The small package, low output skew, and low additive jitter make for a flexible device in demanding applications.

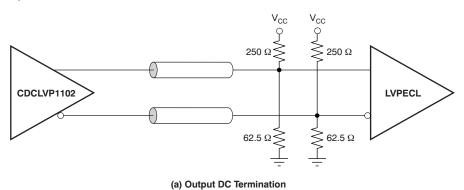
8.4 Device Functional Modes

The CDCLVP1102 is an open emitter for LVPECL outputs. Therefore, proper biasing and termination are required to ensure correct operation of the device and to minimize signal integrity. The proper termination for LVPECL outputs is 50 Ω to (V_{CC} –2) V, but this dc voltage is not readily available on PCB. Therefore, a Thevenin equivalent circuit is worked out for the LVPECL termination in both direct-coupled (DC) and AC-coupled configurations. These configurations are shown in Figure 12a and b for V_{CC} = 2.5 V and Figure 13a and b for V_{CC} = 3.3 V, respectively. It is recommended to place all resistive components close to either the driver end or the receiver end. If the supply voltage for the driver and receiver is different, ac coupling is required.

Product Folder Links: CDCLVP1102



8.4.1 LVPECL Output Termination



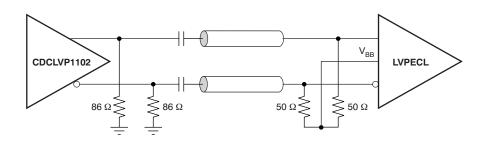
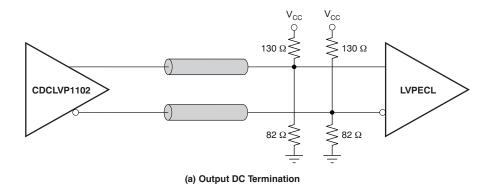


Figure 12. LVPECL Output DC and AC Termination for V_{CC} = 2.5 V

(b) Output AC Termination





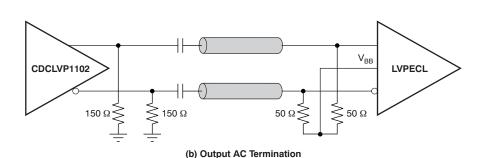


Figure 13. LVPECL Output DC and AC Termination for $V_{CC} = 3.3 \text{ V}$

8.4.2 Input Termination

The CDCLVP1102 inputs can be interfaced with LVPECL, LVDS, or LVCMOS drivers. Figure 14 illustrates how to dc couple an LVCMOS input to the CDCLVP1102. The series resistance (R_S) should be placed close to the LVCMOS driver; its value is calculated as the difference between the transmission line impedance and the driver output impedance.

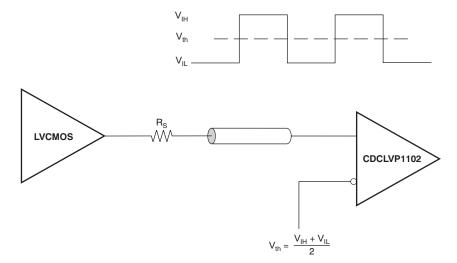


Figure 14. DC-Coupled LVCMOS Input to CDCLVP1102

Figure 15 shows how to dc couple LVDS inputs to the CDCLVP1102. Figure 16 and Figure 17 describe the method of dc coupling LVPECL inputs to the CDCLVP1102 for $V_{CC} = 2.5 \text{ V}$ and $V_{CC} = 3.3 \text{ V}$, respectively.



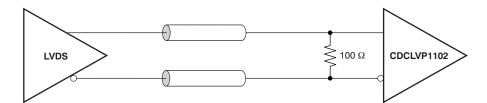


Figure 15. DC-Coupled LVDS Inputs to CDCLVP1102

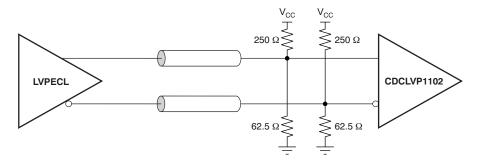


Figure 16. DC-Coupled LVPECL Inputs to CDCLVP1102 ($V_{CC} = 2.5 \text{ V}$)

