



14-Bit, 125Msps ANALOG-TO-DIGITAL CONVERTER

#### **FEATURES**

- 14-Bit Resolution
- 125 Msps Sample Rate
- High SNR: 71.2 dBFS at 100-MHz f<sub>IN</sub>
- High SFDR: 82 dBc at 100-MHz f<sub>IN</sub>
- 2.3-V<sub>PP</sub> Differential Input Voltage
- Internal Voltage Reference
- 3.3-V Single-Supply Voltage
- Analog Power Dissipation: 578 mW
- Serial Programming Interface
- TQFP-64 PowerPAD™ Package
- Recommended Amplifiers: OPA695, OPA847, THS3201, THS3202, THS4503, THS4509, THS9001

#### **APPLICATIONS**

- Wireless Communication
  - Communication Receivers
  - Base Station Infrastructure
- Test and Measurement Instrumentation
- Single and Multichannel Digital Receivers
- Communication Instrumentation
  - Radar, Infrared
- Video and Imaging
- Medical Equipment

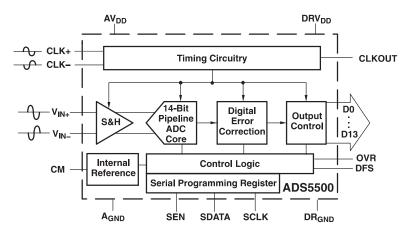
#### DESCRIPTION

The ADS5500 is a high-performance, 14-bit, 125 Msps analog-to-digital converter (ADC). To provide a converter solution, it includes high-bandwidth linear sample-and-hold stage (S&H) and internal reference. Designed for applications demanding the highest speed and highest dynamic performance in little space, the ADS5500 has excellent power consumption of 578 mW at 3.3-V single-supply voltage. This allows an even higher system integration density. The provided internal reference simplifies system design requirements. Parallel CMOS-compatible output ensures seamless interfacing with common logic.

The ADS5500 is available in a 64-pin TQFP PowerPAD™ package and in both a commercial and industrial temperature grade device.

#### **ADS5500 PRODUCT FAMILY**

	80 Msps	105 Msps	125 Msps
12 Bit	ADS5522	ADS5521	ADS5520
14 Bit	ADS5542	ADS5541	ADS5500



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## PACKAGE/ORDERING INFORMATION(1)

PRODUCT	PACKAGE LEAD	PACKAGE DESIGNATOR	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER	TRANSPORT MEDIA, QUANTITY
			–40°C to 85°C	ADS5500I	ADS5500IPAP	Tray, 160
ADS5500	HTQFP-64 <sup>(2)</sup>	PAP	-40 C to 65 C	AD355001	ADS5500IPAPR	Tape and Reel, 1000R
AD35500	PowerPAD	FAF	0°C to 70°C	ADS5500CPAP		Tray, 160
			0.0 10 10.0	ADS5500C	ADS5500CPAPR	Tape and Reel, 100

- (1) For the most current product and ordering information, see the Package Option Addendum located at the end of this data sheet.
- (2) Thermal pad size: 3,5 mm × 3,5 mm (min), 4 mm × 4 mm (max). θ<sub>JA</sub> = 21.47°C/W and θ<sub>JC</sub> = 2.99°C/W, when used with 2 oz. copper trace and pad soldered directly to a JEDEC standard four layer 3 in × 3 in PCB.

## **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range (unless otherwise noted) (1)

		ADS5500	UNIT
Supply voltage	AV <sub>DD</sub> to A <sub>GND</sub> , DRV <sub>DD</sub> to DR <sub>GND</sub>		V
Supply voltage	A <sub>GND</sub> to DR <sub>GND</sub>		V
Analog input to	A <sub>GND</sub> <sup>(2)(2)</sup>	-0.3 to Min (AV <sub>DD</sub> + 0.3 V, 3.6 V)	V
Logic input to D	R <sub>GND</sub>	– 0.3 to DRV <sub>DD</sub>	V
Digital data outp	out to DR <sub>GND</sub>	-0.3 to DRV <sub>DD</sub>	V
Operating temp	oratura ranga	0 to 70	°C
Operating temper	erature range	-40 to 85	C
Junction temper	ature	105	°C
Storage tempera	ature range	-65 to 150	°C

- (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.
- (2) If the input signal can exceed 3.6 V, then a resistor greater than or equal to 25 Ω should be added in series with each of the analog input pins to support input voltages up to 3.8 V. For input voltages above 3.8 V, the device can only handle transients and the duty cycle of the overshoot should be limited to less than 5% for inputs up to 3.9 V.

#### RECOMMENDED OPERATING CONDITIONS

			MIN	NOM	MAX	UNIT
SUPPLI	ES					
$AV_{DD}$	Analog supply voltage		3	3.3	3.6	V
$DRV_DD$	Output driver supply voltage		3	3.3	3.6	V
ANALO	G INPUT					
	Differential input range			2.3		$V_{PP}$
$V_{CM}$	Input common-mode voltage <sup>(1)</sup>		1.45	1.55	1.65	V
DIGITAL	OUTPUT					
	Maximum output load			10		pF
CLOCK	INPUT					
	ADCLK input sample rate (sine	DLL ON	60		125	
	wave) 1/t <sub>C</sub>	DLL OFF	2		80	Msps
	Clock amplitude, sine wave, diffe	erential (see Figure 50 for more information)	1	3		$V_{PP}$

(1) Input common-mode should be connected to CM.



## RECOMMENDED OPERATING CONDITIONS (continued)

	MIN	NOM	MAX	UNIT
Clock duty cycle (see Figure 49 for more information)		50%		
Open free air temperature range	0		70	9
Open free-air temperature range	-40		85	

## **ELECTRICAL CHARACTERISTICS**

Typ values given at  $T_A$  = 25°C, min and max specified over the full recommended operating temperature range, AVDD = DRV<sub>DD</sub> = 3.3 V, sampling rate = 125 Msps, 50% clock duty cycle, DLL On, 3-V<sub>PP</sub> differential clock, and -1-dBFS differential input, unless otherwise noted

	PARAMETER	TEST C	CONDITIONS	MIN	TYP	MAX	UNIT
Resolut	ion				14		Bits
ANALO	G INPUTS						
	Differential input range				2.3		$V_{PP}$
	Differential input impedance	See Figure 41			6.6		Ω
	Differential input capacitance	See Figure 41			4		pF
	Analog input common-mode current (per input)				300		μΑ
	Analog input bandwidth	Source impeda	nce = $50\Omega$		750		MHz
	Voltage overload recovery time				4		Clock Cycles
INTERN	AL REFERENCE VOLTAGES						
$V_{REFM}$	Reference bottom voltage				0.97		V
$V_{REFP}$	Reference top voltage				2.11		V
	Reference error			- 4%	±0.9%	4%	
$V_{CM}$	Common-mode voltage output			1	1.55 ±0.05		V
DYNAM	IC DC CHARACTERISTICS AND ACCURACY						
	No missing codes				Tested		
DNL	Differential linearity error	f <sub>IN</sub> = 10 MHz		-0.9	±0.75	1.1	LSB
INL	Integral linearity error	f <sub>IN</sub> = 10 MHz		<b>-</b> 5	±2.5	5	LSB
	Offset error			-11	±1.5	11	mV
	Offset temperature coefficient 0.02						mV/°C
	DC power supply rejection ratio, DC PSRR	$\Delta$ offset error/ $\Delta$ AV to AV <sub>DD</sub> = 3.6	$AV_{DD}$ from $AV_{DD} = 3$		0.25		mV/V
	Gain error <sup>(1)</sup>			-2	±0.45	2	%FS
	Gain temperature coefficient				0.01		∆%/°C
DYNAM	IC AC CHARACTERISTICS						
		f 10 MHz	25°C to T <sub>MAX</sub>	71.5	73.2		
		$f_{IN} = 10 \text{ MHz}$	Full temp range	70.5	72.8		
		f <sub>IN</sub> = 30 MHz			72.7		
		f <sub>IN</sub> = 55 MHz			71.9		
SNR	Signal-to-noise ratio	( 70 MH	25°C to T <sub>MAX</sub>	70.8	72.3		dBFS
		$f_{IN} = 70 \text{ MHz}$	Full temp range	69.8	72		
		f <sub>IN</sub> = 100 MHz			71.2		
		f <sub>IN</sub> = 150 MHz			70.1		
		f <sub>IN</sub> = 225 MHz			69.1		
	RMS output noise	Input tied to cor	mmon-mode		1.1		LSB

<sup>(1)</sup> Gain error is specified by design and characterization; it is not tested in production.



Typ values given at  $T_A$  = 25°C, min and max specified over the full recommended operating temperature range, AVDD = DRV<sub>DD</sub> = 3.3 V, sampling rate = 125 Msps, 50% clock duty cycle, DLL On, 3-V<sub>PP</sub> differential clock, and -1-dBFS differential input, unless otherwise noted

	PARAMETER	TEST C	ONDITIONS	MIN	TYP I	XAN	UNIT	
		f 40 MH-	25°C	82	84			
		$f_{IN} = 10 \text{ MHz}$	Full temp range	78	84			
		$f_{IN} = 30 \text{ MHz}$			84			
		$f_{IN} = 55 \text{ MHz}$			79			
SFDR	Spurious-free dynamic range	( 70 MIL	25°C	80	83		dBc	
		$f_{IN} = 70 \text{ MHz}$	Full temp range	77	82			
		f <sub>IN</sub> = 100 MHz			82			
		f <sub>IN</sub> = 150 MHz			78			
		f <sub>IN</sub> = 225 MHz			74			
		( 40 MH-	25°C	82	91			
		$f_{IN} = 10 \text{ MHz}$	Full temp range	78	86			
		f <sub>IN</sub> = 30 MHz			86			
		f <sub>IN</sub> = 55 MHz			84			
HD2	Second-harmonic		25°C	80	87		dBc	
		$f_{IN} = 70 \text{ MHz}$	Full temp range	77	83			
		f <sub>IN</sub> = 100 MHz			84			
		f <sub>IN</sub> = 150 MHz	f <sub>IN</sub> = 150 MHz		78			
		f <sub>IN</sub> = 225 MHz			74			
		( 40 MH-	25°C	82	89			
			$f_{IN} = 10 \text{ MHz}$	Full temp range	78	88		
			$f_{IN} = 30 \text{ MHz}$			90		
		$f_{IN} = 55 \text{ MHz}$			79			
HD3	Third-harmonic	4 70 MH-	25°C	80	85		dBc	
		$f_{IN} = 70 \text{ MHz}$	Full temp range	77	82			
		$f_{IN} = 100 \text{ MHz}$			82			
		f <sub>IN</sub> = 150 MHz			80			
		f <sub>IN</sub> = 225 MHz			76			
	Worst-harmonic/spur	f <sub>IN</sub> = 10 MHz	25°C		88		40.	
	(other than HD2 and HD3)	f <sub>IN</sub> = 70 MHz	25°C		86		dBc	
		4 40 MIL	25°C to T <sub>MAX</sub>	71	72.8			
		$f_{IN} = 10 \text{ MHz}$	Full temp range	69.5	72.2			
		f <sub>IN</sub> = 30 MHz			72.3			
SINAD		$f_{IN} = 55 \text{ MHz}$			70.7			
	Signal-to-noise + distortion		25°C to T <sub>MAX</sub>	70.3	71.6		dBFS	
		$f_{IN} = 70 \text{ MHz}$	Full temp range	69	71.3			
		f <sub>IN</sub> = 100 MHz			70.5			
		f <sub>IN</sub> = 150 MHz			69.1			
		f <sub>IN</sub> = 225 MHz			67.4			



Typ values given at  $T_A$  = 25°C, min and max specified over the full recommended operating temperature range, AVDD = DRV<sub>DD</sub> = 3.3 V, sampling rate = 125 Msps, 50% clock duty cycle, DLL On, 3-V<sub>PP</sub> differential clock, and -1-dBFS differential input, unless otherwise noted

	PARAMETER	TEST C	ONDITIONS	MIN	TYP	MAX	UNIT
		f _ 10 MHz	25°C	80	85		
		f <sub>IN</sub> = 10 MHz	Full temp range	78	83		
		f <sub>IN</sub> = 30 MHz			82		
		$f_{IN} = 55 \text{ MHz}$			77		
THD	Total harmonic distortion	f _ 70 MHz	25°C	77.5	81		dBc
		$f_{IN} = 70 \text{ MHz}$	Full temp range	76	79.5		
		f <sub>IN</sub> = 100 MHz			79		
	$f_{IN} = 150 \text{ MHz}$		75				
		f <sub>IN</sub> = 225 MHz			71.8		
ENOB	Effective number of bits	f <sub>IN</sub> = 70 MHz			11.3		Bits
		f = 10.1 MHz, 19 (-7 dBFS each			95		
IMD	Two-tone intermodulation distortion	f = 30.1 MHz, 39 (-7 dBFS each			94		dBFS
		f = 50.1 MHz, 55 (-7 dBFS each			94		
ACPSRR	AC power supply rejection ratio	Supply noise fre	equency ≤ 100 MHz		35		dB
POWER S	SUPPLY		·				
I <sub>CC</sub>	Total supply current	f <sub>IN</sub> = 70 MHz			236	265	mA
I <sub>AVDD</sub>	Analog supply current	f <sub>IN</sub> = 70 MHz			175	190	mA
I <sub>DRVDD</sub>	Output buffer supply current	f <sub>IN</sub> = 70 MHz, 10 digital outputs to			61	75	mA
		Analog only			578	627	mW
	Power dissipation	Output buffer policy load on digital o			202	248	mW
	Standby power	With clocks runn	ning		181	250	mW

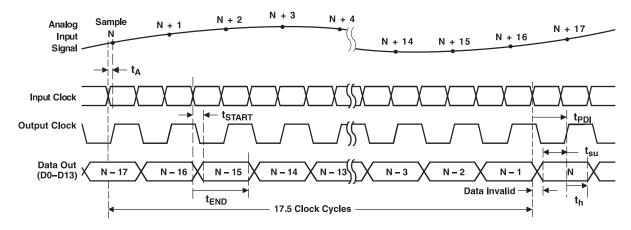
## **DIGITAL CHARACTERISTICS**

Valid over the full recommended operating temperature range,  $AV_{DD} = DRV_{DD} = 3.3 \text{ V}$ , unless otherwise noted

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DIGITAL INPUTS					
High-level input voltage		2.4			V
Low-level input voltage				8.0	V
High-level input current				10	μΑ
Low-level input current				-10	μΑ
Input current for RESET			-20		μΑ
Input capacitance			4		pF
DIGITAL OUTPUTS	·				
Low-level output voltage	C <sub>LOAD</sub> = 10 pF		0.3		V
High-level output voltage	C <sub>LOAD</sub> = 10 pF	2.8	3		V
Output capacitance			3		pF



#### TIMING CHARACTERISTICS



NOTE: It is recommended that the loading at CLKOUT and all data lines are accurately matched to ensure that the above timing matches closely with the specified values.

NOTE: It is recommended that the loading at CLKOUT and all data lines are accurately matched to ensure that the above timing matches closely with the specified values.

Figure 1. Timing Diagram

## TIMING CHARACTERISTICS(1)(2)

Typ values given at  $T_A$  = 25°C, min and max specified over the full recommended operating temperature range, sampling rate = 125 Msps, 50% clock duty cycle,  $AV_{DD}$  = DRV<sub>DD</sub> = 3.3 V, 3-V<sub>PP</sub> differential clock, and  $C_{LOAD}$  = 10 pF, (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SWITCH	HING SPECIFICATION		Į.		'!	
t <sub>A</sub>	Aperture delay	Input CLK falling edge to data sampling point		1		ns
	Aperture jitter (uncertainty)	Uncertainty in sampling instant		300		fs
t <sub>su</sub>	Data setup time	Data valid (3) to 50% of CLKOUT rising edge	2.1	2.5		ns
t <sub>h</sub>	Data hold time	50% of CLKOUT rising edge to data becoming invalid <sup>(3)</sup>	1.7	2.1		ns
t <sub>START</sub>	Input clock to output data valid start (4)(5)	Input clock rising edge to Data valid start delay		2.2	2.9	ns
t <sub>END</sub>	Input clock to output data valid end <sup>(4)(5)</sup>	Input clock rising edge to Data valid end delay	5.8	6.9		ns
t <sub>JIT</sub>	Output clock jitter	Uncertainty in CLKOUT rising edge, peak-to-peak		150	210	ps
t <sub>r</sub>	Output clock rise time	Rise time of CLKOUT measured from 20% to 80% of DRVDD		1.7	1.9	ns
t <sub>f</sub>	Output clock fall time	Fall time of CLKOUT measured from 80% to 20% of DRVDD		1.5	1.7	ns
t <sub>PDI</sub>	Input clock to output clock delay	Input clock rising edge, zero crossing, to output clock rising edge 50%	4.2	4.8	5.5	ns
t <sub>r</sub>	Data rise time	Data rise time measured from 20% to 80% of DRVDD		3.6	4.6	ns
t <sub>f</sub>	Data fall time	Data fall time measured from 80% to 20% of DRVDD		2.8	3.7	ns
	Output enable (OE) to data output delay	Time required for outputs to have stable timings w.r.t input clock( after OE is activated			1000	Clock Cycles

- (1) Timing parameters are ensured by design and characterization and not tested in production.
- (2) See Table 5 through Table 8 in the Application Information section for timing information at additional sampling frequencies.
- (3) Data valid refers to 2 V for LOGIC high and 0.8 V for LOGIC low.
- (4) See the Output Information section for details on using the input clock for data capture.
- These specifications apply when the CLKOUT polarity is set to rising edge (according to Table 3). Add ½ clock period for the valid number for a falling edge CLKOUT polarity.



## TIMING CHARACTERISTICS (continued)

Typ values given at  $T_A$  = 25°C, min and max specified over the full recommended operating temperature range, sampling rate = 125 Msps, 50% clock duty cycle,  $AV_{DD}$  =  $DRV_{DD}$  = 3.3 V, 3- $V_{PP}$  differential clock, and  $C_{LOAD}$  = 10 pF, (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Wake-up time	Time to valid data after coming out of software power down			1000	Clock
·	Time to valid data after stopping and restarting the clock			1000	Cycles
Latency Time for a sample to propagate to the ADC outputs 17.5 Clock Cycles			17.5		Clock Cycles

## **RESET TIMING CHARACTERISTICS**

Typ values given at  $T_A = 25$ °C, min and max specified over the full recommended operating temperature range,  $AV_{DD} = DRV_{DD} = 3.3 \text{ V}$ ,  $3-V_{PP}$  differential clock(unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
SWIT	CHING SPECIFICATION				
t <sub>1</sub>	Power-on delay	Delay from power on of AV <sub>DD</sub> and DRV <sub>DD</sub> to RESET pulse	10		ms
t <sub>2</sub>	Reset pulse width	Pulse width of active RESET signal	2		μs
t <sub>3</sub>	Register write delay	Delay from RESET disable to SEN active	2		μs
	Power-up time	Delay from power-up of AV <sub>DD</sub> and DRV <sub>DD</sub> to output stable		40	ms

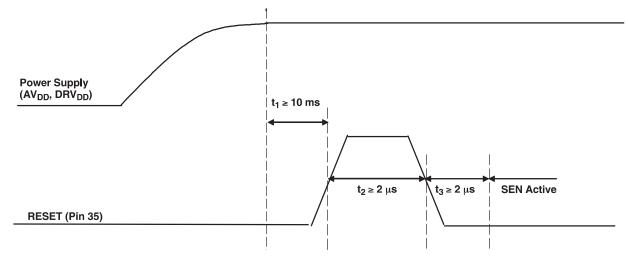


Figure 2. Reset Timing Diagram

## SERIAL PROGRAMMING INTERFACE CHARACTERISTICS

The device has a three-wire serial interface. The device latches the serial data SDATA on the falling edge of serial clock SCLK when SEN is active.