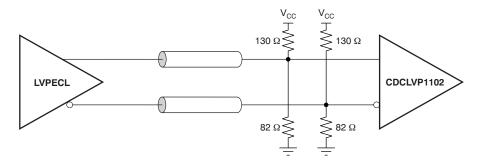


Figure 17. DC-Coupled LVPECL Inputs to CDCLVP1102 (V_{CC} = 3.3 V)

Figure 18 and Figure 19 show the technique of AC coupling differential inputs to the CDCLVP1102 for $V_{CC} = 2.5$ V and $V_{CC} = 3.3$ V, respectively. It is recommended to place all resistive components close to either the driver end or the receiver end. If the supply voltages of the driver and receiver are different, ac coupling is required.

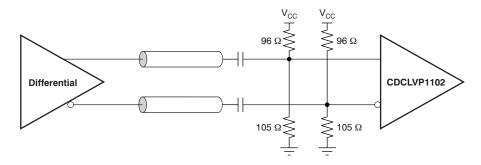


Figure 18. AC-Coupled Differential Inputs to CDCLVP1102 ($V_{CC} = 2.5 \text{ V}$)



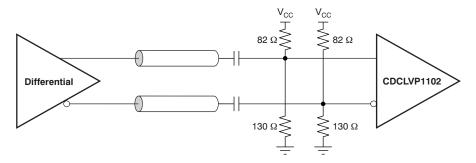


Figure 19. AC-Coupled Differential Inputs to CDCLVP1102 ($V_{CC} = 3.3 \text{ V}$)



9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The CDCLVP1102 is a low additive jitter LVPECL fanout buffer that can generate two copies of a LVPECL, LVDS, or LVCMOS input. The CDCLVP1102 can accept reference clock frequencies up to 2 GHz while providing low output skew.

9.2 Typical Application

9.2.1 Fanout Buffer for Line Card Application

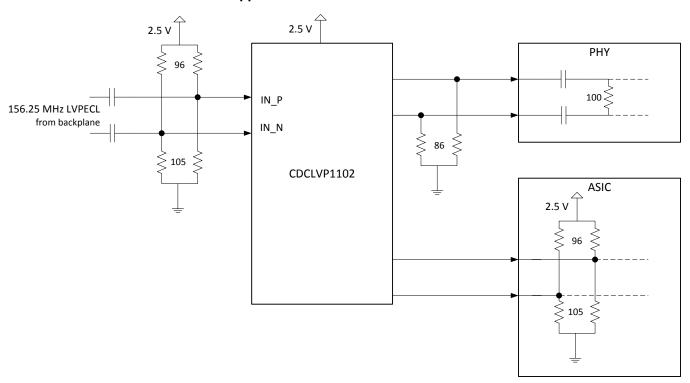


Figure 20. CDCLVP1102 Typical Application

9.2.1.1 Design Requirements

The CDCLVP1102 shown in Figure 20 is configured to receive a 156.25-MHz LVPECL clock from the backplane. Either signal can be then fanned out to desired devices, as shown. The configuration example is driving 2 LVPECL receivers in a line card application with the following properties:

- The PHY device has internal AC coupling and appropriate termination and biasing. The CDCLVP1102 will need to be provided with $86-\Omega$ emitter resistors near the driver for proper operation.
- The ASIC is capable of DC coupling with a 2.5-V LVPECL driver such as the CDCLVP1102. This ASIC features internal termination so no additional components are needed.



Typical Application (continued)

9.2.1.2 Detailed Design Procedure

Refer to Input Termination for proper input terminations, dependent on single ended or differential inputs.

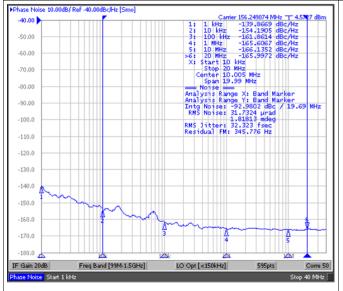
Refer to LVPECL Output Termination for output termination schemes depending on the receiver application. Unused outputs can be left floating.

In this example, the PHY, ASIC, and FPGA/CPU require different schemes. Power supply filtering and bypassing is critical for low noise applications.