- Serial shift of bits is enabled when SEN is low. SCLK shifts serial data at falling edge.
- Minimum width of data stream for a valid loading is 16 clocks.
- Data is loaded at every 16th SCLK falling edge while SEN is low.
- In case the word length exceeds a multiple of 16 bits, the excess bits are ignored.
- Data can be loaded in multiple of 16-bit words within a single active SEN pulse.
- The first 4-bit nibble is the address of the register while the last 12 bits are the register contents.



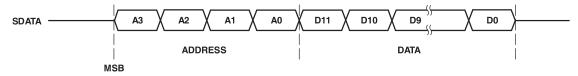


Figure 3. DATA Communication is 2-Byte, MSB First

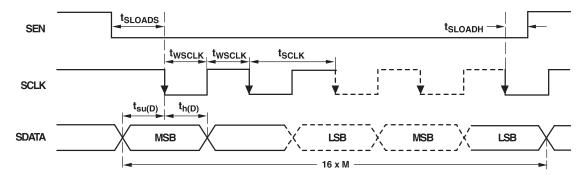


Figure 4. Serial Programming Interface Timing Diagram

**Table 1. Serial Programming Interface Timing Characteristics** 

	PARAMETER	MIN <sup>(1)</sup>	TYP <sup>(1)</sup>	MAX <sup>(1)</sup>	UNIT
t <sub>SCLK</sub>	SCLK period	50			ns
t <sub>WSCLK</sub>	SCLK duty cycle	25%	50%	75%	
t <sub>SLOADS</sub>	SEN to SCLK setup time	8			ns
t <sub>SLOADH</sub>	SCLK to SEN hold time	6			ns
t <sub>DS</sub>	Data setup time	8			ns
t <sub>DH</sub>	Data hold time	6			ns

(1) Min, typ, and max values are characterized, but not production tested.

Table 2. Serial Register Table<sup>(1)</sup>

<b>G</b>																
А3	A2	A1	A0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	DESCRIPTION
														DLL CTR L		Clock DLL
1	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0	Internal DLL is on, recommended for 60–125 Msps clock speed
1	1	0	1	0	0	0	0	0	0	0	0	0	0	1	0	Internal DLL is off, recommended for 2-80 Msps clock speed
					TP<1>	TP<0>										Test Mode
1	1	1	0	0	0	0	0	0	0	0	0	0	0	Х	0	Normal mode of operation
1	1	1	0	0	0	1	0	0	0	0	0	0	0	Х	0	All outputs forced to 0.(2)
1	1	1	0	0	1	0	0	0	0	0	0	0	0	Х	0	All outputs forced to 1. (2)
1	1	1	0	0	1	1	0	0	0	0	0	0	0	Х	0	Each output bit toggles between 0 and 1. (2)(3)
				PDN												Power Down
1	1	1	1	0	0	0	0	0	0	0	0	0	0	Х	0	Normal mode of operation
1	1	1	1	1	0	0	0	0	0	0	0	0	0	Х	0	Device is put in power down (low current) mode

<sup>(1)</sup> The register contents default to the appropriate setting for normal operation upon RESET.

<sup>(2)</sup> The patterns given are applicable to the straight offset binary output format. If two's complement output format is selected, the test mode outputs will be the two's complement equivalent of these patterns as described in the Output Information section.

<sup>(3)</sup> While each bit toggles between 1 and 0 in this mode, there is no assured phase relationship between the data bits D0 through D13. For example, when D0 is a 1, D1 in not assured to be a 0, and vice versa.

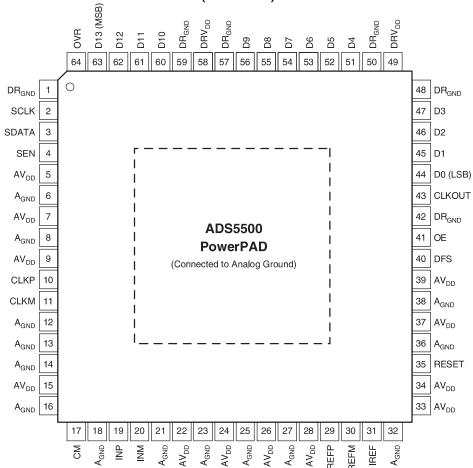


## Table 3. Data Format Select (DFS Table)

DFS-PIN VOLTAGE (V <sub>DFS</sub> )	DATA FORMAT	CLOCK OUTPUT POLARITY
$V_{DFS} < \frac{2}{12} \times AV_{DD}$	Straight Binary	Data valid on rising edge
$\frac{4}{12} \times AV_{DD} < V_{DFS} < \frac{5}{12} \times AV_{DD}$	Two's Complement	Data valid on rising edge
$\frac{7}{12} \times \text{AV}_{\text{DD}} < \text{V}_{\text{DFS}} < \frac{8}{12} \times \text{AV}_{\text{DD}}$	Straight Binary	Data valid on falling edge
$V_{DFS} > \frac{10}{12} \times AV_{DD}$	Two's Complement	Data valid on falling edge

## **PIN CONFIGURATION**

## PAP PACKAGE (TOP VIEW)



## **PIN ASSIGNMENTS**

	TERMINAL	NO. OF	1/0	DESCRIPTION				
NAME	NO.	PINS	1/0	DESCRIPTION				
AV <sub>DD</sub>	5, 7, 9, 15, 22, 24, 26, 28, 33, 34, 37, 39	12	I	Analog power supply				
A <sub>GND</sub>	6, 8, 12-14, 16, 18, 21, 23, 25, 27, 32, 36, 38	14	I	Analog ground				
DRV <sub>DD</sub>	49, 58	2	I	Output driver power supply				



## PIN CONFIGURATION (continued)

#### **PIN ASSIGNMENTS (continued)**

TERM	IINAL	NO. OF	I/O	DESCRIPTION
NAME	NO.	PINS	1/0	DESCRIPTION
DR <sub>GND</sub>	1, 42, 48, 50, 57, 59	6	I	Output driver ground
INP	19	1	I	Differential analog input (positive)
INM	20	1	I	Differential analog input (negative)
REFP	29	1	0	Reference voltage (positive); 1-μF capacitor to GND
REFM	30	1	0	Reference voltage (negative); 1-μF capacitor to GND
IREF	31	1	I	Current set; $56.2$ -k $\Omega$ resistor to GND; do not connect capacitors
CM	17	1	0	Common-mode output voltage
RESET	35	1	I	Reset (active high), internal 200-kΩ resistor to AVDD <sup>(1)</sup>
OE	41	1	I	Output enable (active high)
DFS	40	1	I	Data format and clock out polarity select (2)(3)
CLKP	10	1	I	Data converter differential input clock (positive)
CLKM	11	1	I	Data converter differential input clock (negative)
SEN	4	1	I	Serial interface chip select (3)
SDATA	3	1	I	Serial interface data (3)
SCLK	2	1	I	Serial interface clock (3)
D0 (LSB)-D13(MSB)	44-47, 51-56, 60-63	14	0	Parallel data output
OVR	64	1	0	Over-range indicator bit
CLKOUT	43	1	0	CMOS clock out in sync with data
NOTE: PowerPAD mus	st be connected to analog	g ground.		

- (1) If unused, the RESET pin should be tied to AGND. See the serial programming interface section for details.
- (2) Table 3 defines the voltage levels for each mode selectable via the DFS pin.
- (3) Pins OE, DFS, SEN, SDĂTA, and SCLK have internal clamping diodes to the DRVDD supply. Any external circuit driving these pins must also run off the same supply voltage as DRVDD.

#### **DEFINITION OF SPECIFICATIONS**

#### **Analog Bandwidth**

The analog input frequency at which the power of the fundamental is reduced by 3 dB with respect to the low frequency value.

#### **Aperture Delay**

The delay in time between the falling edge of the input sampling clock and the actual time at which the sampling occurs.

#### **Aperture Uncertainty (Jitter)**

The sample-to-sample variation in aperture delay.

#### **Clock Pulse Width/Duty Cycle**

The duty cycle of a clock signal is the ratio of the time the clock signal remains at a logic high (clock pulse width) to the period of the clock signal. Duty cycle is typically expressed as a percentage. A perfect differential sine-wave clock results in a 50% duty cycle.

#### **Maximum Conversion Rate**

The maximum sampling rate at which certified operation is given. All parametric testing is performed at this sampling rate unless otherwise noted.

#### **Minimum Conversion Rate**

The minimum sampling rate at which the ADC functions.

## **Differential Nonlinearity (DNL)**

An ideal ADC exhibits code transitions at analog input values spaced exactly 1 LSB apart. The DNL is the deviation of any single step from this ideal value, measured in units of LSBs.

## Integral Nonlinearity (INL)

The INL is the deviation of the ADC's transfer function from a best fit line determined by a least squares curve fit of that transfer function, measured in units of LSBs.