See *Power Supply Recommendations* for recommended filtering techniques. A reference layout is provided on the CDCLVP1102 Evaluation Module at SCAU035.

9.2.1.3 Application Curves

The CDCLVP1102's low additive noise can be shown in this line card application. The low noise 156.25 MHz XO with 32-fs RMS jitter drives the CDCLVP1102, resulting in 57-fs RMS when integrated from 10 kHz to 20 MHz. The resultant additive jitter is a low 47-fs RMS for this configuration.



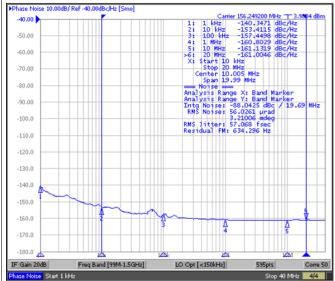


Figure 21. CDCLVP1102 Reference Phase Noise 32 fs rms (10 kHz to 20 MHz)

Figure 22. CDCLVP1102 Output Phase Noise 57 fs rms (10 kHz to 20 MHz)

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10 Power Supply Recommendations

High-performance clock buffers are sensitive to noise on the power supply, which can dramatically increase the additive jitter of the buffer. Thus, it is essential to reduce noise from the system power supply, especially when jitter/phase noise is very critical to applications.

Filter capacitors are used to eliminate the low-frequency noise from the power supply, where the bypass capacitors provide the very low impedance path for high-frequency noise and guard the power-supply system against the induced fluctuations. These bypass capacitors also provide instantaneous current surges as required by the device and should have low equivalent series resistance (ESR). To properly use the bypass capacitors, they must be placed very close to the power-supply pins and laid out with short loops to minimize inductance. It is recommended to add as many high-frequency (for example, 0.1-µF) bypass capacitors as there are supply pins in the package. It is recommended, but not required, to insert a ferrite bead between the board power supply and the chip power supply that isolates the high-frequency switching noises generated by the clock driver; these beads prevent the switching noise from leaking into the board supply. Choose an appropriate ferrite bead with very low dc resistance because it is imperative to provide adequate isolation between the board supply and the chip supply, as well as to maintain a voltage at the supply pins that is greater than the minimum voltage required for proper operation.

Figure 23 illustrates this recommended power-supply decoupling method.

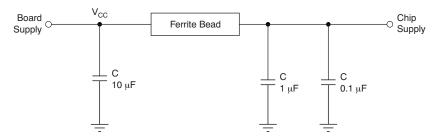


Figure 23. Power-Supply Decoupling

20



11 Layout

11.1 Layout Guidelines

Power consumption of the CDCLVP1102 can be high enough to require attention to thermal management. For reliability and performance reasons, the die temperature should be limited to a maximum of 125°C. That is, as an estimate, ambient temperature (T_A) plus device power consumption times $R_{A,A}$ should not exceed 125°C.

The device package has an exposed pad that provides the primary heat removal path to the printed circuit board (PCB). To maximize the heat dissipation from the package, a thermal landing pattern including multiple vias to a ground plane must be incorporated into the PCB within the footprint of the package. The exposed pad must be soldered down to ensure adequate heat conduction out of the package. Figure 24 shows a recommended land and via pattern.

11.2 Layout Example

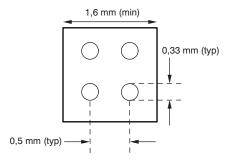


Figure 24. Recommended PCB Layout

11.3 Thermal Considerations

The CDCLVP1102 supports high temperatures on the printed circuit board (PCB) measured at the thermal pad. The system designer needs to ensure that the maximum junction temperature is not exceeded. Ψ_{jb} can allow the system designer to measure the board temperature with a fine gauge thermocouple and back calculate the junction temperature using Equation 1. Note that Ψ_{jb} is close to $R_{\theta JB}$ as 75 to 95% of a device's heat is dissipated by the PCB. Further information can be found at SPRA953 and SLUA566.

$$T_{\text{iunction}} = T_{\text{PCB}} + (\Psi_{\text{ib}} \times \text{Power}) \tag{1}$$

Example:

Calculation of the junction-lead temperature with a 4-layer JEDEC test board using four thermal vias:

$$T_{PCB} = 105 \, ^{\circ}C$$

$$\Psi_{ib} = 21.7 \, ^{\circ}\text{C/W}$$

Power_{inclTerm} = $I_{max} \times V_{max}$ = 105 mA × 3.6 V = 378 mW (max power consumption including termination resistors)

 $Power_{exclTerm} = 302.5 \text{ mW}$ (max power consumption excluding termination resistors, see SLYT127 for further details)

$$\Delta T_{Junction} = \Psi_{jb} \times Power_{exclTerm} = 21.7 \text{ °C/W} \times 302.5 \text{ mW} = 6.56 \text{ °C}$$

 $T_{Junction} = \Delta T_{Junction} + T_{Chassis} = 6.56 \, ^{\circ}\text{C} + 105 \, ^{\circ}\text{C} = 111.56 \, ^{\circ}\text{C}$ (the maximum junction temperature of 125 $\,^{\circ}\text{C}$ is not violated)

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12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

For additional information, see the following:

- Using Thermal Calculation Tools for Analog Components (SLUA566)
- CDCLVP1102 Evaluation Module (SCAU035)
- Power Consumption of LVPECL and LVDS (SLYT127)

12.2 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.3 Trademarks

E2E is a trademark of Texas Instruments.

All other trademarks are the property of their respective owners.

12.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

www.ti.com 10-Nov-2025

PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/	MSL rating/	Op temp (°C)	Part marking
	(1)	(2)			(3)	Ball material	Peak reflow		(6)
						(4)	(5)		
CDCLVP1102RGTR	Active	Production	VQFN (RGT) 16	3000 LARGE T&R	Yes	Call TI Nipdau	Level-2-260C-1 YEAR	-40 to 85	1102
CDCLVP1102RGTR.B	Active	Production	VQFN (RGT) 16	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1102
CDCLVP1102RGTT	Active	Production	VQFN (RGT) 16	250 SMALL T&R	Yes	Call TI Nipdau	Level-2-260C-1 YEAR	-40 to 85	1102
CDCLVP1102RGTT.B	Active	Production	VQFN (RGT) 16	250 SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1102
CDCLVP1102RGTTG4	Active	Production	VQFN (RGT) 16	250 SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1102
CDCLVP1102RGTTG4.B	Active	Production	VQFN (RGT) 16	250 SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1102

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.



PACKAGE OPTION ADDENDUM

www.ti.com 10-Nov-2025

PACKAGE MATERIALS INFORMATION

www.ti.com 12-Aug-2025

TAPE AND REEL INFORMATION



TAPE DIMENSIONS + K0 - P1 - B0 W Cavity - A0 -

	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	U	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CDCLVP1102RGTTG4	VQFN	RGT	16	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

PACKAGE MATERIALS INFORMATION

www.ti.com 12-Aug-2025



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CDCLVP1102RGTTG4	VQFN	RGT	16	250	210.0	185.0	35.0



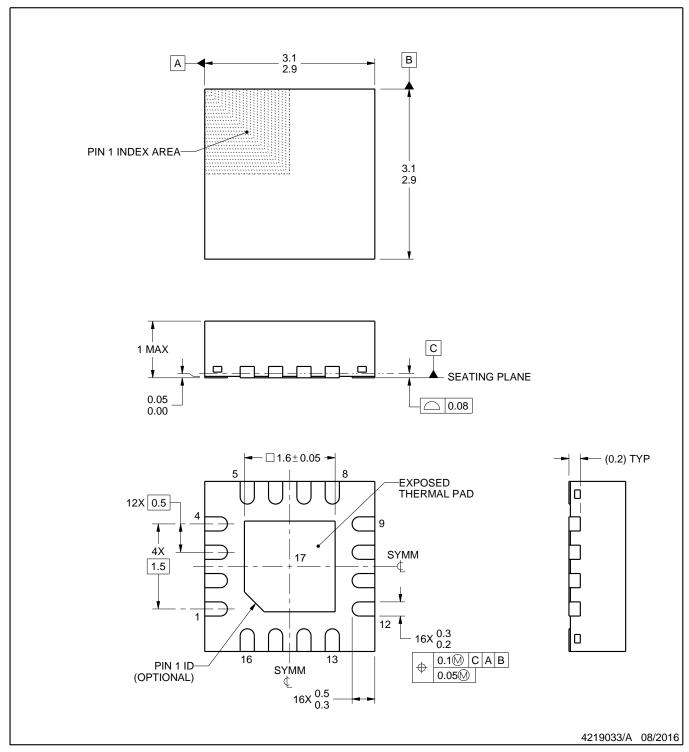
Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.