#### **Gain Error**

The gain error is the deviation of the ADC's actual input full-scale range from its ideal value. The gain error is given as a percentage of the ideal input full-scale range. Gain error does not account for variations in the internal reference voltages (see the *Electrical Specifications* section for limits on the variation of  $V_{REFP}$  and  $V_{REFM}$ ).



#### **Offset Error**

The offset error is the difference, given in number of LSBs, between the ADC's actual average idle channel output code and the ideal average idle channel output code. This quantity is often mapped into mV.

#### **Temperature Drift**

The temperature drift coefficient (with respect to gain error and offset error) specifies the change per degree Celsius of the parameter from  $T_{MIN}$  to  $T_{MAX}$ . It is calculated by dividing the maximum deviation of the parameter across the  $T_{MIN}$  to  $T_{MAX}$  range by the difference  $T_{MAX}$ - $T_{MIN}$ .

## Signal-to-Noise Ratio

SNR is the ratio of the power of the fundamental ( $P_S$ ) to the noise floor power ( $P_N$ ), excluding the power at DC and the first eight harmonics.

$$SNR = 10Log_{10} \frac{P_S}{P_N}$$
 (1)

SNR is either given in units of dBc (dB to carrier) when the absolute power of the fundamental is used as the reference, or dBFS (dB to full scale) when the power of the fundamental is extrapolated to the converter's full-scale range.

#### Signal-to-Noise and Distortion (SINAD)

SINAD is the ratio of the power of the fundamental  $(P_S)$  to the power of all the other spectral components including noise  $(P_N)$  and distortion  $(P_D)$ , but excluding dc.

$$SINAD = 10Log_{10} \frac{P_S}{P_N + P_D}$$
 (2)

SINAD is either given in units of dBc (dB to carrier) when the absolute power of the fundamental is used as the reference, or dBFS (dB to full scale) when the power of the fundamental is extrapolated to the converter's full-scale range.

#### **Effective Number of Bits (ENOB)**

The ENOB is a measure of a converter's performance as compared to the theoretical limit based on quantization noise.

ENOB = 
$$\frac{\text{SINAD} - 1.76}{6.02}$$
 (3)

#### **Total Harmonic Distortion (THD)**

THD is the ratio of the power of the fundamental  $(P_S)$  to the power of the first eight harmonics  $(P_D)$ .

$$THD = 10Log_{10} \frac{P_S}{P_D}$$
 (4)

THD is typically given in units of dBc (dB to carrier).

## Spurious-Free Dynamic Range (SFDR)

The ratio of the power of the fundamental to the highest other spectral component (either spur or harmonic). SFDR is typically given in units of dBc (dB to carrier).

#### **Two-Tone Intermodulation Distortion**

IMD3 is the ratio of the power of the fundamental (at frequencies f1 and f2) to the power of the worst spectral component at either frequency  $2f_1-f_2$  or  $2f_2-f_1$ . IMD3 is either given in units of dBc (dB to carrier) when the absolute power of the fundamental is used as the reference, or dBFS (dB to full scale) when the power of the fundamental is extrapolated to the converter's full-scale range.

## DC Power Supply Rejection Ratio (DC PSRR)

The DC PSSR is the ratio of the change in offset error to a change in analog supply voltage. The DC PSRR is typically given in units of mV/V.

#### Reference Error

The reference error is the variation of the actual reference voltage ( $V_{REFP}-V_{REFM}$ ) from its ideal value. The reference error is typically given as a percentage.

### **Voltage Overload Recovery Time**

The voltage overload recovery time is defined as the time required for the ADC to recover to within 1% of the full-scale range in response to an input voltage overload of 10% beyond the full-scale range.



## TYPICAL CHARACTERISTICS

Typical values are at  $T_A = 25$ °C,  $AV_{DD} = DRV_{DD} = 3.3$  V, differential input amplitude = -1 dBFS, sampling rate = 125 Msps, DLL On, and 3-V differential clock unless otherwise noted

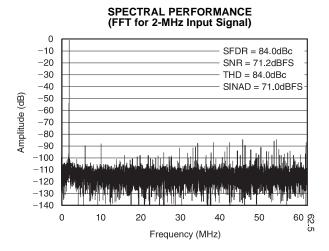


Figure 5.

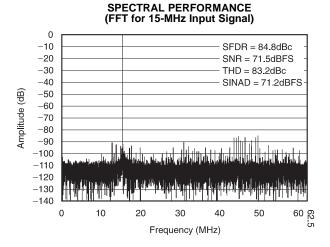


Figure 6.



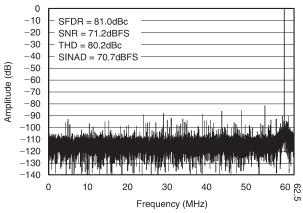


Figure 7.

# SPECTRAL PERFORMANCE (FFT for 70-MHz Input Signal)

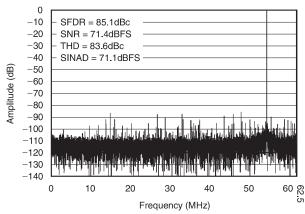


Figure 8.

# SPECTRAL PERFORMANCE (FFT for 80-MHz Input Signal)

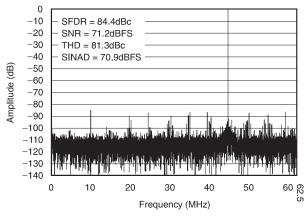


Figure 9.

# SPECTRAL PERFORMANCE (FFT for 100-MHz Input Signal)

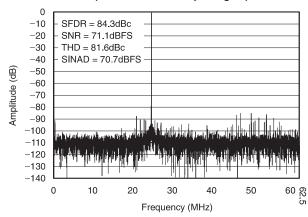


Figure 10.



Typical values are at  $T_A = 25$ °C,  $AV_{DD} = DRV_{DD} = 3.3$  V, differential input amplitude = -1 dBFS, sampling rate = 125 Msps, DLL On, and 3-V differential clock unless otherwise noted

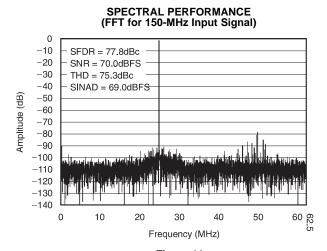


Figure 11.

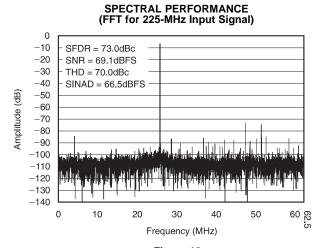


Figure 12.



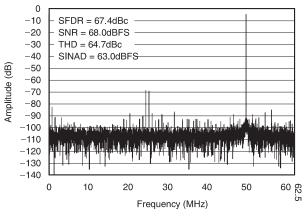


Figure 13.

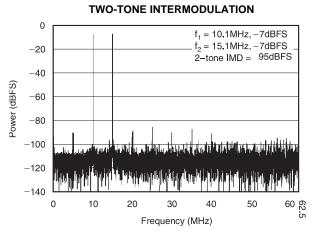


Figure 14.

## TWO-TONE INTERMODULATION

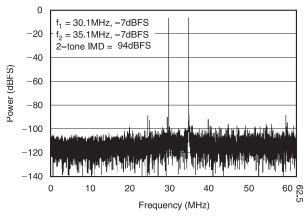


Figure 15.

## TWO-TONE INTERMODULATION

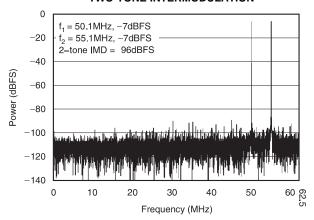


Figure 16.



Typical values are at  $T_A = 25$ °C,  $AV_{DD} = DRV_{DD} = 3.3$  V, differential input amplitude = -1 dBFS, sampling rate = 125 Msps, DLL On, and 3-V differential clock unless otherwise noted

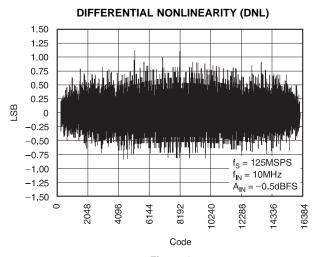


Figure 17.

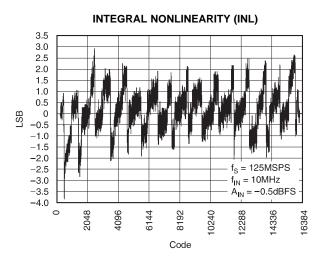


Figure 18.



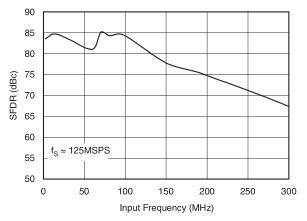


Figure 19.

#### SIGNAL-TO-NOISE RATIO vs INPUT FREQUENCY

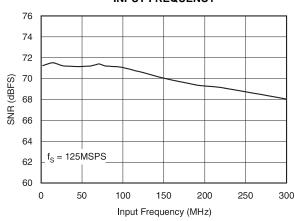
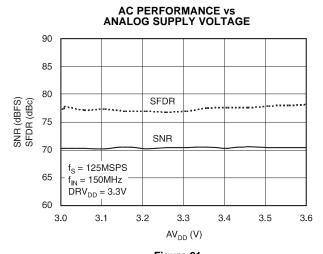
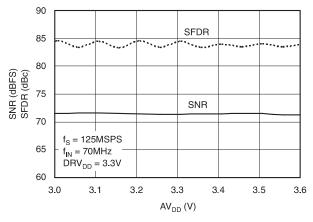


Figure 20.



Typical values are at  $T_A = 25$ °C,  $AV_{DD} = DRV_{DD} = 3.3$  V, differential input amplitude = -1 dBFS, sampling rate = 125 Msps, DLL On, and 3-V differential clock unless otherwise noted





AC PERFORMANCE vs ANALOG SUPPLY VOLTAGE

Figure 21.

Figure 22.



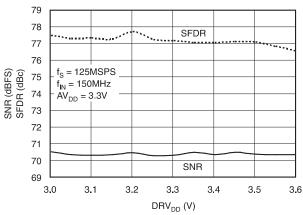


Figure 23.

#### AC PERFORMANCE vs DIGITAL SUPPLY VOLTAGE

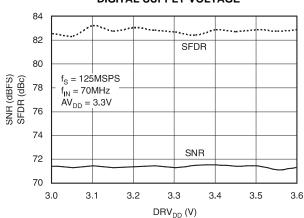


Figure 24.



Typical values are at  $T_A = 25$ °C,  $AV_{DD} = DRV_{DD} = 3.3$  V, differential input amplitude = -1 dBFS, sampling rate = 125 Msps, DLL On, and 3-V differential clock unless otherwise noted

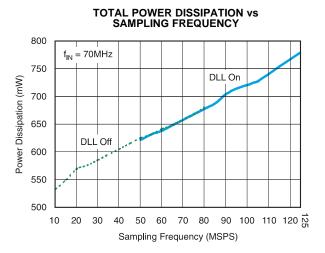


Figure 25.

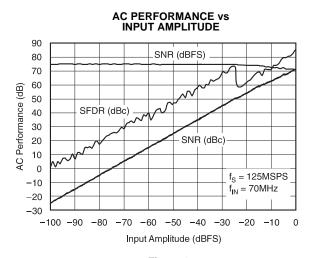


Figure 27.

# SIGNAL-TO-NOISE RATIO AND SPURIOUS-FREE DYNAMIC RANGE vs TEMPERATURE

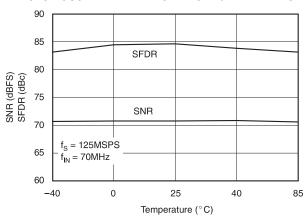


Figure 26.

#### AC PERFORMANCE vs INPUT AMPLITUDE

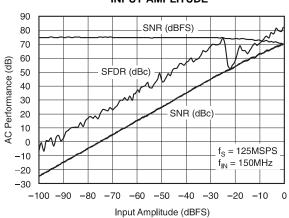


Figure 28.



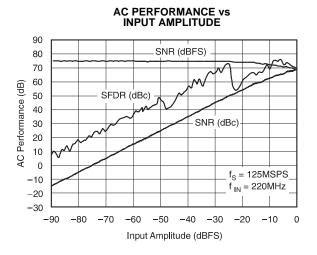


Figure 29.

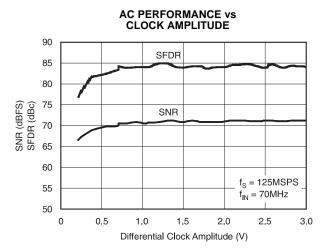


Figure 31.

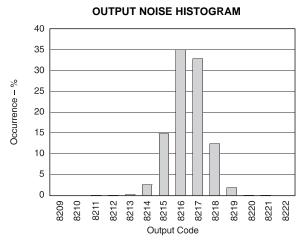


Figure 30.

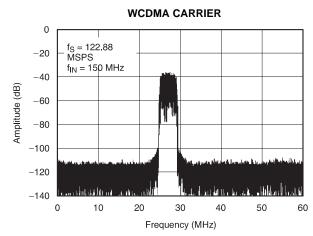


Figure 32.



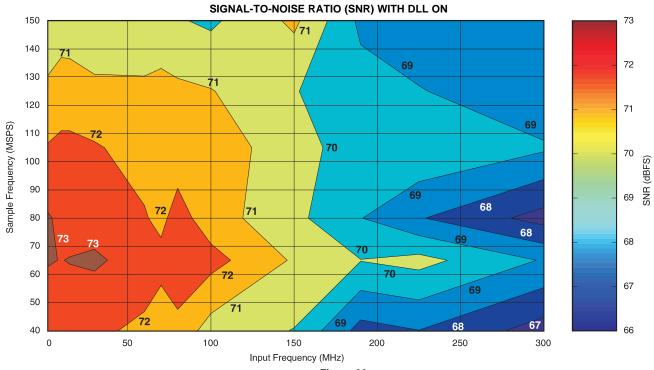


Figure 33.

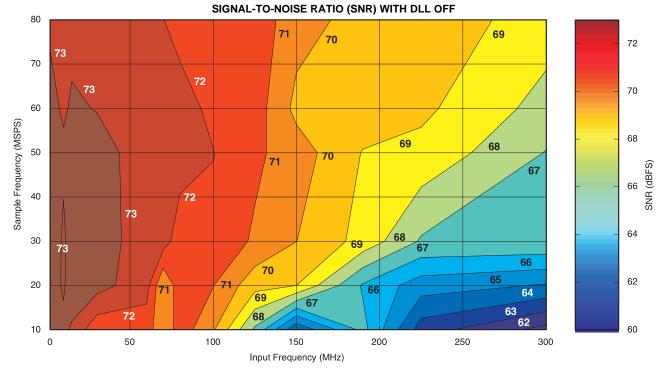
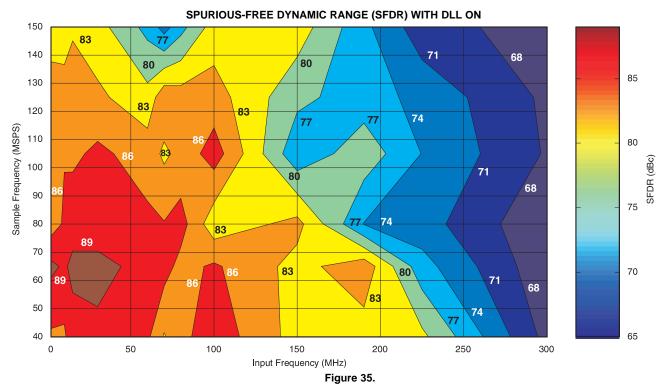


Figure 34.





SPURIOUS-FREE DYNAMIC RANGE (SFDR) WITH DLL OFF

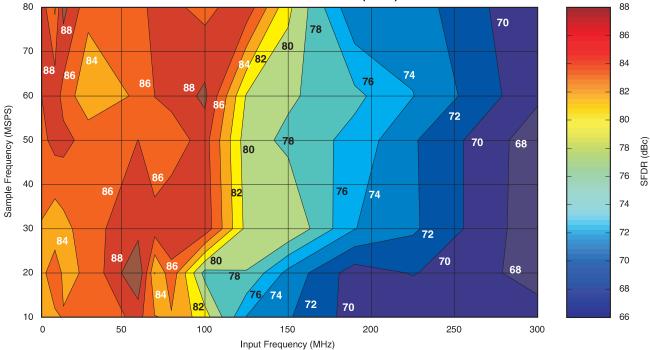


Figure 36.



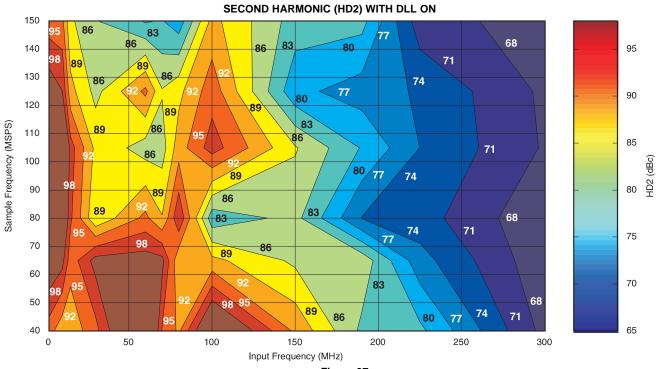


Figure 37.

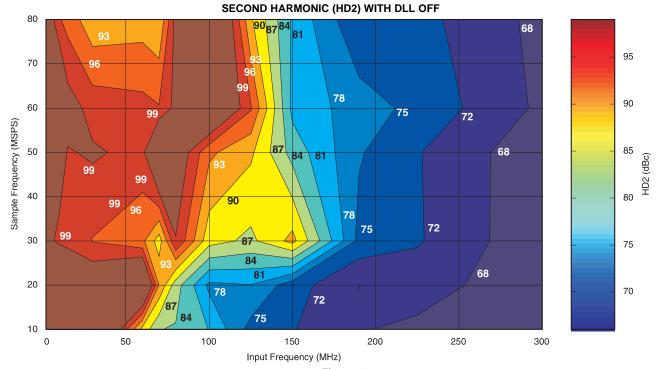
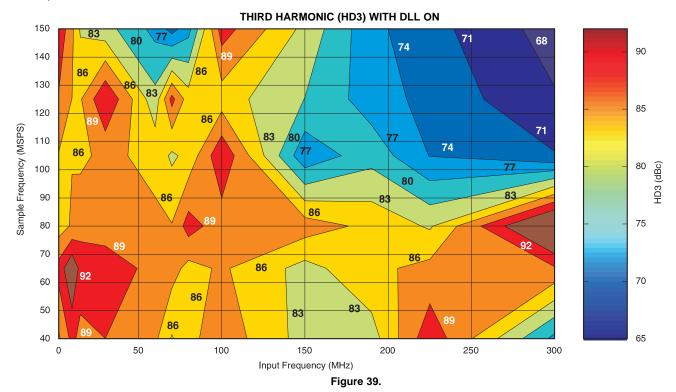


Figure 38.