PLASTIC QUAD FLATPACK - NO LEAD

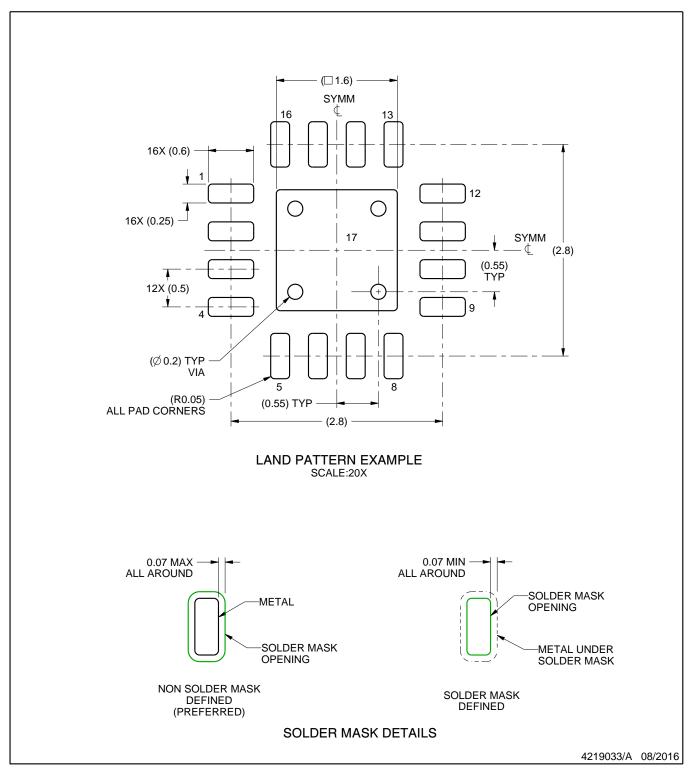


NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK - NO LEAD

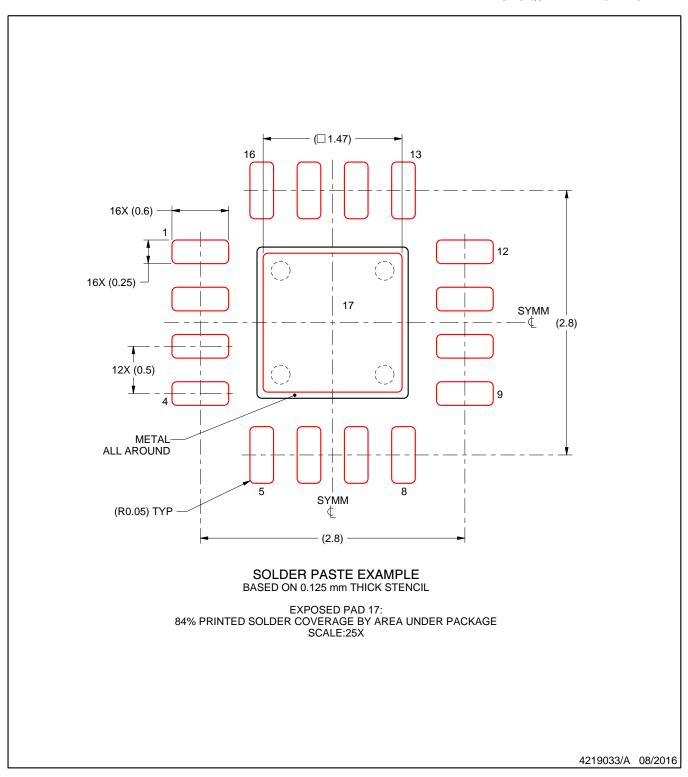


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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