Typical values are at  $T_A$  = 25°C,  $AV_{DD}$  = DRV<sub>DD</sub> = 3.3 V, differential input amplitude = -1 dBFS, sampling rate = 125 Msps, DLL On, and 3-V differential clock unless otherwise noted



THIRD HARMONIC (HD3) WITH DLL OFF Sample Frequency (MSPS) 84 81 HD3 (dBc) Input Frequency (MHz)

Figure 40.



#### APPLICATION INFORMATION

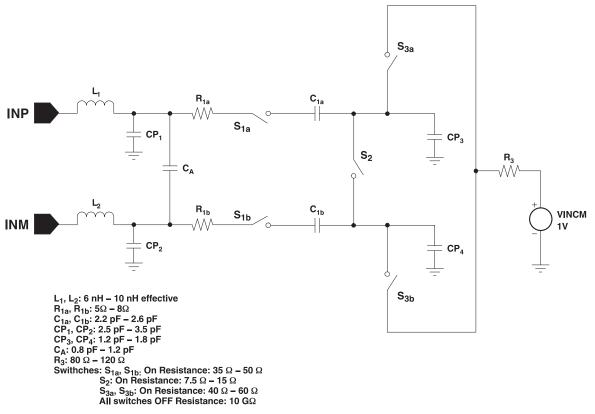
#### THEORY OF OPERATION

The ADS5500 is a low-power, 14-bit, 125 Msps, CMOS, switched capacitor, pipeline ADC that operates from a single 3.3-V supply. The conversion process is initiated by a falling edge of the external input clock. Once the signal is captured by the input S&H, the input sample is sequentially converted by a series of small resolution stages, with the outputs combined in a digital correction logic block. Both the rising and the falling clock edges are used to propagate the sample through the pipeline every half

clock cycle. This process results in a data latency of 17.5 clock cycles, after which the output data is available as a 14-bit parallel word, coded in either straight offset binary or binary two's complement format.e sample through the pipeline every half clock cycle. This process results

#### INPUT CONFIGURATION

The analog input for the ADS5500 consists of a differential sample-and-hold architecture implemented using a switched capacitor technique, shown in Figure 41.



NOTE: All Switches are ON in sampling phase which is approximately one half of a clock period.

#### Figure 41. Analog Input Stage

This differential input topology produces a high level of AC performance for high sampling rates. It also results in a high usable input bandwidth, especially important for high intermediate-frequency (IF) or undersampling applications. The ADS5500 requires each of the analog inputs (INP, INM) to be externally biased around the common-mode level of the internal circuitry (CM, pin 17). For a full-scale differential input, each of the differential lines of the input signal (pins 19 and 20) swings symmetrically between CM + 0.575 V and CM – 0.575 V. This means that each input is driven with a signal of up to CM 0.575 V, so that each input has a maximum

differential signal of 1.15  $V_{PP}$  for a total differential input signal swing of 2.3  $V_{PP}$ . The maximum swing is determined by the two reference voltages, the top reference (REFP, pin 29), and the bottom reference (REFM, pin 30).

The ADS5500 obtains optimum performance when the analog inputs are driven differentially. The circuit shown in Figure 42 shows one possible configuration using an RF transformer.



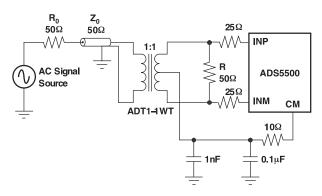


Figure 42. Transformer Input to Convert Single-Ended Signal to Differential Signal

The single-ended signal is fed to the primary winding of an RF transformer. Since the input signal must be biased around the common-mode voltage of the internal circuitry, the common-mode voltage ( $V_{CM}$ ) from the ADS5500 is connected to the center-tap of the secondary winding. To ensure a steady low-noise  $V_{CM}$  reference, best performance is obtained when the CM (pin 17) output is filtered to ground with 0.1  $\mu$ F and 0.001- $\mu$ F low-inductance capacitors.

Output  $V_{CM}$  (pin 17) is designed to directly drive the ADC input. When providing a custom CM level, be aware that the input structure of the ADC sinks a common-mode current in the order of 600  $\mu$ A (300  $\mu$ A per input) at 125 Msps. Equation 5 describes the dependency of the common-mode current and the sampling frequency:

$$\frac{600 \ \mu\text{A} \times \text{f}_{\text{s}}}{125\text{MSPS}} \tag{5}$$

Where:

 $f_S > 2$  Msps.

This equation helps to design the output capability and impedance of the driving circuit accordingly.

When it is necessary to buffer or apply a gain to the incoming analog signal, it is possible to combine single-ended operational amplifiers with an RF transformer, or to use a differential input/output amplifier without a transformer, to drive the input of the ADS5500. TI offers a wide selection of single-ended operational amplifiers (including the THS3201, THS3202, OPA847, and OPA695) that

can be selected depending on the application. An RF gain block amplifier, such as Tl's THS9001, can also be used with an RF transformer for very high input frequency applications. The THS4503 is a recommended differential input/output amplifier. Table 4 lists the recommended amplifiers.

When using single-ended operational amplifiers (such as the THS3201, THS3202, OPA847, or OPA695) to provide gain, a three-amplifier circuit is recommended with one amplifier driving the primary of an RF transformer and one amplifier in each of the legs of the secondary driving the two differential inputs of the ADS5500. These three amplifier circuits minimize even-order harmonics. For very high frequency inputs, an RF gain block amplifier can be used to drive a transformer primary; in this case, the transformer secondary connections can drive the input of the ADS5500 directly, as shown in Figure 42, or with the addition of the filter circuit shown in Figure 43.

Figure 43 illustrates how RIN and CIN can be placed to isolate the signal source from the switching inputs of the ADC and to implement a low-pass RC filter to limit the input noise in the ADC. It is recommended that these components be included in the ADS5500 circuit layout when any of the amplifier circuits discussed previously are used. The components allow fine-tuning of the circuit performance. Any mismatch between the differential lines of the ADS5500 input produces a degradation performance at high input frequencies, mainly characterized by an increase in the even-order harmonics. In this case, special care should be taken to keep as much electrical symmetry as possible between both inputs.

Another possible configuration for lower-frequency signals is the use of differential input/output amplifiers that can simplify the driver circuit for applications requiring dc-coupling of the input. Flexible in their configurations (see Figure 44), such amplifiers can be used for single-ended-to-differential conversion, signal amplification.



Table 4. Recommended Amplifiers to Drive the Input of the ADS5500

INPUT SIGNAL FREQUENCY	RECOMMENDED AMPLIFIER	TYPE OF AMPLIFIER	USE WITH TRANSFORMER?
DC to 20 MHz	THS4503	Differential In/Out Amp	No
DC to 50 MHz	OPA847	Operational Amp	Yes
DC to 100 MHz	THS4509	Differential In/Out Amp	NO
	OPA695	Operational Amp	Yes
10 MHz to 120 MHz	THS3201	Operational Amp	Yes
	THS3202	Operational Amp	Yes
Over 100 MHz	THS90016	RF Gain Block	Yes

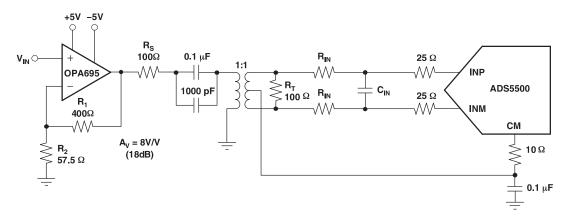


Figure 43. Converting a Single-Ended Input Signal to a Differential Signal Using an RF Transformer

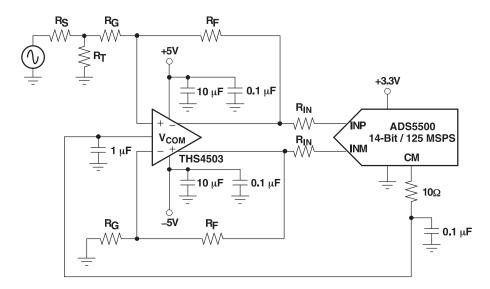


Figure 44. Using the THS4503 With the ADS5500

## **POWER SUPPLY SEQUENCING**

The preferred power-up sequence is to ramp  $AV_{DD}$  first, followed by  $DRV_{DD}$ , including a simultaneous ramp of  $AV_{DD}$  and  $DRV_{DD}$ . In the case that  $DRV_{DD}$  ramps up first in the system, care must be taken to ensure that  $AV_{DD}$  ramps up within 10 ms.

## **POWER DOWN**

The device enters power-down in one of two ways: either by reducing the clock speed or by setting the PDN bit through the serial programming interface. Using the reduced clock speed, power-down may be initiated for clock frequencies below 2 Msps. The exact frequency at which the power down occurs varies from device to device.



Using the serial interface PDN bit to power down the device places the outputs in a high-impedance state and only the internal reference remains on to reduce the power-up time. The power-down mode reduces power dissipation to approximately 180 mW.

#### REFERENCE CIRCUIT

The ADS5500 has built-in internal reference generation, requiring no external circuitry on the printed circuit board (PCB). For optimum performance, it is best to connect both REFP and REFM to ground with a 1- $\mu$ F decoupling capacitor (the 1- $\Omega$  series resistor shown in Figure 45 is optional). In addition, an external 56.2-k $\Omega$  resistor should be connected from IREF (pin 31) to AGND to set the proper current for the operation of the ADC, as shown in Figure 45. No capacitor should be connected between pin 31 and ground; only the 56.2-k $\Omega$  resistor should be used.

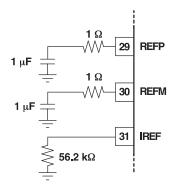


Figure 45. REFP, REFM, and IREF Connections for Optimum Performance

## **CLOCK INPUT**

The ADS5500 clock input can be driven with either a differential clock signal or a single-ended clock input, with little or no difference in performance between both configurations. The common-mode voltage of the clock inputs is set internally to CM (pin 17) using internal 5-k $\Omega$  resistors that connect CLKP (pin 10) and CLKM (pin 11) to CM (pin 17), as shown in Figure 46.

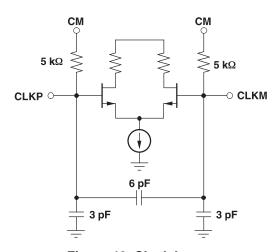


Figure 46. Clock Inputs

When driven with a single-ended CMOS clock input, it is best to connect CLKM (pin 11) to ground with a 0.01- $\mu F$  capacitor, while CLKP is ac-coupled with a 0.01- $\mu F$  capacitor to the clock source, as shown in Figure 47.

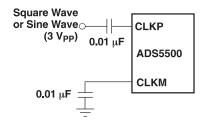


Figure 47. AC-Coupled, Single-Ended Clock Input

The ADS5500 clock input can also be driven differentially, reducing susceptibility to common-mode noise. In this case, it is best to connect both clock inputs to the differential input clock signal with 0.01- $\mu$ F capacitors, as shown in Figure 48.

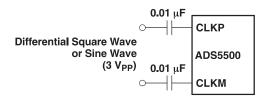


Figure 48. AC-Coupled, Differential Clock Input

For high input frequency sampling, it is recommended to use a clock source with very low jitter. Additionally, the internal ADC core uses both edges of the clock for the conversion process. This means that, ideally, a 50% duty cycle should be provided. Figure 49 shows the performance variation of the ADC versus clock duty cycle.



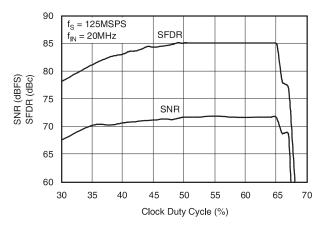


Figure 49. AC Performance vs Clock Duty Cycle

Bandpass filtering of the clock source can help produce a 50% duty cycle clock and reduce the effect of jitter. When using a sinusoidal clock, the clock jitter further improves as the amplitude is increased. In that sense, using a differential clock allows for the use of larger amplitudes without exceeding the supply rails and absolute maximum ratings of the ADC clock input. Figure 50 shows the performance variation of the device versus input clock amplitude. For detailed clocking schemes based on transformer or PECL-level clocks, see the ADS5500EVM user's guide (SLWU010), available for download from www.ti.com.

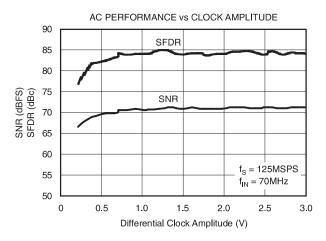


Figure 50. AC Performance vs Clock Amplitudes

#### **INTERNAL DLL**

In order to obtain the fastest sampling rates achievable with the ADS5500, the device uses an internal digital delay lock loop (DLL). Nevertheless, the limited frequency range of operation of DLL degrades the performance at clock frequencies below 60 Msps. In order to operate the device below 60 Msps, the internal DLL must be shut off using the

DLL OFF mode described in the Serial Interface Programming section. The *Typical Performance Curves* show the performance obtained in both modes of operation: DLL ON (default) and DLL OFF. In either of the two modes, the device enters power-down mode if no clock or slow clock is provided. The limit of the clock frequency where the device functions properly with default settings is ensured to be over 2 MHz.

#### **OUTPUT INFORMATION**

The ADC provides 14 data outputs (D13 to D0, with D13 being the MSB and D0 the LSB), a data-ready signal (CLKOUT, pin 43), and an out-of-range indicator (OVR, pin 64) that equals one when the output reaches the full-scale limits.

Two different output formats (straight offset binary or two's complement) and two different output clock polarities (latching output data on rising or falling edge of the output clock) can be selected by setting DFS (pin 40) to one of four different voltages. Table 3 details the four modes. In addition, output enable control (OE, pin 41, active high) is provided to put the outputs into a high-impedance state.

In the event of an input voltage overdrive, the digital outputs go to the appropriate full scale level. For a positive overdrive, the output code is 0x3FFF in straight offset binary output format, and 0x1FFF in 2's complement output format. For a negative input overdrive, the output code is 0x0000 in straight offset binary output format and 0x2000 in two's complement output format. These outputs to an overdrive signal are ensured through design and characterization.

The output circuitry of the ADS5500, by design, minimizes the noise produced by the data switching transients and, in particular, its coupling to the ADC analog circuitry. Output D4 (pin 51) senses the load capacitance and adjusts the drive capability of all the output pins of the ADC to maintain the same output slew rate described in the timing diagram of Figure 1. Care should be taken to ensure that all output lines (including CLKOUT) have nearly the same load as D4 (pin 51). This circuit also reduces the sensitivity of the output timing versus supply voltage or temperature. Placing external resistors in series with the outputs is not recommended.

The timing characteristics of the digital outputs change for sampling rates below the 125 Msps maximum sampling frequency. Table 5 through Table 7 show the values of various timing parameters for lower sampling frequencies, both with DLL on and off.



To use the input clock as the data capture clock, it is necessary to delay the input clock by a delay  $(t_d)$  that results in the desired setup or hold time. Use either of the following equations to calculate the value of  $t_d$ .

Desired setup time =  $t_d$ -  $t_{START}$ Desired hold time =  $t_{END}$ -  $t_d$ 

Table 5. Timing Characteristics at Additional Sampling Frequencies (DLL ON)

F <sub>s</sub> (Msps)	t <sub>su</sub> (ns)		t <sub>h</sub> (ns)		t <sub>START</sub> (ns)				t <sub>END</sub> (ns)		tr (ns)			t <sub>f</sub> (ns)				
r <sub>S</sub> (Wisps)	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX
105	2.2	2.8		2.2	2.5			1.9	2.8	5.8	7.3			4.4	5.1		3.3	3.8
80	2.8	3.7		2.8	3.3			0.5	1.7	5.3	7.9			5.8	6.6		4.4	5.3
65	3.8	4.6		3.6	4.1			-0.5	0.8	5.3	8.5			6.7	7.2		5.5	6.4

Table 6. Timing Characteristics at Additional Sampling Frequencies (DLL OFF)

F <sub>s</sub> (Msps)	t <sub>su</sub> (ns)		t <sub>h</sub> (ns)			t <sub>START</sub> (ns	)		t <sub>END</sub> (ns)			tr (ns)		t <sub>f</sub> (ns)				
r <sub>S</sub> (IVISPS)	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX
80	3.2	4.2		1.8	3			3.8	5	8.4	11			5.8	6.6		4.4	5.3
65	4.3	5.7		2	3			2.8	4.5	8.3	11.8			6.6	7.2		5.5	6.4
40	8.5	11		2.6	3.5			-1	1.5	8.9	14.5			7.5	8		7.3	7.8
20	17	25.7		2.5	4.7			-9.8	2	9.5	21.6			7.5	8		7.6	8
10	27	51		4	6.5			-30	-3	11.5	31							
2	284	370		8	19			185	320	515	576			50	82		75	150

Table 7. Timing Characteristics at Additional Sampling Frequencies (DLL ON)

F <sub>s</sub> (Msps)	CLKOUT, Rise Time t <sub>r</sub> (ns)			CLKOUT, Fall Time $t_f$ (ns)			CLKOUT	Jitter, Peak-to-Pe	eak t <sub>JIT</sub> (ps)	Input-to-Output Clock Delay t <sub>PDI</sub> (ns)			
	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
105		2	2.2		1.7	1.8		175	250	4	4.7	5.5	
80		2.5	2.8		2.1	2.3		210	315	3.7	4.3	5.1	
65		3.1	3.5		2.6	2.9		260	380	3.5	4.1	4.8	

Table 8. Timing Characteristics at Additional Sampling Frequencies (DLL OFF)

F <sub>s</sub> (Msps)	CLKOUT, Rise Time t <sub>r</sub> (ns)			CLKOUT, Fall Time t <sub>f</sub> (ns)			CLKOUT J	itter, Peak-to-P	eak t <sub>JIT</sub> (ps)	Input-to-Output Clock Delay t <sub>PDI</sub> (ns)			
	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
80		2.5	2.8		2.1	2.3		210	315	7.1	8	8.9	
65		3.1	3.5		2.6	2.9		260	380	7.8	8.5	9.4	
40		4.8	5.3		4	4.4		445	650	9.5	10.4	11.4	
20		8.3	9.5		7.6	8.2		800	1200	13	15.5	18	
10										16	20.7	25.5	
2		31	52		36	65		2610	4400	537	551	567	

#### SERIAL PROGRAMMING INTERFACE

The ADS5500 has internal registers for the programming of some of the modes described in the previous sections. The registers should be reset after power-up by applying a 2  $\mu s$  (minimum) high pulse on RESET (pin 35); this also resets the entire ADC and sets the data outputs to low. This pin has a 200-k $\Omega$  internal pullup resistor to AV $_{DD}$ . The programming is done through a three-wire interface. The timing diagram and serial register setting in the Serial Programing Interface section describe the programming of this register.

Table 2 shows the different modes and the bit values to be written on the register to enable them.

Note that some of these modes may modify the standard operation of the device and possibly vary the performance with respect to the typical data shown in this data sheet.

Applying a RESET signal is required to set the internal registers to their default states for normal operation. If the hardware RESET function is not used in the system, the RESET pin must be tied to ground and it is necessary to write the default values to the internal registers through the serial programming interface. The registers must be written in the following order.

Write 9000h (Address 9, Data 000) Write A000h (Address A, Data 000) Write B000h (Address B, Data 000) Write C000h (Address C, Data 000)



Write D000h (Address D, Data 000) Write E000h (Address E, Data 804) Write 0000h (Address 0, Data 000) Write 1000h (Address 1, Data 000) Write F000h (Address F, Data 000).

NOTE: This procedure is only required if a RESET pulse is not provided to the device.

#### PowerPAD PACKAGE

The PowerPAD package is a thermally enhanced standard size IC package designed to eliminate the use of bulky heat sinks and slugs traditionally used in thermal packages. This package can be easily mounted using standard printed circuit board (PCB) assembly techniques and can be removed and replaced using standard repair procedures. AVDD. The programming is done through a three-wire interface. The timing diagram and serial register setting in the Serial Programing Interface section describe the programming of this register.

The PowerPAD package is designed so that the lead frame die pad (or thermal pad) is exposed on the bottom of the IC. This provides a low thermal resistance path between the die and the exterior of the package. The thermal pad on the bottom of the IC can then be soldered directly to the printed circuit board (PCB), using the PCB as a heatsink. Programming is done through a three-wire interface. The timing diagram and serial register setting in the Serial Programing Interface section describe the programming of this register.

#### **Assembly Process**

 Prepare the PCB top-side etch pattern including etch for the leads as well as the thermal pad as illustrated in the Mechanical

- Data section. The recommended thermal pad dimension is 8 mm x 8 mm.
- Place a 5-by-5 array of thermal vias in the thermal pad area. These holes should be 13 mils in diameter. The small size prevents wicking of the solder through the holes.
- It is recommended to place a small number of 25 mil diameter holes under the package, but outside the thermal pad area to provide an additional heat path.
- 4. Connect all holes (both those inside and outside the thermal pad area) to an internal copper plane (such as a ground plane).
- 5. Do not use the typical web or spoke via connection pattern when connecting the thermal vias to the ground plane. The spoke pattern increases the thermal resistance to the ground plane.
- The top-side solder mask should leave exposed the terminals of the package and the thermal pad area.
- 7. Cover the entire bottom side of the PowerPAD vias to prevent solder wicking.
- 8. Apply solder paste to the exposed thermal pad area and all of the package terminals.

For more detailed information regarding the PowerPAD package and its thermal properties, see either the SLMA004B application brief PowerPAD Made Easy or SLMA002 technical brief PowerPAD Thermally Enhanced Package.



## **REVISION HISTORY**

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#### PACKAGING INFORMATION

Orderable part number	Status (1)	Material type	Package   Pins	Package qty   Carrier	<b>RoHS</b> (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
ADS5500IPAP	Active	Production	HTQFP (PAP)   64	160   JEDEC TRAY (10+1)	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	ADS5500I
ADS5500IPAP.A	Active	Production	HTQFP (PAP)   64	160   JEDEC TRAY (10+1)	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	ADS5500I
ADS5500IPAPG4	Active	Production	HTQFP (PAP)   64	160   JEDEC TRAY (10+1)	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	ADS5500I
ADS5500IPAPG4.A	Active	Production	HTQFP (PAP)   64	160   JEDEC TRAY (10+1)	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	ADS5500I

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

## PACKAGE OPTION ADDENDUM

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#### OTHER QUALIFIED VERSIONS OF ADS5500:

● Enhanced Product : ADS5500-EP

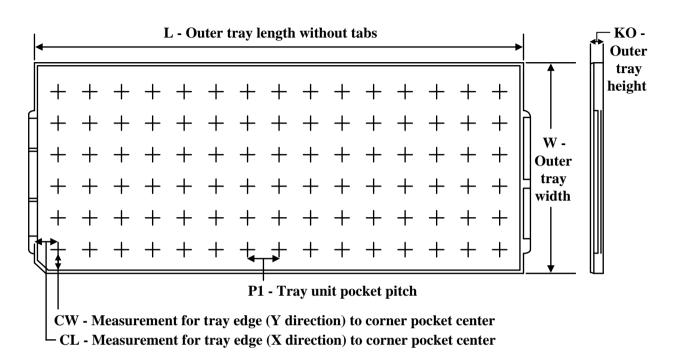
NOTE: Qualified Version Definitions:

• Enhanced Product - Supports Defense, Aerospace and Medical Applications



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## **TRAY**



Chamfer on Tray corner indicates Pin 1 orientation of packed units.

#### \*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	Unit array matrix	Max temperature (°C)	L (mm)	W (mm)	Κ0 (μm)	P1 (mm)	CL (mm)	CW (mm)
ADS5500IPAP	PAP	HTQFP	64	160	8 x 20	150	315	135.9	7620	15.2	13.1	13
ADS5500IPAP.A	PAP	HTQFP	64	160	8 x 20	150	315	135.9	7620	15.2	13.1	13
ADS5500IPAPG4	PAP	HTQFP	64	160	8 x 20	150	315	135.9	7620	15.2	13.1	13
ADS5500IPAPG4.A	PAP	HTQFP	64	160	8 x 20	150	315	135.9	7620	15.2	13.1	13

10 x 10, 0.5 mm pitch

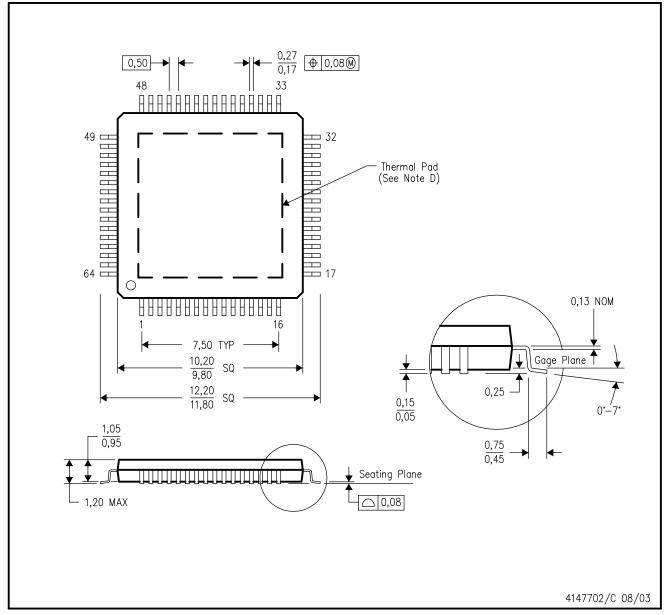
QUAD FLATPACK

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.



# PAP (S-PQFP-G64)

## PowerPAD™ PLASTIC QUAD FLATPACK



NOTES:

- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <a href="https://www.ti.com">www.ti.com</a>.
- E. Falls within JEDEC MS-026

PowerPAD is a trademark of Texas Instruments.



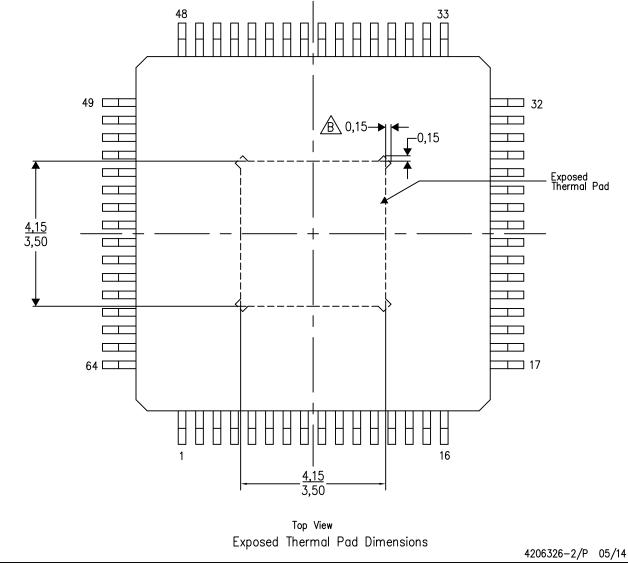
PowerPAD™ PLASTIC QUAD FLATPACK

## THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTES: A. All linear dimensions are in millimeters

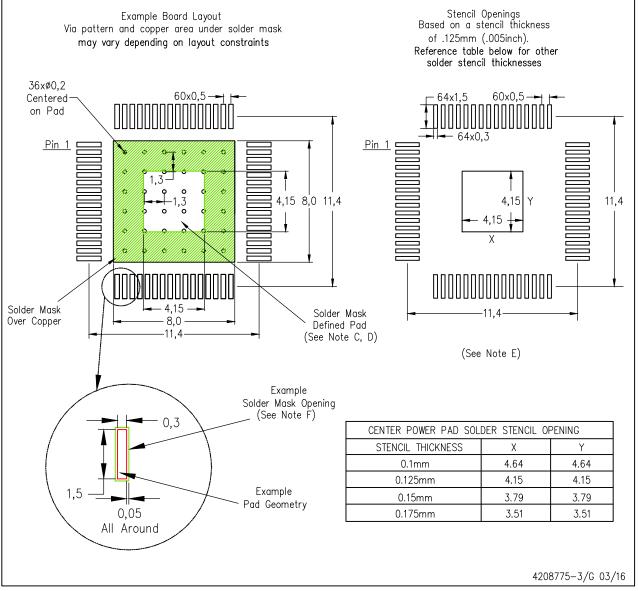
(B) Tie strap features may not be present.

PowerPAD is a trademark of Texas Instruments



# PAP (S-PQFP-G64)

## PowerPAD™ PLASTIC QUAD FLATPACK



NOTES:

- A. All linear dimensions are in millimeters.
- 3. This drawing is subject to change without notice.

- PowerPAD is a trademark of Texas Instruments
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">www.ti.com</a>. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
- F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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